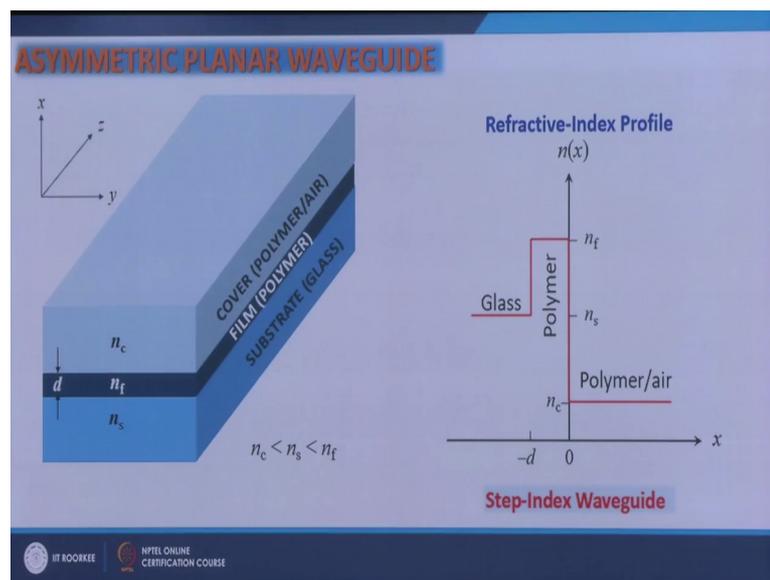


Fiber Optics
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Lecture - 17
Electromagnetic Analysis of Waveguides – VII

Till now we have carried out the analysis of symmetric planar wave guides where the high index region is sandwiched between 2 lower index regions of same refractive index. Now in this lecture we will extend the analysis to asymmetric planar waveguide asymmetric planar waveguides are more practical wave guides. So, we will extend this analysis to asymmetric planar waveguide and see how the waveguide hence is done in these kind of waveguides.

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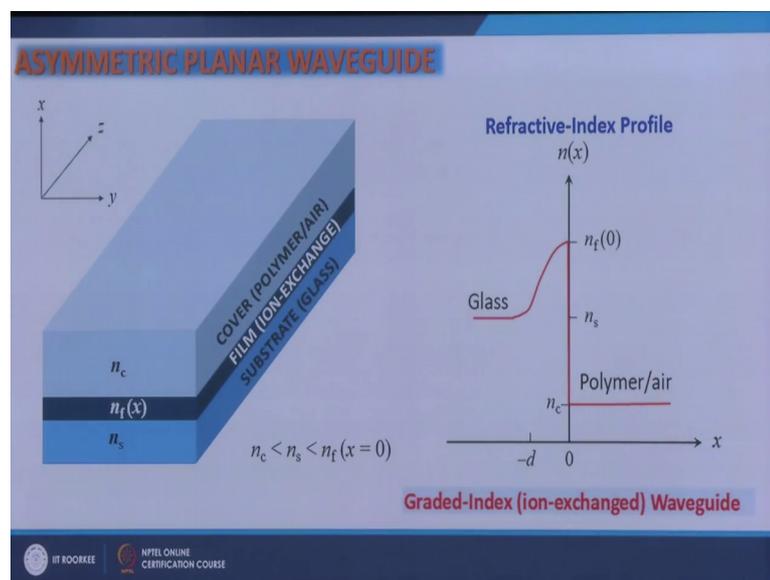
So, what is an asymmetric planar waveguide you for example, take a glass of substrate of refractive index n_s , and you deposit a film of refractive index n_f of thickness d , this film can be for example, of a polymer material, the refractive index n_f is higher than the refractive index n_s and then it is surrounded by air. So, air works as cover; also you can put a cover of polymer material itself of refractive index n_c , where n_c is smaller than n_s and n_s is smaller than n_f .

So, this kind of waveguide we can see that it has asymmetric structure, because n_s is not equal to n_c here. If I look at the refractive index profile of this waveguide, then this is

the substrate an example is glass, this is the guiding film the example is polymer, and this is the cover region which can be air or another polymer of lower refractive index. Here at x is equal to 0, I have made the interface between the film and the cover, and at x is equal to minus d represents the interface between the substrate and the film.

Since the refractive indices of the individual layers are uniform. So, this waveguide is step index waveguide. I can also have a waveguide in which the refractive index in the guiding film is not uniform, but it varies with x .

