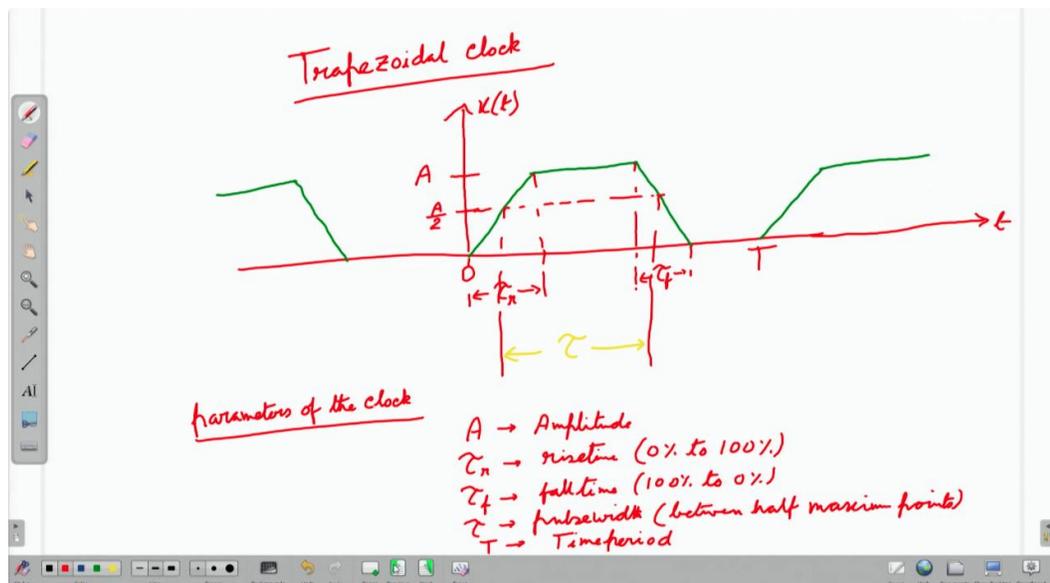


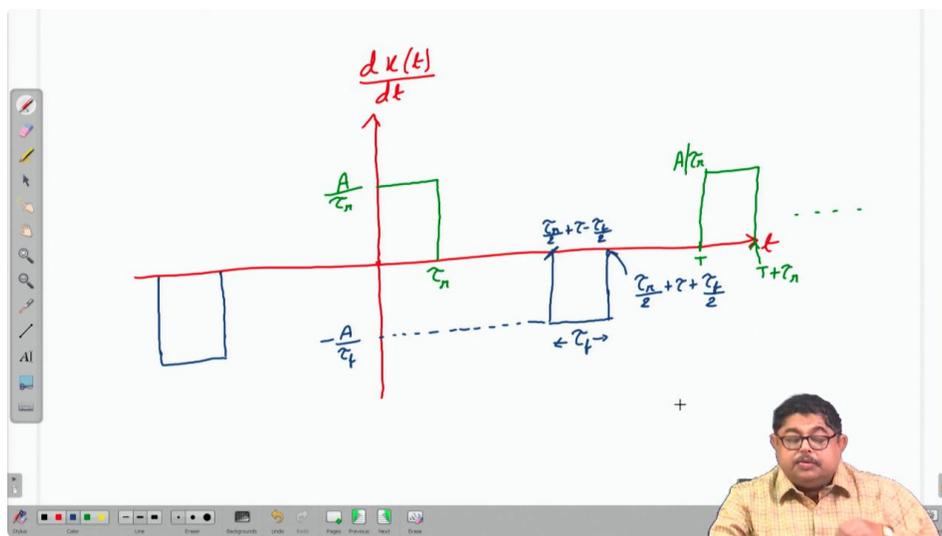
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 Week :02
 Lecture 10: Trapezoidal Clock

Welcome to this 10th lecture of the course on EMIMC and Signal Integrity Principles Techniques and Applications. In the previous lecture, we have seen the algorithm for getting the Fourier coefficient in very elegant way. Now, today we will apply all that to the trapezoidal clock. Now, with this knowledge we will try to find the what is the Fourier coefficient for a clock which has finite rise time and fall time that is a trapezoidal clock. So, let us see a trapezoidal clock, trapezoidal clock or trapezoidal pulse tran . So, let me first draw there are some parameters of the clock that we need to understand.

