

Course name: EMI /EMC and Signal Integrity: Principles, Techniques and Applications.

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Week :02

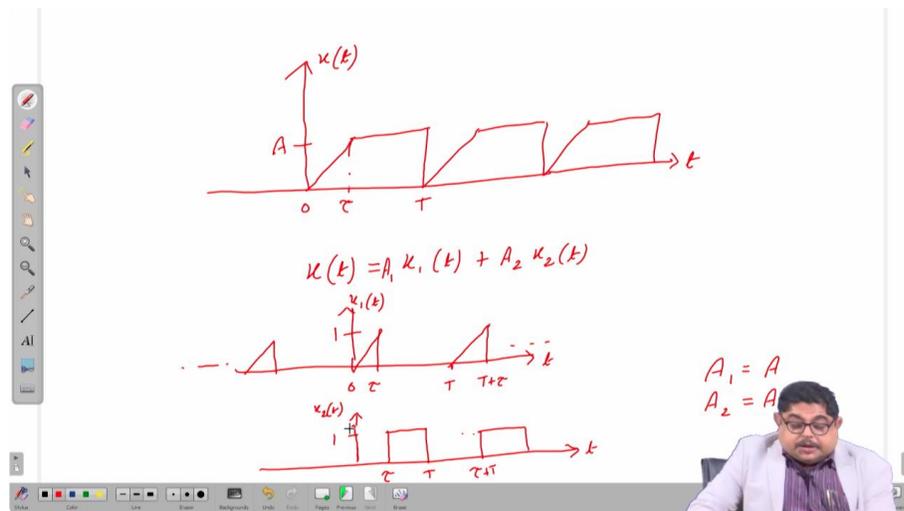
Lecture 7: Important Computational Techniques

Welcome to the 7th lecture of the course on EMI EMC and Signal Integrity Principles Techniques and Applications. today we will discuss some important computational techniques. As I said in the last class that calculating the spectrum of the output signal is quite mathematically intensive. So, can we have some shortcuts to our class of signals or systems. So, I will again go back to your basic knowledge of signals and systems Fourier analysis. So, there we will I will recall the property of linearity of any waveform. So, and it is linearity of your it is Fourier coefficients. So, if a waveform is decomposed into 2 waveforms $x_1(t)$ and $x_2(t)$ with some coefficients that means, if I can write $x(t)$ is equal to $a_1 x_1(t)$ plus $a_2 x_2(t)$ plus $a_3 x_3(t)$ etcetera. Then if $x_1(t)$ has its spectral components as c_{1n} and $x_2(t)$ has its spectral components as c_{2n} and $x_3(t)$ has its spectral components as c_{3n} etcetera. Then $x(t)$ can be written as. So, you see that the spectral component c_n can be now decomposed as. So, we can say that if $x(t)$ is spectral components are $c_n e^{jn\omega_0 t}$, then we can write c_n is $a_1 c_{1n}$ plus $a_2 c_{2n}$ plus $a_3 c_{3n}$ etcetera. This can be easily proved I think you know that also.

