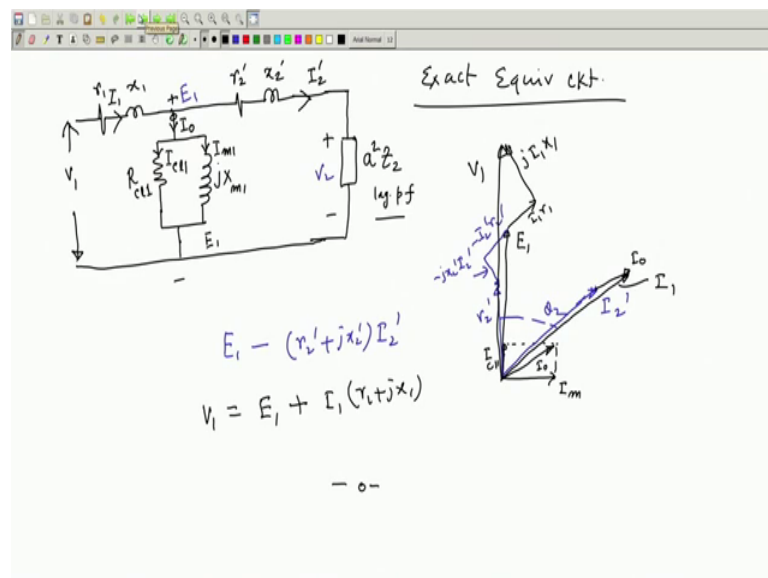


Electrical Machines - I
Prof. Tapas Kumar Bhattacharya
Department of Electrical Engineering
Indian Institute of Technology, Kharagpur

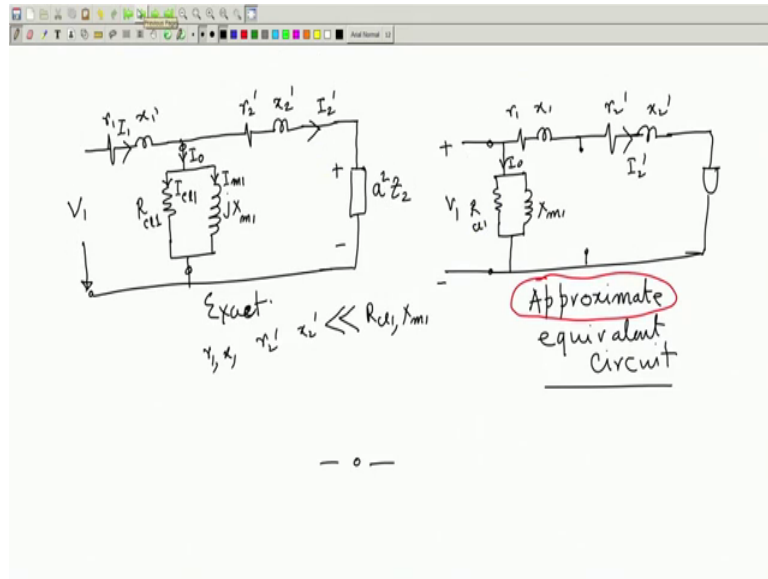
Lecture - 16
Determination of Equivalent Circuit Parameters - No Load Test

Welcome to lecture 16 and we have been discussing about the exact equivalent circuit, its phasor diagram, then approximate equivalent circuit and I will also draw its phasor diagram.

(Refer Slide Time: 00:33)