So, first let me draw the clock. So, let us draw the clock. For our analysis we will take that the time the clock takes to go from 0 to the peak value A that is τ_r . You know, but in industry they sometimes take 10 percent to 90 percent of the waveform the time it takes to go from 10 percent of the peak to 90 percent of the peak sometimes they take as 20 percent to 80 percent of the peak different industry uses different things, but here we will take this definition that τ_r is this similarly our fall time is this and for this one what is our pulse width that will take that it is between two half amplitudes amplitude is A , A by 2 is half amplitude. So, if I take this this is our τ and this is our time period when it is. So, what are the parameters of the clock? Parameters of the clock one is A is the amplitude, τ_r is rise time 0 percent to 100 percent, τ_f is fall time 100 percent to 0 percent, τ pulse width between half maximum points and T is the time period. So, with this we will try to find the spectrum that is the c_n values. So, let us first differentiate it once.

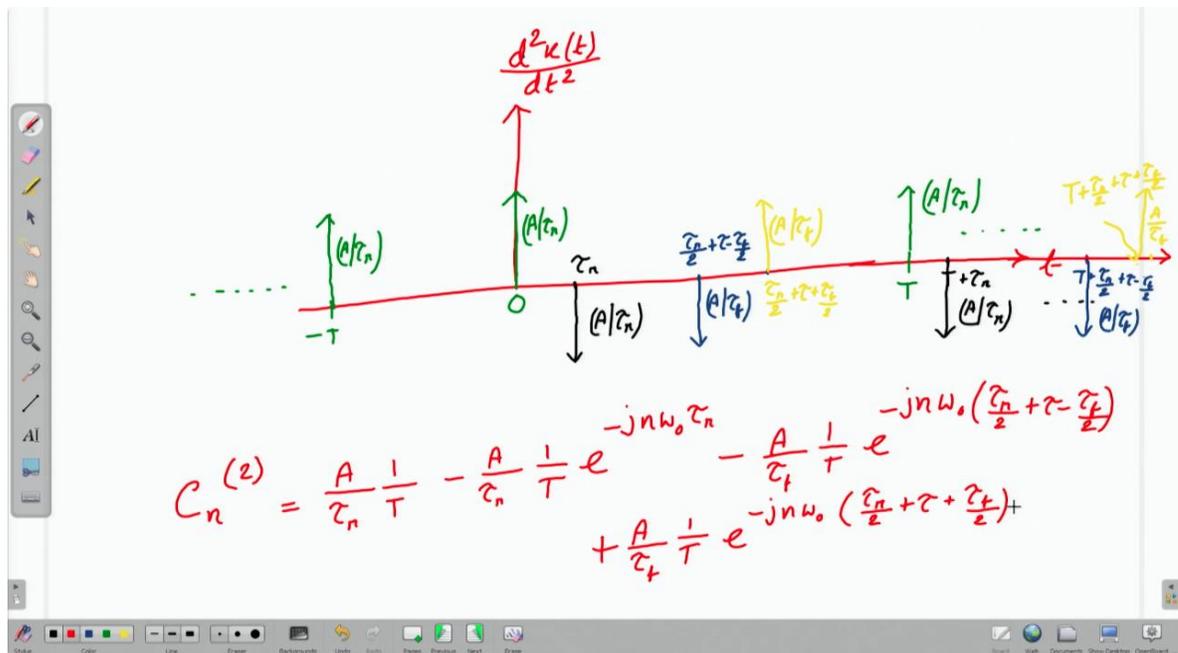


So, what will happen after first differentiation $\frac{dx(t)}{dt}$. So, this portion will give me a rectangular portion again you know that amplitude will be A by τ_r of that rectangular portion. So, let me use green colour. So, it is A by τ_r and it will go up to τ_r . Then there is a after differentiation the portion from this τ_r to this line the flat line that will be 0 then again at τ_f there will be something. So, let us say this is the falling portion. Now, what is the value of the falling portion that we will have to find? That means what is the value when the from the maximum it starts falling. If you see that if I look at that it is the half maximum of the rising portion that is I can say occurring at τ_r by 2. So, from τ_r by 2 if I give τ it is coming at the second half maximum point and from that if I subtract τ_f by 2 and coming to the point where the clock is starting falling. So, this value of this will be let me. So, at time τ_r by 2 plus τ minus τ_f by 2 I will see that there is again a τ_r falling the means falling. So, after differentiation it becomes this. So, this width is τ_f or you can find out. So, this point is τ_r 2 by a this point will be τ_r by 2 plus τ plus τ_f by 2 that means this width is τ_f . Then at T there will be this repetition that it will be a by τ_r this point will be T plus τ_r that is it. So, this will go. Similarly, here again this value is what that means this value will be minus a by τ_f this side. So, this is rising edge let us see what will happen here this is again a falling portion. So, I can say that there will be. So, like this it will go.