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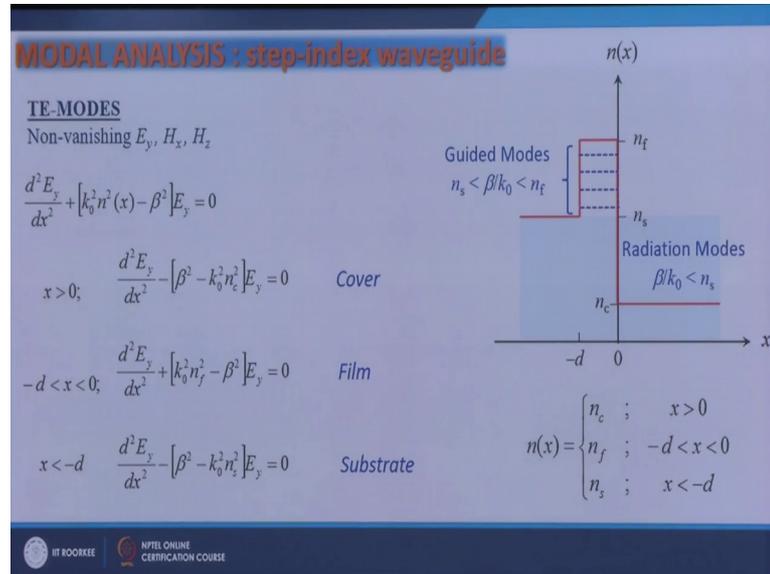


So, how I can make this waveguide well I take a substrate of material glass, and then I diffuse certain ions into this pay over a certain distance. So, I will have a diffusion profile for example, I can diffuse silver ions into glass and that will change the refractive index of the material near the surface.

So, now this n_f would be a function of x now, and then I can have cover as air or cover of a polymer layer. So, now, this n_c is smaller than n_s , and then the refractive index of the film at the surface of the film is n_f at f is equal to 0. So, n_s is smaller than that. If I look at the refractive index profile of this structure now, so, I have a substrate and this is the guiding film. This guiding film now has refractive index with varies which varies with x and ultimately it merges into the refractive index of glass.

So, this is the surface index n_f , and this n_f is larger than n_s and which is larger than n_c .

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So, this is a typical graded index ion exchange waveguide. Let us do the modal analysis of this step index waveguide, the step index planar asymmetric waveguide. So, this is the refractive index profile and I can write down the refractive index in various regions as $n(x)$ is equal to n_c , which is defined by the region x greater than 0 which is the cover, then in the region minus d less than x less than 0 the refractive index is n_f which is the film region, and for x smaller than minus d which represents the substrate region, the refractive index is n_s .

Now, the modes of this waveguide are here guided modes, whose refractive index lies between n_s and n_f . So, these are some representative guided modes, and for β/k_0 smaller than n_s , the field would radiate out in the substrate region and for β/k_0 is smaller than n_c the field would radiate out in the cover region as well. So, in this region you will have radiation modes.

We are interested in the guided modes of the structure, which are defined in this range, and let us first do the TE mode analysis of the waveguide, and as I can see that the refractive index variation is in x direction, and I am considering propagation in z direction. So, the non vanishing components corresponding to TE modes are E_y, H_x and H_z .

I know from my previous analysis that the wave equation satisfied by TE modes of such a waveguide is given by $\frac{d^2 E_y}{dx^2} + k_0^2 n^2 E_y - \beta^2 E_y = 0$. So, the procedure is again the same I write down this equation in all the three regions, I write down the solutions of these three equations and then apply boundary conditions to match the solutions.

So, this is the equation in cover region $x > 0$, which is $\frac{d^2 E_y}{dx^2} + k_0^2 n_c^2 E_y - \beta^2 E_y = 0$, because in this region n_x is equal to n_c , and I have taken minus sign outside and represented it in this form because β/k_0 is greater than n_c , so that this quantity is positive for guided modes.

In the film region which is defined by $-d < x < 0$, the equation becomes $\frac{d^2 E_y}{dx^2} + k_0^2 n_f^2 E_y - \beta^2 E_y = 0$, again for guided modes this quantity in the square brackets is positive. In the substrate region defined by $x < -d$, the equation is $\frac{d^2 E_y}{dx^2} + k_0^2 n_s^2 E_y - \beta^2 E_y = 0$ and again for guided modes this quantity is positive.

So, now I can represent this as γ_c^2 , this as κ_f^2 and this as γ_s^2 and then find out what are the solutions of these three equations.