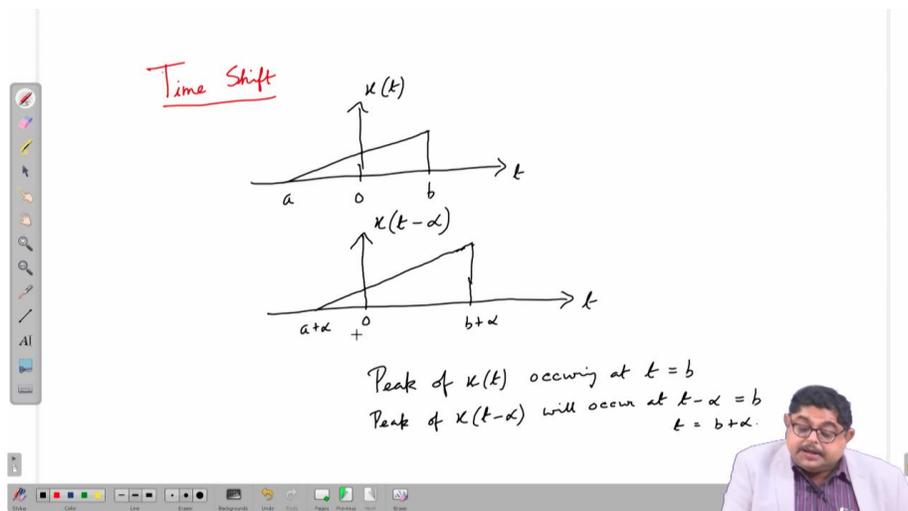
LECTURE 7: IMPORTANT COMPUTATIONAL TECHNIQUES

$$x(t) = \sum_{n=-\infty}^{\infty} C_n e^{jn\omega_0 t}$$
$$x(t) = A_1 x_1(t) + A_2 x_2(t) + A_3 x_3(t) + \dots$$
$$x_1(t) = \sum_{n=-\infty}^{\infty} c_{1n} e^{jn\omega_0 t}$$
$$x_2(t) = \sum_{n=-\infty}^{\infty} c_{2n} e^{jn\omega_0 t}$$
$$x_3(t) = \sum_{n=-\infty}^{\infty} c_{3n} e^{jn\omega_0 t}$$
$$x(t) = \sum_{n=-\infty}^{\infty} (A_1 c_{1n} + A_2 c_{2n} + A_3 c_{3n} + \dots) e^{jn\omega_0 t}$$
$$C_n = A_1 c_{1n} + A_2 c_{2n} + A_3 c_{3n} + \dots$$

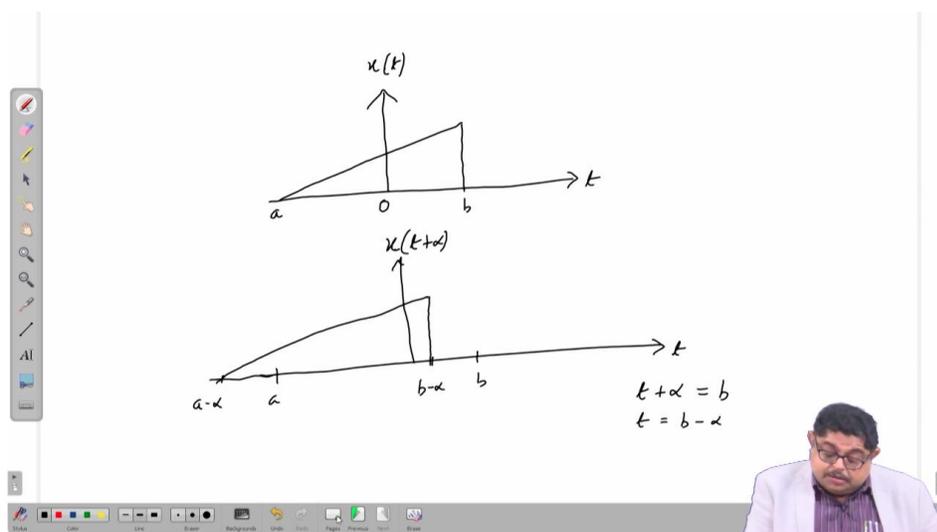
So, you see that when we have this type of signal we have already seen this clock signal. So, here you have noted that this $x(t)$ over one period it was broken into or it can be written as $x_1(t)$ plus $x_2(t)$. So, I can write now that all are $x_1(t)$ is. This is $x_1(t)$ obviously, there were this side also they were there. So, I can write this easily as $x(t)$ equal to $A_1 x_1(t) + A_2 x_2(t)$ where A_1 is A A_2 is also A , but then I will have to take these ones as 1. This one 1 if I take that then A_1 is A A_2 is also A . So, like this. So, this is the first property linearity property that we have seen.



The next one is the time shift property time shift. So, suppose I have a signal $x(t)$ which is when I give it a time shift. So, let me call it I am giving a time shift of α actually I generally we call it τ , but since in pulse strain we have τ reserved for on time of the pulse. So, I am calling it α . So, it will be b plus α , it will be a plus α this is 0. Now, how you find this? You see that the corresponding points on the waveforms of $x(t)$ and $x(t - \alpha)$ occur when the arguments of $x(t)$ and $x(t - \alpha)$ are identical. The meaning is what is the corresponding point? Suppose this is the peak, this is also the peak. Now, where it will occur? When the arguments of $x(t)$ and arguments of $x(t - \alpha)$ are identical. So, peak of $x(t)$ is occurring at t is equal to b . So, peak of $x(t - \alpha)$ will occur at $t - \alpha$ is equal to b or t is equal to $b + \alpha$. You can think corresponding point as the nulls. So, the null or 0 of $x(t)$ is occurring at a . So, that means, at t is equal to a then 0 is occurring. So, for this function the time delayed function the delay null will occur at $t - \alpha$ is equal to a that is $a + \alpha$.