So, still we have not got any impulse train. So, we will have to make another differentiation. So, we will have this one. If I differentiate you see this this one should be green. So, this one will give me an impulse this one will give me a rising impulse. Now, this one that means at τ_r what it that will give me a falling impulse this one will give me a falling impulse. Similarly, here you see this will give me a falling impulse this will give me a rising impulse. So, there will be 4 types of things. So, let us do that carefully

that $\frac{dx}{dt}$. So, let me take the first rising impulse as this one. So, this is a by τ_r it will be repeated at T there will be another a by τ_r . Similarly, this is at 0 this is at $-\tau_r$ there will be a by τ_r . Then there will be at τ_r there will be a falling one. So, let me choose this color that at τ_r there will be a falling one with value a by τ_r . Similarly, it will be at $T + \tau_r$ $T + \tau_r$ there will be again its next occurrence. So, like this it will also go this will go like this this will go like this. Now, let us see the other ones. So, at τ_r plus that you see previous one τ_r by 2 plus τ_r minus τ_r a by 2 there will be a falling impulse train. So, τ_r by 2 plus τ_r minus τ_r a by 2 there will be a falling impulse strain again that value will be a by τ_r and since I do not have much space I am writing it here. So, what color is still remain in this one. So, can write at τ_r by 2 plus τ_r plus τ_r a by 2 there will be a rising impulse strain a by τ_r this whole things will be again repeated. So, we can say that here if I look at this look at at $T + \tau_r$ by 2 plus τ_r minus τ_r a by 2 there will be a falling impulse strain a by τ_r at $T + \tau_r$. $T + \tau_r$ by 2 plus τ_r plus τ_r a by 2 like this. So, there are 4 different pulse trains. So, all of them were obtained after second differentiation. So, can I write $C_n^{(2)}$ is equal to a by τ_r the green one 1 by T minus a by τ_r 1 by T it is shifted by τ_r . So, e to the power minus j in ω naught τ_r minus a by τ_r 1 by T e to the power minus j in this has been shifted by j in ω naught τ_r by 2 plus τ_r minus τ_r a by 2 plus a by τ_r 1 by T e to the power minus j in ω naught τ_r by 2 plus τ_r plus τ_r a by 2 this is my $C_n^{(2)}$.