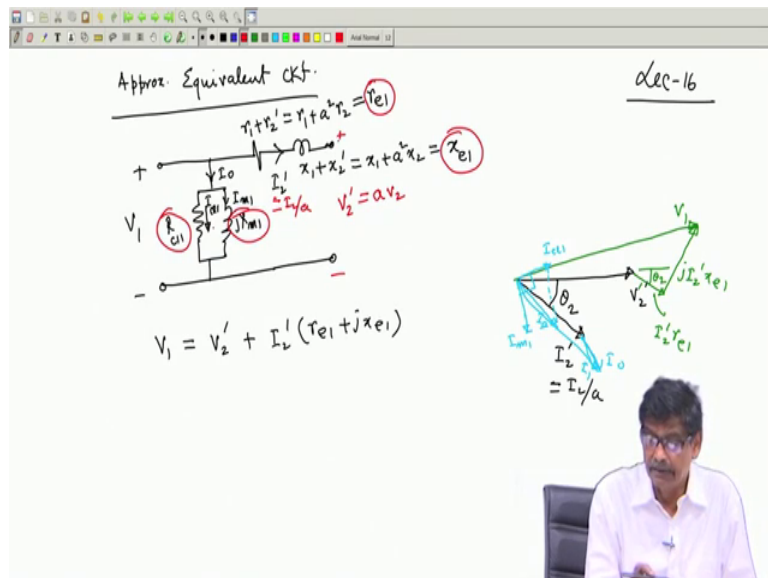


(Refer Slide Time: 00:38)



Recall that the approximate equivalent circuit was like this where this parallel branch has been brought right across the supply. The approximation made is under no load condition as if there is no current through the winding, but nonetheless core loss will always take place. So, it was like that now if you see this r_1 x_1 and r_2 dashed x_2 dashed, they become in series ok. So, this is one of the greatest advantage of looking at the transformer equivalent circuit in terms of its approximate representation.

(Refer Slide Time: 01:31)



So, what will happen is this equivalent circuit, then can be further simplified. So, approximate equivalent circuit refer to primary it will then boil down to this one. These r_1 and r_2 dashed can be represented by a single resistance as $r_1 + r_2$ dashed which is nothing, but r_1 plus a square into actual resistance of the secondary term and this can be written simply as r_e . That is equivalent resistance in terms of primary winding; we see equivalent resistance.

Similarly, the two reactants can be grouped together and written as x_1 plus x_2 dashed and that is equal to x_1 plus a square x_2 and this is called equivalent reactance; equivalent reactants. Equivalent reactants referred to primary side x_e and that is all. These two are the secondary terminals where an impedance will be connected if it is connected a square z_2 .

And this will be R_{c1} and this is jX_{m1} and this is your V_1 and mind you this current is I_2 dashed and this is no load current and no load current comprises of two parts; core loss component c_1 and I_{m1} . Then this voltage if I show it by different colours. So, that you understand, this voltage is V_2 dashed which is equal to a into V_2 actual secondary. This current I_2 dashed is equal to I_2 by a. This is how things will go, then the phasor diagram drawing becomes so simple.

Let us see; suppose we start drawing the phasor diagram with respect to V_2 dashed. Now as you can see V_1 will be equal to V_2 dashed; this voltage plus this drop plus I_2 dashed into r_e plus jx_e . This will be the phasor diagram.

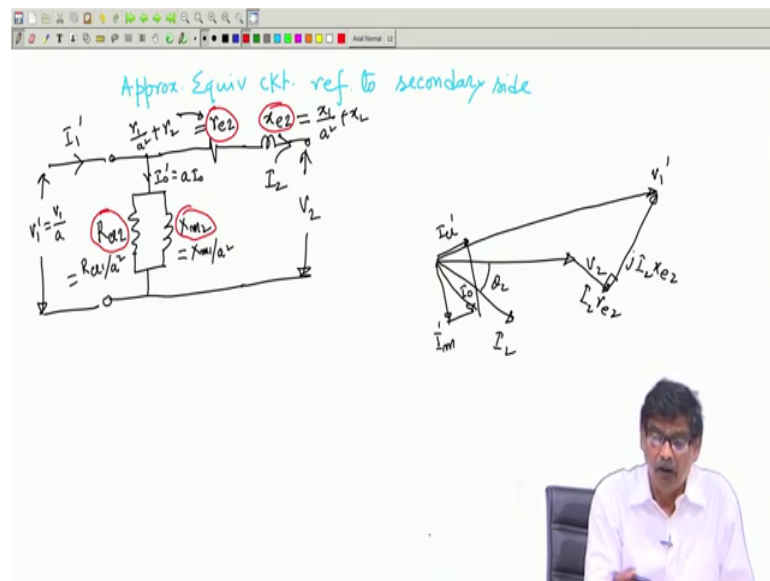
So, we start the phasor diagram drawing in this way. First draw V_2 dashed and then draw I_2 dashed. Suppose it is supplying, what is I_2 dashed? Actual current by a and this is then the load power factor angle mind you is the load power factor angle θ_2 whatever impedance you have connected. Then to get V_1 you add V_2 dashed; this drop simply.