Similarly, this if I have again I am drawing $x(t)$ versus t a b this is 0 $x(t)$ plus α . So, now, you will be able to say that where will be the peak. So, for $x(t)$ the peak is occurring at t is equal to b . So, for t plus α the peak will occur at t plus α is equal to b that means, t will be b minus α that means, now if this is a this is b it will be b minus α . So, something may be here b minus α . So, and so, you can say that peak will be here. So, and this is a minus α provided α is positive if α is negative then it will be other side ok. Now, so, this is an example of time shifted signal.



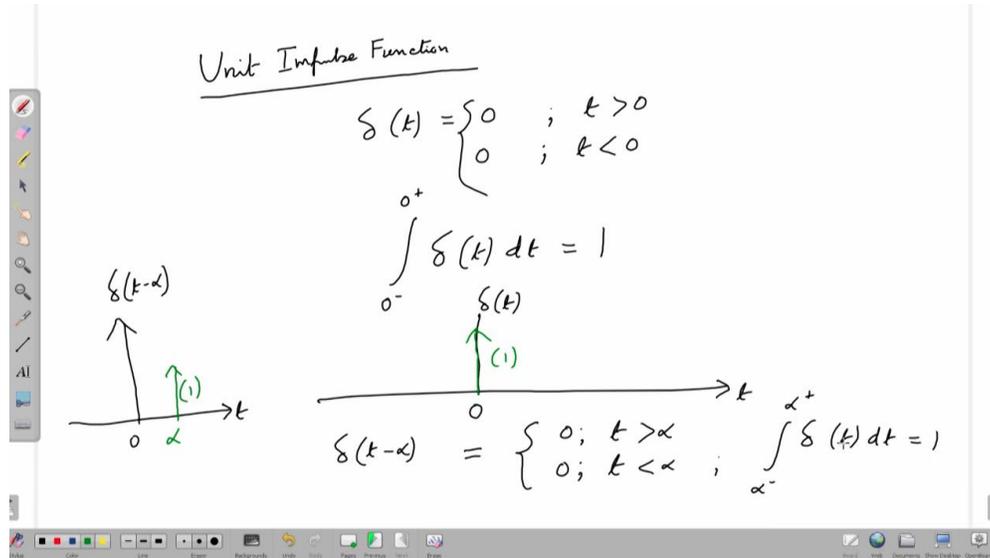
So, from Fourier series we know that if $x(t)$ is its spectral coefficients are C_n then what will be spectral coefficients of $x(t - \alpha)$ that will be $C_n e^{-jn\omega\alpha}$. So, n is equal to minus α to α . So, what is happening $C_n e^{-jn\omega\alpha}$ into $e^{jn\omega\alpha}$ this thing is now our C_n dashed if I call it for this thing. So, we can say that C_n dashed is

nothing, but $C_n e^{j n \omega_0 t}$ to the power minus $j n \omega_0 \alpha$. Similarly, for $x(t + \alpha)$ if I call that as C_n'' then it will be $C_n e^{j n \omega_0 t}$ to the power plus $j n \omega_0 \alpha$. So, this is the time shift property this also you know easily from Fourier series analysis you can prove this.