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Solutions		
Cover	$x > 0;$	$\frac{d^2 E_y}{dx^2} - \gamma_c^2 E_y = 0$ $E_y(x) = A \exp(-\gamma_c x)$
Film	$-d < x < 0;$	$\frac{d^2 E_y}{dx^2} + \kappa_f^2 E_y = 0$ $E_y(x) = B \exp(i\kappa_f x) + C \exp(-i\kappa_f x)$
Substrate	$x < -d$	$\frac{d^2 E_y}{dx^2} - \gamma_s^2 E_y = 0$ $E_y(x) = D \exp(\gamma_s x)$

where, $\gamma_c^2 = \beta^2 - k_0^2 n_c^2$, $\kappa_f^2 = k_0^2 n_f^2 - \beta^2$, $\gamma_s^2 = \beta^2 - k_0^2 n_s^2$

A, B, C and D are determined by boundary conditions

So, these are the three equations now, where γ_c^2 is equal to β^2 minus $k_0^2 n_c^2$, κ_f^2 is $k_0^2 n_f^2$ minus β^2 , and γ_s^2 is β^2 minus $k_0^2 n_s^2$.

So, what are the solutions? The solution of this differential equation is $A e^{-\gamma_c x}$ and $e^{\gamma_c x}$. And as you know that we are interested in guided modes. So, we cannot have exponentially amplifying solution. So, I am only considering the exponentially decaying solutions in cover regions and substrate region.

Now, in the film this equation will give you the oscillatory solutions, which I write as $B e^{i\kappa_f x}$ plus $C e^{-i\kappa_f x}$, and in the substrate region it is $D e^{\gamma_s x}$ because x is smaller than $-d$. So, it is negative. So, this again gives you exponentially decaying solutions. Now a , b , c and d are constants and these constants are determined by the boundary conditions.

So, let us apply boundary conditions and obtain relationships between a , b , c , d and obtain the Eigen value equation or characteristic equation which is satisfied by the propagation constants of the guided modes.

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Region: $x < -d$ (substrate) $-d < x < 0$ (film) $x > 0$ (cover)

Solution: $E_y(x) = D \exp(\gamma_s x)$ $E_y(x) = B \exp(i\kappa_f x) + C \exp(-i\kappa_f x)$ $E_y(x) = A \exp(-\gamma_c x)$

Boundary conditions: E_y and $\frac{dE_y}{dx}$ are continuous at $x = 0$ and at $x = -d$

E_y at $x = 0$: $B + C = A$

E_y at $x = -d$: $B \exp(-i\kappa_f d) + C \exp(i\kappa_f d) = D \exp(-\gamma_s d)$

E'_y at $x = 0$: $i\kappa_f B - i\kappa_f C = -\gamma_c A$

E'_y at $x = -d$: $i\kappa_f B \exp(-i\kappa_f d) - i\kappa_f C \exp(i\kappa_f d) = \gamma_s D \exp(-\gamma_s d)$

$\tan(\kappa_f d) = \frac{\frac{\gamma_s}{\kappa_f} + \frac{\gamma_c}{\kappa_f}}{1 - \frac{\gamma_s \gamma_c}{\kappa_f \kappa_f}}$

Eigenvalue equation

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So, these are the solutions in different regions in substrate in the film and in the cover and what are the boundary conditions? Boundary conditions are again the tangential

components of e and h are continuous, and the tangential components here are E_y and h_z . So, they give me that E_y and dE_y/dx because h_z is related to E_y as dE_y/dx . So, they are continuous at the interfaces x is equal to 0 and at x is equal to minus t .

So, let us apply these boundary conditions. So, if I look at the continuity of E_y at x is equal to 0. So, I apply it here. So, if I approach from the film and if I approach from the cover towards x is equal to 0. So, I will get $B + C$ is equal to A . Now continuity of E_y at x is equal to minus d which is the substrate film interface, this gives me $B e^{-\gamma_c d}$ to the power x is equal to minus d . So, $B e^{-\gamma_c d}$ plus $C e^{-\gamma_c d}$ is equal to D , $D e^{-\gamma_s d}$ to the power minus $\gamma_s d$.

Now, let us see the continuity of the derivatives at x is equal to 0 and at x is equal to minus t . So, E_y' at x is equal to 0 gives me $I \kappa_f B - I \kappa_f C$ is equal to minus $\gamma_c A$ and continuity of E_y' at x is equal to minus d interface gives me $I \kappa_f B e^{-\gamma_c d}$ to the power minus $I \kappa_f d$ minus $I \kappa_f C e^{-\gamma_c d}$ to the power $I \kappa_f d$ is equal to $\gamma_s d e^{-\gamma_s d}$.

So, I have now four equations relating these four constants a , b , c and d , and by mathematical manipulation of these four equations I can do 2 things one is obtain b , c and d in terms of a , and then eliminating a , b , c , d all together and form a transcendental equation in β which is also known as Eigen value equation or characteristic equation. So, that equation comes out to be $\tan \kappa_f d$ is equal to γ_s over κ_f plus γ_c over κ_f divided by $1 - \gamma_c$ over κ_f times γ_c over κ_f .