So, this will be V_2 dashed plus I_2 dashed r_e plus $j I_2$ dashed x_e and you will get V_1 . No point in showing e_1 e_2 etcetera. So, as I am telling you as you go to the approximate equivalent circuit, you on the plea that the series impedances are much smaller compared to the parallel impedances of this core loss resistance and magnetizing impedance. Then it is the phasor diagram that is all and you this is the angle θ_2 and this is your V_2 dashed.

So, to V_2 dashed, you add this ok. Now where is your I_1 ? If this is V_1 you see this I_1 naught will be lagging this V_1 by whatever that current is. Aare you getting? After you see after you get the V_1 to V_1 , you have in this phasor in this approximate phasor diagram your I_{c1} will be in phase with your applied voltage. So, this will be I_{c1} and your magnetizing current you show 90 degree lagging in this circuit.

So, this angle will be then 90 degree as per the approximate equivalent circuit I_m is not and if you wish to calculate where your I_1 will be that is what I am doing. So, I_m and then add this to get your I_{naught} to I_2 dashed you add your this I_{naught} to get I_1 an input power factor will be angle between V_1 and I_1 . I think you have got the idea, but this is the thing. Similarly the same transformer if you draw the equivalent circuit referred to I will do that because you practice them.

(Refer Slide Time: 08:27)



So, suppose approximate equivalent circuit referred to secondary side or load side; secondary side. I will draw it very quickly because I know what is what. So, equivalent first you draw the equivalent circuit. Equivalent circuit referred to secondary side, we have already drawn.

All parameters now it will look like same thing. Only thing this is R_{cl2} . This is X_{m2} referred to secondary side and this R_{cl2} is nothing, but R_{cl1} by a square from this side to that side things are to be divided in terms of X_{m1} , it will be by a square. And I will now show it and leave it to you to verify it is indeed true; I will be right here as r_2 and

x_2 where, this x_2 will be equal to x_1 by a square plus x_2 and this r_2 will be equal to r_1 by a square plus r_2 ; this resistance.

And this voltage is certainly not V_1 now it is V_1 dashed which will be equal to V_1 by a and this voltage will be your V_2 . And this current is actual current I_2 and this is your I naught dashed which is equal to I naught a I naught is not. This is I naught dashed; I can transform this current it is by a . So, it will be a I naught dashed is equal to a into I naught.

So, get used to this transformation multiplying any impedance transfer from this side to this side will involve a and any current transformation or voltage transformation will involve a ; whether multiplication division you should be very careful. I gave you some hints also that is you just simply calculate the turns ratios and you must know that impedance values in the lb sides are lower; voltage value on the low voltage side is low definitely and so on current value in the lb side is higher.

So, all these things will you get used to it. So, this is the thing and in this case this current will be I_1 dashed and then the phasor diagram for this in the same way. Here this you start with V_2 actual secondary voltage V_2 . This is suppose the current delivered by the transformer to the load which am not drawing. Henceforth I will not draw because V_2 , it is supplying current means some impedance is connected.

This is the actual power factor of the load, then to V_2 you add these drop $I_2 r_2$ plus this angle is 90 degree $j I_2 x_2$ and this is $I_2 r_2$ and then you will get this V_1 dashed. You will not get V_1 mind you; V_1 dashed and then you can fix up where your I naught will be because V_1 by r_{cl} will give you magnetizing current. This will give you magnetizing current V_1 dashed divided by X_m and the core loss component of current will be there I_m dashed I_{cl} dashed. These two will give you I mean same stuff; I do not want to also mix.