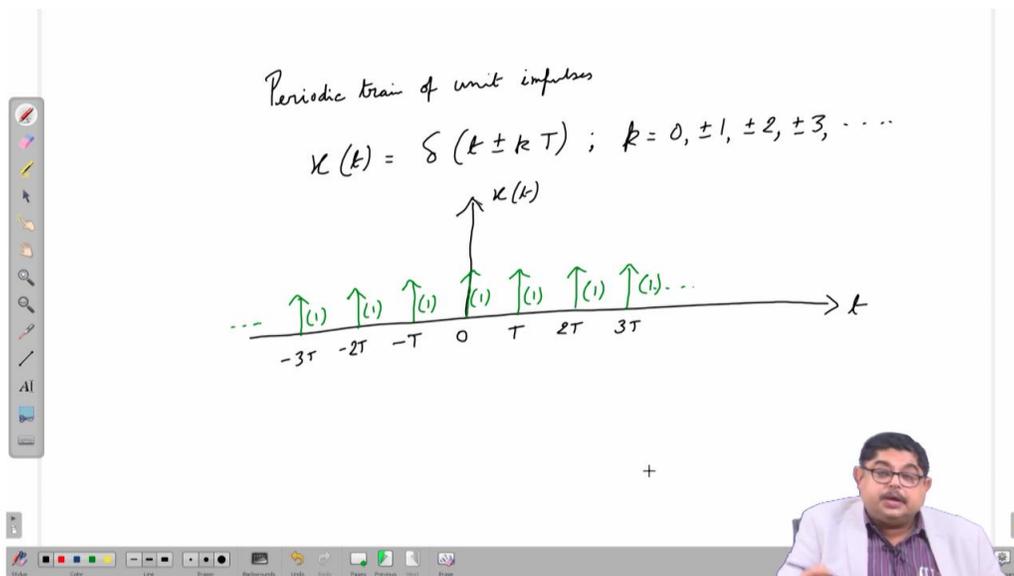
$$\begin{aligned}
 x(t) &= \sum_{n=-\infty}^{\infty} C_n e^{j n \omega_0 t} \\
 x(t-\alpha) &= \sum_{n=-\infty}^{\infty} C_n e^{j n \omega_0 (t-\alpha)} \\
 &= \sum_{n=-\infty}^{\infty} C_n e^{-j n \omega_0 \alpha} e^{j n \omega_0 t} \\
 &= \sum_{n=-\infty}^{\infty} C_n' e^{j n \omega_0 t} \\
 C_n' &= C_n e^{-j n \omega_0 \alpha} \\
 C_n'' &= C_n e^{+j n \omega_0 \alpha}
 \end{aligned}$$

The next one I think you know or every electronics engineer should know what is unit impulse function. Its definition is $\delta(t)$ is equal to ∞ at $t = 0$ and 0 for $t > 0$ and $t < 0$ and you see that it exist only at $t = 0$. So, it is not a function actually in the proper sense what we learnt in our basic courses on calculus that a function means it is continuous. So, it should have its left hand limit is equal to right hand limit is equal to the function at a point. So, at $t = 0$ you see its left hand limit is 0 its right hand limit is 0 . So, limit exist, but none of the limit is equal to the function at $t = 0$. So, it is not continuous, but this is in physics physical reality it comes that is why it was defined as a special function and the this definition is not complete you will have to define that its $\int_{-\infty}^{\infty} \delta(t) dt = 1$ that means, integration of $\delta(t) dt$ is equal to finite is equal to finite since we are having unit impulse it is this its representation is like this that this exist here and we denote its strength here. Now, if it was not unit impulse it could have been some constant that we should have written. So, this is your $\delta(t)$ this is your t this is and this is your $\delta(t - \alpha)$ ok. If I have $\delta(t - \alpha)$ then that definition will be again let us say unit impulse. So, ∞ for $t = \alpha$ 0 for $t > \alpha$ and 0 for $t < \alpha$ and also $\int_{-\infty}^{\infty} \delta(t - \alpha) dt = 1$. So, all that that means, this makes the definition perfect and its representation will be that time. So, this is the delta function. It is represented by a vertical arrow pointing upward if it is positive if it is vertical arrow going downward it is its value is negative the height of the arrow is immaterial the

strength of the impulse is denoted in parenthesis adjacent to this arrow. So, instead of unit impulse if you have a impulse of strength a you will write it as a here.