So, this is the equation which is satisfied by β because the only unknown here is β , β appears in κ_f , γ_s and γ_c . So, if I solve this equation and find out the roots of this equation then I can know what are the propagation constants of the modes, what are the modes, which are supported by this asymmetric step index planar waveguide.

As we have done earlier also that it is always a good idea to obtain everything in normalized parameters. So, that we are more or less independent of waveguide parameters, here we cannot have complete independence from waveguide parameters, but to certain extent it is still we can draw some universal curve. So, how do we define normalized parameters here? First is normalized frequency v .

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Normalized Parameters

$$V = \frac{2\pi d}{\lambda_0} \sqrt{n_f^2 - n_s^2}$$

$$b = \frac{(\beta/k_0)^2 - n_s^2}{n_f^2 - n_s^2}$$

$$a = \frac{n_s^2 - n_c^2}{n_f^2 - n_s^2}$$

$$b = \frac{\beta^2 - k_0^2 n_s^2}{k_0^2 (n_f^2 - n_s^2)} \Rightarrow 1 - b = \frac{k_0^2 n_f^2 - \beta^2}{k_0^2 (n_f^2 - n_s^2)} = \frac{\kappa_f^2}{V^2 / (d/2)^2} = \frac{(\kappa_f d/2)^2}{V^2}$$

$$\therefore \frac{\kappa_f d}{2} = V \sqrt{1 - b} \quad \& \quad \frac{\gamma_s d}{2} = V \sqrt{b}$$

$$b + a = \frac{\beta^2 - k_0^2 n_c^2}{k_0^2 (n_f^2 - n_s^2)} = \frac{\gamma_c^2}{V^2 / (d/2)^2} \quad \therefore \frac{\gamma_c d}{2} = V \sqrt{b + a}$$

If you remember that in symmetric planar waveguide we had only 2 refractive indices n_1 and n_2 , n_1 is greater than n_2 . So, there we had defined it in terms of n_1 square minus n_2 square.

Now, now in place of n_1 I have n_f , but low refractive indices are now up to n_c and n_s which one we should take in order to define V . So, it is very simple to choose because n_s defines the cut-off of guided modes. So, I should choose n_s because as soon as the effective index of a mode falls below n_s it is no more guided. So, I choose n_s to define the normalized frequency v here, and similarly for defining the normalized propagation constant b I choose n_s .

So, b is defined as β over k naught square minus n_s square over n_f square minus n_s square. There is another parameter here in this structure, which tells you how asymmetric this waveguide is. The asymmetry in the structure is introduced by the different values of n_s and n_c . So, what is the difference between n_s and n_c ? So, let us define by asymmetry parameter a , which is given as n_s square minus n_c square divided by n_f square minus n_s square.

So, now I have here three normalized parameters in place, and now I can represent my transcendental equation or Eigen value equation in terms of these normalized parameters. If you remember that in the transcendental equation the three terms appear which are κ_f , γ_s and γ_c , which contain β .

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The image shows handwritten mathematical derivations and a diagram of a step-index optical fiber cross-section. The diagram is a rectangle with a top horizontal line labeled n_f and a bottom horizontal line labeled n_c . A vertical dashed line is drawn from the top line down to the bottom line, representing the core-cladding interface. The left vertical line is labeled n_s . The derivations are as follows:

$$\kappa_f^2 = k_0^2 n_f^2 - \beta^2$$

$$\gamma_s^2 = \beta^2 - k_0^2 n_s^2$$

$$\gamma_c^2 = \beta^2 - k_0^2 n_c^2$$

$$b = \frac{\beta^2/k_0^2 - n_s^2}{n_f^2 - n_s^2}$$

$$\beta/k_0 < n_f \quad 0 < b < 1$$

$$V = \frac{2\pi}{\lambda_0} \frac{d}{2} \sqrt{n_f^2 - n_s^2}$$

$$V = \frac{2\pi}{\lambda_0} \frac{d}{2} \sqrt{n_f^2 - n_s^2}$$

So, now I will have to represent these kappa f gamma s and gamma c in terms of v b and a. And we can see that kappa f square is given by k naught square n f square, minus beta square. Gamma s square is given by beta square minus k naught square and s square, and gamma c square is given by beta square minus k naught square and c square.

So, keeping these in mind and looking at the expressions here I would now try to obtain kappa f gamma c and gamma s in terms of v b and a. So, let us first try for kappa f, kappa f has k naught square n f square minus beta square, and I can see that if I do 1 minus b then I can obtain this k naught square n f square minus beta square in the numerator of this. So, b is given by this. So, if I do 1 minus b it becomes k naught square n f square minus beta square, divided by k naught square n f square minus n s square, and 2 pi over lambda naught is k naught.