So, I naught so, to I naught you add I_2 to get I_1 dashed and so on; I will not continue. I have done is this I have represented a practical transformer modelled it in terms of some external parameters connected either in series or in parallel. For example, r_1 r_2 R_{cl} core loss component of resistance magnetizing reactants to an ideal transformer and got these things and after getting the exact equivalence circuit we did another approximation not without any reason.

You cannot approximate anything just like that the argument is the winding resistance and leakage reactants are much smaller compared to the resistance representing core loss and the reactants representing the magnetizing current and that comes in parallel. So, it can be approximately pushed in this way [FL].

Now, after doing this, I will now tell you given a practical transformer how do you know this parameter values? By doing some tests it can be simple test, these parameters can be evaluated.

(Refer Slide Time: 15:14)

$1 \text{ Ph}, 1 \text{ KVA}, 200\text{V}/100\text{V}, 50\text{Hz tfo}$
 $a = 200/100\text{V} = 2$
 $I_{\text{HV rated}} = I_1 = \frac{1000}{200} \text{A} = 5 \text{A}$
 $I_{\text{LV rated}} = I_2 = \frac{1000}{100} \text{A} = 10 \text{A}$

no load current
 $I_0 (= I_{c1} + I_{m1})$
 2 to 5% of the rated
 for well designed tfo
 • $I_{01} = 5\% \text{ of } 5 = 5 \times 0.05 = 0.25 \text{A}$
 if energised from LV side
 $I_{02} = 5\% \text{ of } 10 \text{A} = 0.5 \text{A}$

But before that let us you would take some rating of a practical transformer. I took one like this say 200 volt, stroke 100 volt, single phase, 50 Hertz transformer and suppose it is practical transformer. And what I have missed KVA rating single phase. Suppose the KVA rating is 1 KVA ok; 1 KVA 200 volt, stroke 100 volt, 50 Hertz single phase transformer; tfo means transformer short way I am writing [FL]; this things given to you.

What does this mean to a practical transformer is this? This is practical transformer, I will assume this way that if you apply 200 volt 50 Hertz source here across this winding with nothing connected on the secondary side, you will get 100 volt 50 Hertz. And this will be terms ratio will be 200 by 100. This is what will happen this is the turns ratio I can get and this one. I also told you that HV side rated current which I will call side 1 and this is side 2 HV side and this is LV side that is equal to I 1 rated is equal to 1000 divided by 200.

So, much ampere these I can immediately calculate ok. So, how much it will be? 5 ampere and also I know I HV I LV rated I LV side rated is equal to I 2 rated is equal to 1000 by 100. Of course, these I could in one stroke right straight away 5 into this is the rated current mind you; these things I know [FL].

So, given the transformer rating we know about this numbers or this turns ratio I can make out calculate rated current that can be allowed to flow through the HV winding and through the LV winding. KVA remain same approximately [FL]. After calculating this, I now make one very interesting point that no load current of a transformer; no load current that is I_{naught} you recall I_{naught} was equal to I_{cl} plus $I_{magnetizing}$ ok.

No load current of a transformer is of the order of say 5 percent of the rated current 2 to 5 percent; 2 to 5 percent of the rated current. As a practicing engineer, the this idea must be there that is it must be a well designed transformer whoever has designed this transformer he must have seen that eddy current, hysteresis losses are reduced is using very good magnetic material.

So, that magnetizing current is also small and it I_{cl} plus I_m . These two together phasor sum gives you the no load current and that number, I am telling for a well designed transformer for a well designed transformer. Therefore, without knowing the equivalence circuit this that I will say that maybe for this transformer, I_{naught} 1 no load current referred to side one will be about say 5 percent 5 percent of rated current of side one that is 5 ampere. So, it will be 5 into 0.05 that is about 0.25 ampere.

Same transformer; if you are energized from the from this side no load current LV side; it will be I_{naught} 2. If energized from LV side, I_{naught} 2 no load current will be 5 percent of 10 ampere. Are you getting? That is 0.5 ampere anyway exact value I do not know what I am telling given a transformer, you can guess what will be the order of the no load current.