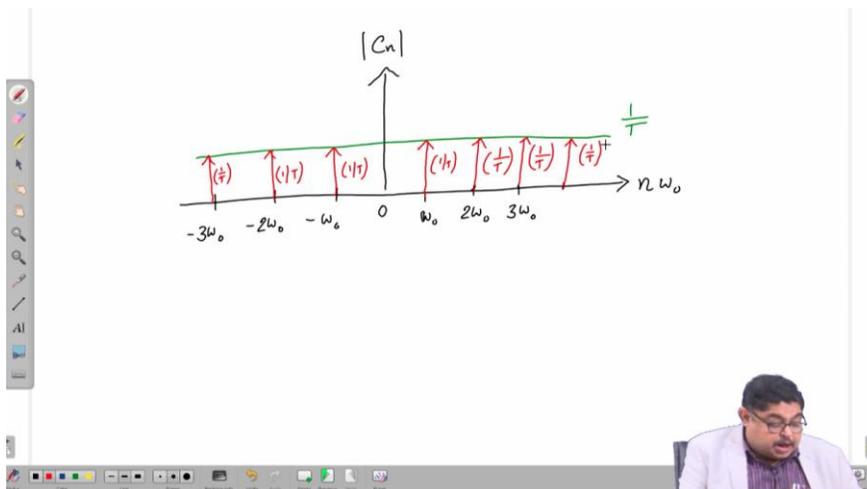
Now, consider a periodic train of periodic train of unit impulses. So, that means, I can write $x(t) = \sum_{k=-\infty}^{\infty} \delta(t - kT)$ is equal to $\delta(t) + \delta(t - T) + \delta(t - 2T) + \delta(t - 3T) + \dots$. So, then what is it delta sorry $x(t)$. So, the waveform will be look like. So, $\delta(t)$ is equal to $\delta(t) + \delta(t - T) + \delta(t - 2T) + \delta(t - 3T) + \dots$. So, this is the periodic pulse train. So, again the complex exponential Fourier it periodic pulse train means it has Fourier series. So, we can determine its complex exponential Fourier coefficients.



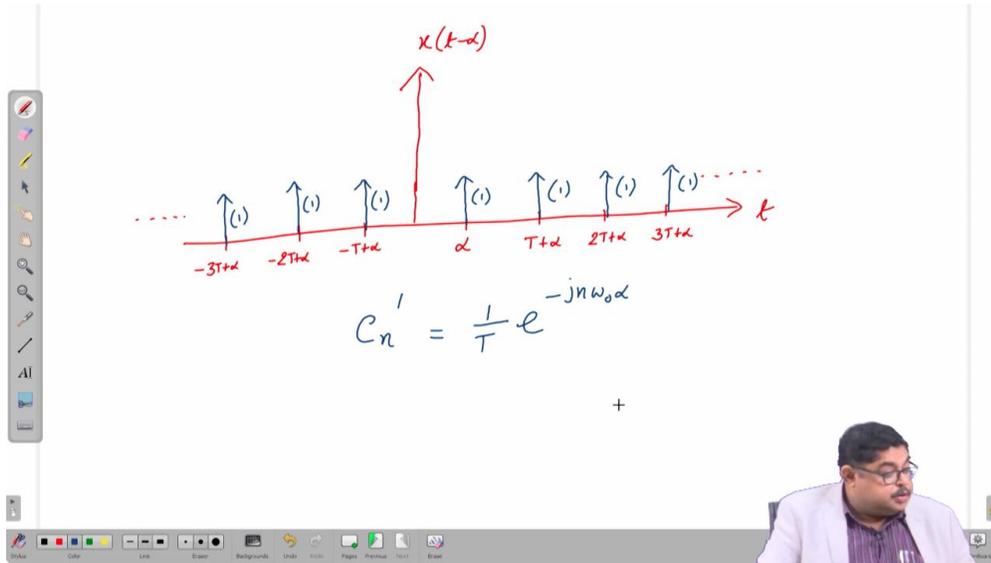
So, that will be C_n is equal to $\frac{1}{T}$ by $t=0$ to t that is its period Δt to the power minus $j n \omega_0 t$ $\int_0^T \delta(t) e^{-j n \omega_0 t} dt$ is equal to $\frac{1}{T}$. So, everywhere $\delta(t)$ is 0 except at $t=0$ that is why I see what happens at 0^- to 0^+ to the power minus $j n \omega_0 t$ $\int_{0^-}^{0^+} \delta(t) e^{-j n \omega_0 t} dt$. So, it exists only between 0^- and 0^+ , but there what is the value of $e^{-j n \omega_0 t}$ to the power minus $j n \omega_0 t$ at $t=0$ is equal to 1. So, we are left with $\int_{0^-}^{0^+} \delta(t) dt$ and that value we know is 1 by T into 1 is equal to that is the definition. So, it is $\frac{1}{T}$.

$$\begin{aligned}
 C_n &= \frac{1}{T} \int_0^T \delta(t) e^{-jn\omega_0 t} dt \\
 &= \frac{1}{T} \int_{0^-}^{0^+} \delta(t) e^{-jn\omega_0 t} dt \\
 &= \frac{1}{T} \int_{0^-}^{0^+} \delta(t) dt = \frac{1}{T} \times 1 = \frac{1}{T}
 \end{aligned}$$