So, I can simplify this I take a by τ_r out then 1 by τ_r e to the power minus j in ω naught τ_r by 2 e to the power j in ω naught τ_r by 2 minus e to the power minus j in ω naught τ_r by 2 minus 1 by τ_r e to the power minus j in ω naught τ_r

$\frac{1}{\tau_n} e^{-jn\omega_0 \frac{\tau_n}{2}} \left(e^{jn\omega_0 \frac{\tau_n}{2}} - e^{-jn\omega_0 \frac{\tau_n}{2}} \right) - \frac{1}{\tau_f} e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \left(e^{jn\omega_0 \frac{\tau_f}{2}} - e^{-jn\omega_0 \frac{\tau_f}{2}} \right)$

$= \frac{A}{T} \left[\frac{2j}{\tau_n} e^{-jn\omega_0 \frac{\tau_n}{2}} \sin(n\omega_0 \frac{\tau_n}{2}) - \frac{2j}{\tau_f} e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \sin(n\omega_0 \frac{\tau_f}{2}) \right]$

$= \frac{jA}{T} \left[(n\omega_0) e^{-jn\omega_0 \frac{\tau_n}{2}} \text{Sinc}(n\omega_0 \frac{\tau_n}{2}) - (n\omega_0) e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \text{Sinc}(n\omega_0 \frac{\tau_f}{2}) \right]$

$$C_n = \frac{A}{T} \left[\frac{1}{\tau_n} e^{-jn\omega_0 \frac{\tau_n}{2}} \left(e^{jn\omega_0 \frac{\tau_n}{2}} - e^{-jn\omega_0 \frac{\tau_n}{2}} \right) - \frac{1}{\tau_f} e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \left(e^{jn\omega_0 \frac{\tau_f}{2}} - e^{-jn\omega_0 \frac{\tau_f}{2}} \right) \right]$$

$$= \frac{A}{T} \left[\frac{2j}{\tau_n} e^{-jn\omega_0 \frac{\tau_n}{2}} \sin(n\omega_0 \frac{\tau_n}{2}) - \frac{2j}{\tau_f} e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \sin(n\omega_0 \frac{\tau_f}{2}) \right]$$

$$= \frac{jA}{T} \left[(n\omega_0) e^{-jn\omega_0 \frac{\tau_n}{2}} \text{Sinc}(n\omega_0 \frac{\tau_n}{2}) - (n\omega_0) e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)} \text{Sinc}(n\omega_0 \frac{\tau_f}{2}) \right]$$

So, $e^{-jn\omega_0 \frac{\tau_n}{2}}$ and $e^{-jn\omega_0 \left(\tau + \frac{\tau_n}{2} \right)}$ are the power minus j in $n\omega_0$ naught by τ and $\tau + \tau_n$ by 2. So, $\text{sinc}(n\omega_0 \frac{\tau_n}{2})$ and $\text{sinc}(n\omega_0 \frac{\tau_f}{2})$ are the power minus j in $n\omega_0$ naught τ by 2. So, this is my C_n check previous this is also C_n this is C_n C_n is still going on. So, what is C_n ? C_n will be 1 by double differentiation $j n \omega_0$ naught whole square C_n naught is equal to 0 because for DC this is not the case. So, is equal to if you put that then minus j a by $n \omega_0$ naught τ by 2 into the power minus j in $n \omega_0$ naught τ by 2 all those things that $\text{sinc}(n \omega_0 \text{ naught } \tau \text{ by } 2)$ minus $e^{-jn\omega_0 \text{ naught } \tau \text{ by } 2}$ into the power minus j in $n \omega_0$ naught what is happening ω_0 naught τ by 2. So, this is the result for trapezoidal clock. Now, this thing can be or in a in most of the cases the in trapezoidal clock the rise time and fall time they are same. So, if we make that that τ_r is equal to τ_f then this can be further simplified.