So, no load current value is much smaller compared to the rated by maybe of the order of 5 percent of the rated current. See this transformer could be energized from the 100 volt side as well that is why LV side, I call it and your no load current can be estimated. That is if I say suppose you are energizing the transformer from the HV side with secondary no load means no impedance which is open here no load means s is opened s opened. Then I am telling suppose I want to measure this current no load current. Then how to

choose the ammeter reading? It guides you from the HV side better taken ammeter which can read whose range is 0.05 ampere do not connect it 0 to 30 ampere range ammeter to record a current of 0.25 ampere. So, I can decide upon the range of the ammeter; if I know this information that is no load of this current is of the order of 0.5 percent and this 5 percent is not unique value that is given a transformer.

You expect it is it you can expect it will be close to this current, but exact current in any case I can calculate because the no load current can be about 2 to 5 percent of the rated current. You must understand that [FL]. After knowing this let us start discussing how to determine the equivalent circuit parameters.

(Refer Slide Time: 23:43)

Determination of Equivalent Ckt. parameters

Open Circuit Test & S.C Test
or no load

Approx.:

$R_{e1} = \frac{V_1}{I_{c1}}$

$X_{m1} = \frac{V_1}{I_{m1}}$

$W_0 = V_1 I_0 \cos \theta_0$

$I_0 \cos \theta_0 = \frac{W_0}{V_1} = I_{c1}$

$\cos \theta_0 = \frac{W_0}{V_1 I_0}$ from this θ_0 is known

practically records loss the loss $I_0^2 R_{e1}$ as flux is rated & I_0 is only about 2 to 5% rated current.

Recs X_{m1}

at rated voltage V_1, f at rated frequency

LV HV

Vector diagram: I_0 is the hypotenuse, I_{c1} is the adjacent side, I_{m1} is the opposite side. $I_{m1} = I_0 \sin \theta_0$, $I_{c1} = I_0 \cos \theta_0$.

Determination of equivalent circuit parameters. So, a transformer practical transformer is given I want to know what will be the values of r_{e2} , what will be the value of x_{e2} , what will be the value of R_{c12} and what should be the value of X_{m2} . On any side if you know, other side can be calculated or in the previous diagram my god.

Student: (Refer Time: 24:51).

That is this is with respect to the secondary side; I drew with respect to the primary side. Similarly our goal is to calculate this one, x_{e1} , R_{c1} and x_{m1} that is how can I calculate it?

The answer to this is we have to do some testing and simple testing; I will do. So, determination of equivalent circuit parameters by conducting two tests; open circuit test, you conduct two tests; open circuit test and short circuit test. You conduct these two tests in the laboratory and you can get those to all the parameters ok. How do I do the test? First will take up this open circuit test, what it is all about? What is done? You take the transformer and connect an ammeter and connect a Watt meter and this is the practical transform of primary.

And generally to do the open circuit test from the LV side, this test is carried out. The reason will tell you later and suppose this is the other side HV side which is open circuit or sometimes called no load test or no load. No load means no impedance is connected across the secondary and also you connect a voltmeter across this. And here you apply the V_1 rated voltage at rated frequency.

So, what I will do? LV side, I will energize with voltage V_1 at rated frequency f and connect an ammeter and Watt meter ok. And keep in mind this is the approximate equivalent circuit I will be referred to what is the approximate equivalent circuit then? Here V_1 at frequency f , you have given r_{e1} and x_{e1} are there, but these are open circuited because secondary nothing is connected in finite. So, this is that open circuit will be reflected at the open circuit in the equivalent circuit also.

So, there is I_2 dashed is 0 nothing is there because no impedance is connected. So, I_2 is not there I_2 dashed is not there and this is R_{cl1} and this is X_{m1} . So, under this condition this condition, you see this is I_{naught} this is also I_{naught} and this is I_{cl1} and this is I_{m1} magnetizing current. And what is the phasor diagram in this approximate equivalent circuit? Phasor diagram is very simple applied voltage V_1 , your magnetizing current will be here 90 degree lagging and your core loss component of current will be in phase with the supply voltage and this will be your no load current; I_{naught} .