So, if I multiple here by d by 2 and d by 2, then I can have kappa f square here or I have kappa f square directly here and this is nothing, but v square divided by d by 2 square. So, this becomes kappa f d by 2 whole square divided by v square. So, this gives me kappa f d by 2 kappa f d by 2 is equal to v times square root of 1 minus b.

If I look at gamma s square, gamma s square is b dash square minus k naught square n s square. So, it is directly appearing here beta square minus k naught square n s square divided by k naught square this thing. So, this will immediately give me gamma as d by 2 is equal to v times square root of b. What about now gamma c square, which is defined

as beta square minus k naught square n c square. So, I can see from here that if I do b plus a then I can only obtain gamma c square in the numerator. So, I do b plus a. So, it becomes beta square minus k naught square n c square divided by this. So, which is gamma c square over v square divided by d by 2 whole square, and this gives me gamma c d by 2 is equal to v times square root of b plus a.

So, in this way I have now obtained kappa f gamma c and gamma s in terms of v b and a. So, I can now represent my Eigen value equation in normalized parameters. So, this equation simply becomes tan 2 v square root of 1 minus b, is equal to square root of b divided by 1 minus b. Which is coming from gamma s term, plus b plus a divided by 1 minus b square root which is coming from the cover term gamma c term. Divided by 1 minus square root of b divided by 1 minus b and the square root of b plus a divided by 1 minus b.

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Normalized Parameters

$$\frac{\kappa_f d}{2} = V\sqrt{1-b}, \quad \frac{\gamma_s d}{2} = V\sqrt{b}, \quad \frac{\gamma_c d}{2} = V\sqrt{b+a}$$

Eigenvalue equation

$$\tan(\kappa_f d) = \frac{\frac{\gamma_s + \gamma_c}{\kappa_f} \frac{\kappa_f}{\kappa_f}}{1 - \frac{\gamma_s \gamma_c}{\kappa_f \kappa_f}} \Rightarrow \tan(2V\sqrt{1-b}) = \frac{\sqrt{\frac{b}{1-b}} + \sqrt{\frac{b+a}{1-b}}}{1 - \sqrt{\frac{b}{1-b}} \sqrt{\frac{b+a}{1-b}}}$$

Cut-offs: $\beta = k_0 n_s$ or $b = 0$

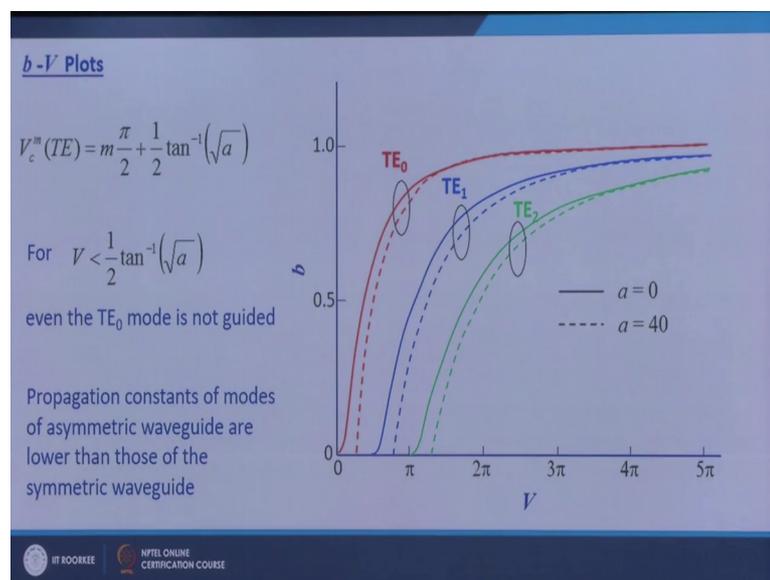
$$\tan(2V_c) = \sqrt{a} \quad \text{or} \quad V_c^m(TE) = m \frac{\pi}{2} + \frac{1}{2} \tan^{-1}(\sqrt{a})$$

And of course, since b is equal to b dash square over k naught square, minus n s square divided by n f square, minus n s square and for cut off beta over k naught is equal to n s or for guided volts, beta over k naught lies between n s and n f. So, b would lie as usual between 0 and 1. So, I solve this equation for a given value of v, v is if you remember v is defined as 2 pi over lambda naught times d by 2, times square root of n f square minus n s square. So, this v contains all the waveguide parameters and the wavelength. So, for a

given waveguide and wavelength I have v and for that value of v I can solve this equation to obtain the normalized propagation constant b .