$$C_n^{(2)} = \frac{jAn\omega_0}{T} e^{-jn\omega_0 \frac{\tau_n}{2}} \left[\text{Sinc}\left(n\omega_0 \frac{\tau_n}{2}\right) - e^{-jn\omega_0 \tau} \text{Sinc}\left(n\omega_0 \frac{\tau}{2}\right) \right]$$

$$C_n = \frac{1}{(jn\omega_0)^2} C_n^{(2)} \quad ; \quad n \neq 0$$

$$= \frac{-jA}{n\omega_0 T} e^{-jn\omega_0 \frac{\tau_n}{2}} \left[\text{Sinc}\left(n\omega_0 \frac{\tau_n}{2}\right) - e^{-jn\omega_0 \tau} \text{Sinc}\left(n\omega_0 \frac{\tau}{2}\right) \right]$$

$$\tau_r = \tau_f$$

So, that we will see now that for τ_r is equal to τ_f that means, rise time and fall time are equal which is most of the cases that happens. So, C_n becomes minus $j a$ by $2 \pi n$ e to the power minus $j n \omega_0 \tau_r$ by 2 then you can now take this sinc is out. So, sinc $n \omega_0 \tau_r$ by 2 if you take it out what you are left with is $1 - e$ to the power minus $j n \omega_0 \tau$ you see from previous this e to the power e to the power minus $j n \omega_0 \tau$ that will be there. So, again you can take e to the power minus $j n \omega_0 \tau$ by 2 common that will give you another sinc term. So, minus if you do that minus $j a$ by $2 \pi n$ e to the power minus $j n \omega_0 \tau_r$ by 2 sinc of $n \omega_0 \tau_r$ by 2 e to the power minus $j n \omega_0 \tau$ by 2 then you will be left with e to the power $j n \omega_0 \tau$ by 2 minus e to the power minus $j n \omega_0 \tau$ by 2. So, you can do this manipulation and you will get finally, this value this you can easily do I have done enough. So, sinc $n \omega_0 \tau_r$ by 2 then sinc $n \omega_0 \tau$ by 2 then e to the power minus $j n \omega_0 \tau$ plus τ_n by 2. So, you see this one and this one these two exponential terms they are here and there are two sinc terms one is dependent on the pulse width τ another is dependent on the rise time or fall time because now rise time or fall time are similar. So, this is the expression there are two sinc terms product in a normal pulse train that means, where rise time fall time is not there you have one sinc function here you have product of two sinc function one is like the pulse train another is due to the rise time. Now, this is the double sided spectrum. So, what is the single sided spectrum that is C_n plus that will be equal to $2 C_n$ is equal to $2 a \tau$ by T if we take the magnitude part only then this e to the power term will go and so we will have $\sin n \omega_0 \tau$ by 2 by $n \omega_0 \tau$ by 2. Then $\sin n \omega_0 \tau$ by 2 $n \omega_0 \tau$ by 2 this is the magnitude spectrum of a trapezoidal clock. Now, here again as before we can remove ω_0 by noting that ω_0 is 2π by t that will bring our t because t is in instead of in terms of frequency this we can write.

$$\begin{aligned}
 \tau_n &= \tau_f \\
 C_n &= -j \frac{A}{2\pi n} e^{-jn\omega_0 \frac{\tau_n}{2}} \left\{ \text{Sinc} \left(n\omega_0 \frac{\tau_n}{2} \right) \right\} \left[1 - e^{-jn\omega_0 \tau} \right] \\
 &= -\frac{A}{2\pi n} e^{-jn\omega_0 \frac{\tau_n}{2}} \text{Sinc} \left(n\omega_0 \frac{\tau_n}{2} \right) e^{-jn\omega_0 \frac{\tau}{2}} \left[e^{jn\omega_0 \frac{\tau}{2}} - e^{-jn\omega_0 \frac{\tau}{2}} \right] \\
 &= A \frac{\tau}{T} \text{Sinc} \left\{ n\omega_0 \frac{\tau}{2} \right\} \cdot \text{Sinc} \left\{ n\omega_0 \frac{\tau_n}{2} \right\} \\
 &\quad e^{-jn\omega_0 \left(\frac{\tau + \tau_n}{2} \right)} \\
 |C_n^+| &= 2|C_n| = 2A \frac{\tau}{T} \left| \frac{\text{Sin} \left(n\omega_0 \frac{\tau}{2} \right)}{\left(n\omega_0 \frac{\tau}{2} \right)} \right| \left| \frac{\text{Sin} \left(n\omega_0 \frac{\tau_n}{2} \right)}{\left(n\omega_0 \frac{\tau_n}{2} \right)} \right|
 \end{aligned}$$