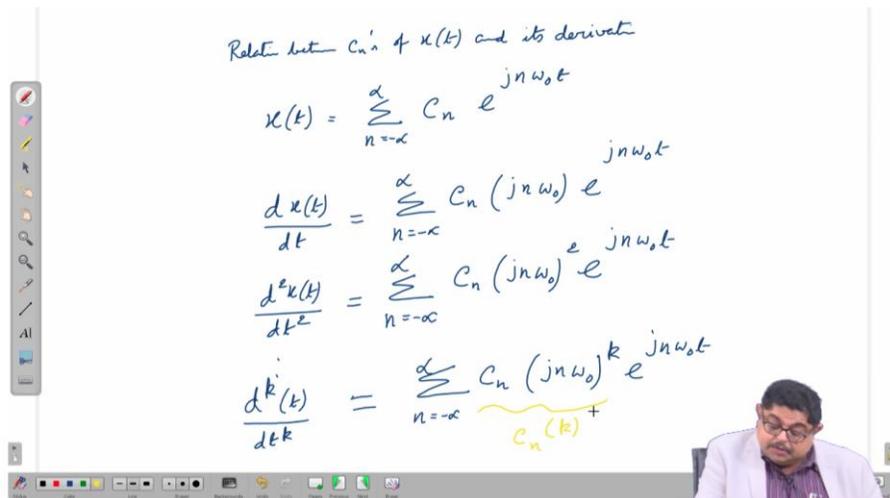
So, it is Fourier coefficient that means, if I plot the C_n magnitudes versus n . So, n is 1 or you can call $n \omega_0$. So, ω_0 , $2\omega_0$, $3\omega_0$, etcetera minus ω_0 , minus $2\omega_0$, minus $3\omega_0$, etcetera the value will be same for all and that value is $\frac{1}{T}$. So, you see spectral coefficients are all same actually it should be discrete spectrum actually I have this is the envelope of that. So, the discrete spectrum will be something like this $\frac{1}{T}$, $\frac{1}{T}$, $\frac{1}{T}$, $\frac{1}{T}$, $\frac{1}{T}$, $\frac{1}{T}$, $\frac{1}{T}$. Now, let us say the this pulse train is shifted ahead in time by α .



So, this pulse train so $x(t - \alpha)$ by t and so the pulse train sorry. So, so this is the shifted ahead in time by α a periodic unit impulse pulse train. So, using time shift property we can say that the C_n shifted version of these is 1 by $t e$ to the power 1 by t was original pulse trains C_n . So, it will be minus $J n \omega_0 \alpha$.



And finally, I have one more property to recall from Fourier analysis that relation between C_n 's of $x(t)$ and its derivatives. Let $x(t)$ is equal to n is equal to minus infinity to infinity $C_n e$ to the power $J n \omega_0 t$ and so what will be $\frac{d}{dt} x(t)$? It will be n is equal to minus infinity to infinity $C_n J n \omega_0 e$ to the power $J n \omega_0 t$. What is $\frac{d^2}{dt^2} x(t)$? It will be n is equal to minus $C_n J n \omega_0^2 e$ to the power $J n \omega_0 t$ up to that what is $\frac{d^k}{dt^k} x(t)$ sorry. So, let us call that this one as $C_n^{(k)}$ then can I say that sorry $C_n^{(k)}$ is nothing, but $C_n J n \omega_0^k e$.



So, we can say that so this is the property that the expansion coefficient of $x(t)$ and expansion coefficient of k th derivative of $x(t)$ they are related by these. So, C_n is 1 by $J^n n \omega_0$ to the power k $C_n^{(k)}$. So, with this we have seen the preliminary properties needed to go to our shortcut. As I said that all this we are doing that we will find an algorithm to derive for our clock pulses periodic pulse trains for them how to find the Fourier coefficients because we will have to estimate the harmonics as an EMC engineer we should know that. So, how to calculate it very fast that will be the topic of our next lecture. Thank you.

The image shows a digital whiteboard interface. In the center, two equations are written in black ink:

$$C_n^{(k)} = C_n (jn\omega_0)^k$$
$$C_n = \frac{1}{(jn\omega_0)^k} C_n^{(k)}$$

Below the equations, there is a small plus sign (+). In the bottom right corner, there is a small video inset showing a man with glasses and a light-colored jacket. The whiteboard has a toolbar on the left side with various drawing tools like a pencil, eraser, and highlighter. At the bottom, there is a Windows taskbar with several application icons.