What are the cut offs? Cut offs are defined by β is equal to $k_0 n_s$ or b is equal to 0. So, if I put it there then the cut off equation becomes $\tan 2V_c$ is equal to square root of a or cut off for TE mode m th TE mode is now given by $V_c m$ is equal to $m\pi$ by 2 plus half $\tan^{-1} a$ square root of a . You can see here that the cut off of even TE 0 mode is finite.

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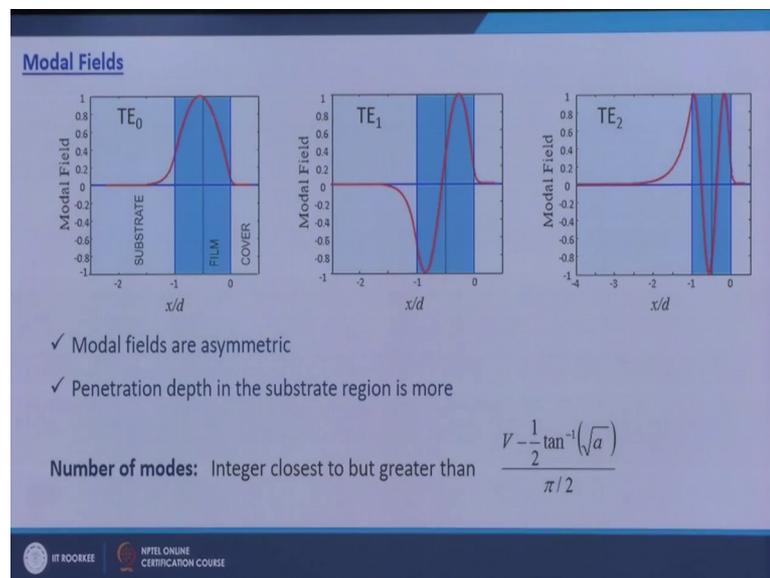
So, in a particular range of V which is defined by V less than half $\tan^{-1} a$ even the TE_0 mode is not guided.

This is the difference as compared to the symmetric planar waveguide. In symmetric planar waveguide TE_0 mode has 0 cut off. So, TE_0 mode was always guided, but here it is not the case. Now let us solve this equation the characteristic equation or Eigen value equation for different values of V , and plot the roots as obtained as a function of v . So, here I have plotted the roots for both the cases a is equal to 0 which represents the symmetric planar waveguide, and then for a very high value of a which is 40 for asymmetric planar waveguide. These solid lines are corresponding to symmetric planar waveguide a is equal to 0, and these dash lines are corresponding to asymmetric planar waveguide

And what I can see that t_0 mode has cut off here, t_1 mode has cut off here, t_2 mode has cut off here. In case of symmetric planar waveguide this was the cut off v_{cm} is equal to $m\pi/2$ was the corresponding cut off. Now all the cut offs have been shown shifted by this much amount which is half tan inverse of square root of a . So, this is one thing another thing that I see here is that if I take a particular value of v then the propagation constant of the mode is now smaller than the propagation constant of the corresponding symmetric waveguide mode.

And it is understandable that because asymmetric is introduced by introducing the cover region. So, if this is the symmetric waveguide, this is the symmetric waveguide. So, this is n_s this is n_f this is again the level n_s , but now in asymmetric planar waveguide I have n_c . So, I am reducing the refractive index in the cover region. So, the effect is to pull down the effective indices of the modes towards lower side so that we can see from here itself that the propagation constants of asymmetric planar waveguides are smaller than those corresponding to symmetric planar waveguide.

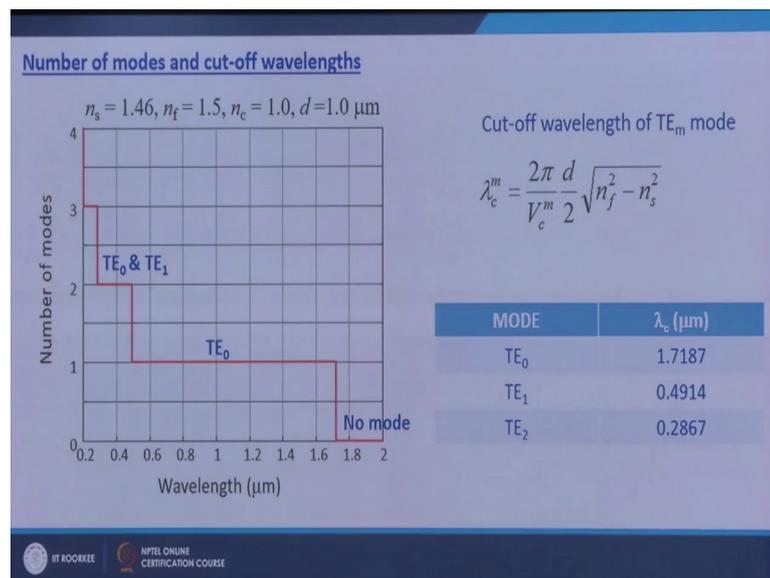
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Let us look at modal fields. So, these are the modal fields of TE_0 , TE_1 and TE_2 modes of a typical asymmetric planar waveguide. So, I can see that the modal fields are asymmetric and you can see that the penetration gap is more in the substrate and less in the cover which is understandable, because here at this interface the index contrast is smaller as compared to the index contrast at this interface ok.