So, if we do that then our C_n one sided spectrum magnitude that will be $2A \frac{\tau}{T} \text{sin} \left(n\pi \frac{\tau}{T} \right) \text{sin} \left(n\pi \frac{\tau_n}{T} \right)$. What will be our angle or phase spectrum? We have already seen that how to for sin what happens? So, that will be plus minus $n\pi \frac{\tau + \tau_n}{T}$ from this that is phase spectrum will be addition of these two. So, that will be sin space and this sin space you know that sin may be plus minus that way is the spectrum and what is the DC value C_0 ? C_0 is a τ by T . So $X(t)$ can be written as C_0 plus one sided spectrum n is equal to 1 to infinity $C_n^+ \cos(n\omega_0 t + \angle C_n)$ Now, you see that one of the popular choice is or most of the cases people try to make the duty cycle 50 percent. So, if I do that that mean then τ by T will be half. So, the first sinc term that will be the first sinc term is $\text{sin} \left(n\pi \frac{\tau}{T} \right)$ by $n\pi \frac{\tau}{T}$. So, this will go to 0 for n even that means, the even harmonics are absent if the clock has 50 percent duty cycle you only have odd harmonics of the clock. However in practice clock duty cycle fluctuates it is the designer makes it 50 percent, but today it may be something 49 percent tomorrow it may fluctuate it is very rare that it will tend at same 50 percent. Now, the problem is near 0 the sinc term varies widely you know a $\text{sin} \theta$ that you see that near 0 that has a very fast variation whereas, near peak it is a quite stable. So, slight variation in duty cycle from 50 percent to some other value can cause even harmonics to fluctuate rapidly whereas, odd harmonics are relatively stable this have is this variation. So, in a radiated emission test if the offending frequency is an even harmonics of the clock then emission level measured at one day may fluctuate widely on another day due to variation on clock duty cycle. So, you should always check whether even harmonic is the culprit or not. So, remember these in a field test you should remember that even harmonics try to find out

and if from your one day it is often not meeting the limit in another day it is meeting the limit sometimes the manufacturer may report you that that he has done it, but at your end you are finding that it is not there check the even harmonic. If the even harmonic is there try to find that if you can bring the duty cycle proper then that may go away. So, we have now got the spectrum of the trapezoidal clock a practical clock waveform. Now, we will have to estimate that what are the bounds of this clock that means, if I tell you that at 5 harmonic what may be the highest value of this spectrum. Now, in field you cannot always calculate everything. So, we will try to get some upper bound of this harmonic spectrum the upper bound of the spectrum. So, that we can find out that if our those harmonics are meeting that bound then we are safe that they cannot cross the limit emission limits. So, that will take up next day.

$$|C_n^+| = 2A \frac{\tau}{T} \left| \frac{\sin(n\pi \frac{\tau}{T})}{(n\pi \frac{\tau}{T})} \right| \left| \frac{\sin(n\pi \frac{\tau_n}{T})}{(n\pi \frac{\tau_n}{T})} \right|$$

$$\angle C_n = \pm n\pi \frac{\tau + \tau_n}{T}$$

$$C_0 = A \frac{\tau}{T}$$

$$x(t) = C_0 + \sum_{n=1}^{\infty} |C_n^+| \cos(n\omega_0 t + \angle C_n)$$

$$\frac{\tau}{T} = \frac{1}{2} \quad \frac{\sin(\frac{n\pi}{2})}{\frac{n\pi}{2}} = 0 \text{ for } n \text{ even.}$$