So, the field extends more into the substrate region as compared to in the cover region. How many modes are supported? If you go back to symmetric planar waveguide the number of modes are the integer which is closest to, but greater than v over v over pi by 2, but now all the cut offs are shifted by this much amount. So, the number of modes would now be v minus half tan inverse square root of a divided by pi by 2. So, you obtain this number and find out the integer closest to, but greater than this number.

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What are the cut off wavelengths of various modes and if I change the wavelength how the number of modes would change. So, I can see from here that the cut off wavelength for m th TE mode would be given by $\lambda_c^m = \frac{2\pi d}{V_c^m} \sqrt{n_f^2 - n_s^2}$ which comes directly from the definition of normalized frequency v . So, I can write it down also which we will quite often use.

V is equal to $2\pi d \sqrt{n_f^2 - n_s^2}$. So, from here I get these cut off wavelengths, now if I find out the cut off wavelengths corresponding to various modes then I see that for TE_0 mode the cut off wavelength is 1.7187 for these parameters of waveguide, and for TE_1 mode it is 0.4914 and for TE_2 mode it is 0.2867.

So, if I start from a longer wavelength let us say 2 micro meter, then until I cross this TE_0 mode is not guided. So, from here to here there is no mode guided by the structure, and

as soon as I cross this go below this wavelength, then t 0 mode starts appearing and t 0 mode would be guided now for all the wavelengths is smaller than this.

If I further reduce the wavelength and as soon as I go below 0.49 micrometer TE 1 mode starts appearing and below these this wavelength I will have both TE 0 and TE 1 and so on. So, this is how the waveguide would guide different modes if I change the wavelength. How the thickness affects the guided modes. So, from here I can find out the cut off thickness of mth TE mode from here itself. So, if I find out d in terms of v now and put the cut off of various modes in terms of v, then I can find out the cut off of thickness ok.

So, if I have 0 thickness it means no waveguide no mode. If I start increasing the thickness then up to 0.58 micron there is no mode guided because t 0 mode has finite cut off and as soon as I cross this then TE 0 mode starts appearing, and when I cross 2.03 micron then TE 1 mode starts appearing and so on. So, as I increase the waveguide thickness the number of modes start increasing I should have the label here which is d in micron. So, the label here is t in micron.

Let us work out few examples. So, I take a dielectric step index asymmetric planar waveguide.

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Example

Q. Consider a dielectric step-index asymmetric planar waveguide with $n_f = 1.5$, $n_s = 1.48$, $n_c = 1$ and $d = 4 \mu\text{m}$. Calculate:

- Number of modes at $\lambda_0 = 0.5 \mu\text{m}$.
- The wavelength range in which the waveguide does not support any mode.
- Cut-off wavelength of TE_2 mode.
- Range of d so that only TE_0 and TE_1 modes are guided at $\lambda_0 = 1 \mu\text{m}$.

Solution

(i) $a = \frac{n_s^2 - n_c^2}{n_f^2 - n_s^2} = 19.97$ for $\lambda_0 = 0.5 \mu\text{m}$ $V = \frac{2\pi d}{\lambda_0} \sqrt{n_f^2 - n_s^2} = 6.1357$

$$\frac{V - \frac{1}{2} \tan^{-1}(\sqrt{a})}{\pi/2} = 3.476, \quad \therefore M = 4$$




Defined by n_f is equal to 1.5 and n_s is equal to 1.48 and n_c is equal to 1, and d is equal to 4 micrometer. Now the first thing is to calculate the number of modes at λ_0 is equal to 0.5 micron. So, I first calculate the value of asymmetric parameter a which comes out to be about 20, then I find out what is the value of v at λ_0 is equal to 0.5 micrometer, and this comes out to be about 6.13.

Then I find out this number v minus half \tan^{-1} of square root a , divided by π by 2. So, this comes out to be 3.4. So, the number of modes are 4.

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Example

Q. Consider a dielectric step-index asymmetric planar waveguide with $n_f = 1.5$, $n_s = 1.48$, $n_c = 1$ and $d = 4 \mu\text{m}$. Calculate:

- Number of modes at $\lambda_0 = 0.5 \mu\text{m}$.
- The wavelength range in which the waveguide does not support any mode.
- Cut-off wavelength of TE_2 mode.
- Range of d so that only TE_0 and TE_1 modes are guided at $\lambda_0 = 1 \mu\text{m}$.

Solution

(ii) $M = 0$ if $V < \frac{1}{2} \tan^{-1}(\sqrt{a})$

or $\frac{2\pi d}{\lambda_0} \frac{1}{2} \sqrt{n_f^2 - n_s^2} < \frac{1}{2} \tan^{-1}(\sqrt{a})$

or $\lambda_0 > \frac{2\pi d \sqrt{n_f^2 - n_s^2}}{\tan^{-1}(\sqrt{a})}$ or $\lambda_0 > 4.5427 \mu\text{m}$

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The second is wavelength range in which the waveguide does not support anymore. So, I know there would not be any more support if v is less than half \tan^{-1} of square root a , which if I put the value V here the expression for V here. Then this gives me the condition in terms of λ_0 as λ_0 should be greater than this, and if I plug in all these numbers I find out for λ_0 greater than 4.5427 micrometer the waveguide would not support anymore.

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Example

Q. Consider a dielectric step-index asymmetric planar waveguide with $n_f = 1.5$, $n_s = 1.48$, $n_c = 1$ and $d = 4 \mu\text{m}$. Calculate:

- Number of modes at $\lambda_0 = 0.5 \mu\text{m}$.
- The wavelength range in which the waveguide does not support any mode.
- Cut-off wavelength of TE_2 mode.
- Range of d so that only TE_0 and TE_1 modes are guided at $\lambda_0 = 1 \mu\text{m}$.

Solution

(iii) $V_c^m (TE) = m \frac{\pi}{2} + \frac{1}{2} \tan^{-1}(\sqrt{a})$

\therefore for TE_2 mode $V_c = \pi + \frac{1}{2} \tan^{-1}(\sqrt{a}) = 3.8169$

\Rightarrow for TE_2 mode $\lambda_c = 0.8038 \mu\text{m}$.

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Third is what is the cut off wavelength of TE 2 mode? So, I find out what is the V value for cut off of TE 2 mode. So, which is given by this with m is equal to 2. So, if I put m is equal to 2 then v c is equal to this, and the cut off V c for cut off v for TE 2 mode is this if I translate it to the wavelength it comes out to be 0.8038 micrometer.

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Example

Q. Consider a dielectric step-index asymmetric planar waveguide with $n_f = 1.5$, $n_s = 1.48$, $n_c = 1$ and $d = 4 \mu\text{m}$. Calculate:

- Number of modes at $\lambda_0 = 0.5 \mu\text{m}$.
- The wavelength range in which the waveguide does not support any mode.
- Cut-off wavelength of TE_2 mode.
- Range of d so that only TE_0 and TE_1 modes are guided at $\lambda_0 = 1 \mu\text{m}$.

Solution

(iv) We know that $d_c^m = \frac{V_c^m \lambda_0}{\pi \sqrt{n_f^2 - n_s^2}}$ and $V_c^m = m \frac{\pi}{2} + \frac{1}{2} \tan^{-1}(\sqrt{a})$

$\therefore V_c^0 = 0.6753$ and $V_c^1 = 2.2461$ and $d_c^0 = 0.88 \mu\text{m}$ and $d_c^1 = 2.93 \mu\text{m}$

Hence for TE_0 and TE_1 modes to be guided $0.88 \mu\text{m} < d < 2.93 \mu\text{m}$

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What is the range of d so, that only TE 0 and TE 1 modes are guided at lambda naught is equal to 1 micrometer.

So, I know that for m th mode the cut off thickness is this, and I want TE 0 and TE 1 both the modes guided. So, I find out this d_c for both the modes, and I also know this $V_{c m}$ for m th mode is given by this. So, I find out $V_{c 0}$ for TE 0 mode, $V_{c 1}$ is this for TE 1 mode $V_{c 1}$ is this correspondingly if I now find out d_c for these 2 modes, then for TE 0 mode cut off thickness is 0.88 micrometer, and for this is 2.93 micrometer. So, if TE 0 and TE 1 modes are to be guided then thickness should be between this and this.

So, this is all in this lecture in the next lecture, we will extend the analysis to TM modes.

Thank you.