So, your ammeter is going to read I_{naught} ok. And this is called no load power factor angle. If this is θ , this is θ ; θ_{naught} , θ_{naught} ; θ_{naught} θ_{naught} will be in an ideal transformer 90 degree, but it cannot be because it is a practical transformer having core losses ok. And this is the thing and the Watt meter suppose read W_{naught} , then I will say that this Watt meter reading must be equal to the voltage applied across its pressure coil that is V_1 current flowing through its current coil that is I_{naught} . Let us

assume wattmeter to be ideal; I mean those losses of Watt meter neglected $V_1 I_{naught}$ and cosine of the angle between these two that is θ_{naught} . This is what I will get.

And from this, I can get the value of $I_{naught} \cos \theta_{naught}$ to be equal to W_{naught} by $V_1 I_{naught}$ and note that $I_{naught} \cos \theta_{naught}$ is also equal to I_{cl1} from this right angled triangle. So, $\cos \theta_{naught} I_{cl1}$ is known [FL]. Then what you do follow? This point very carefully some students make very I mean big mistake. It is better you try to calculate I_m also; magnetizing branch ok. Instead of trying to calculate V_1 by I_{naught} that will give you the equivalent impedance things become complicated; it is much more easier as you see.

What do you do now? Since so, from these also I calculate $\cos \theta_{naught} W_{naught}$ have I done some mistake here this is not.

Student: (Refer Time: 32:42).

This will be the thing now. So, W_{naught} by $V_1 I_{naught} \cos \theta_{naught}$, but in any case I can also calculate the value of $\cos \theta_{naught}$ as $V_1 I_{naught}$.

So, from this θ_{naught} is known. If θ_{naught} is known, then from this triangle I can say I_{m1} is equal to I_{naught} which is known ammeter reading into $\sin \theta_{naught}$. And also I_{cl1} I have already written it is equal to $I_{naught} \cos \theta_{naught}$. Therefore, what we have done I have got this current I_{cl1} and I_{m1} and my problem is to calculate R_{cl1} ; R_{cl1} , then simply will become V_1 by I_{cl1} . This is known this is known and X_{m1} is a equal to V_1 by I_{m1} . So, these two values can be calculated.

Instead of trying to calculate the impedance V_1 by I_{naught} , do not do it because that impedance will then be the series representation of R_{cl} and X_m equivalent. And then you are gone and do not say the series part of this equivalent circuit is the core loss resistance. Then it will define the physical reasoning that is the magnetizing current requires a fixed voltage and this one.

So, you should be very careful. So, mind you then I have been able to calculate R_{cl1} and X_m . So, doing this open circuit test, I can very quickly calculate by noting down the ammeter reading and wattmeter reading and these voltmeter reading that is that will read

equal to V_1 . I can calculate R_{cl1} and X_{m1} . And then am telling, if it is needed, you can then after knowing R_{cl1} , I can always calculate R_{cl2} and X_{m2} .

For these, I do not have to carry out the experiment once again because I know it will be either divided or multiplied by s^2 . So, correctly if I do that I will get R_{cl2} as well as X_{m2} . So, in this equivalent circuit, I have got this one only just one point. As I told you, this is the equivalent circuit approximate equivalent circuit approximate equivalent circuit. But if you honestly look at the circuit, you will find this no load current is flowing through winding resistance as well, is not.

Because the exact equivalence circuit is what? R_{1x1} , then your this parallel branch is not suppose and then of course, these two are open circuited exact equivalence circuit. And you are doing on a that practical transformer whose exact equivalent circuit is this what we have done. Here we have applied a voltage and this current you are telling I naught fine, but then I will say look here this wattmeter reading W_{naught} ok. We have done some approximation that I have understood, but this Watt meter reading strictly speaking, you will read the copper loss in R_{cl} as well as copper loss in r_1 .

Some of these two will be read is not, but what happens is this how to justify that why we neglect this copper loss. It is because of that in open circuit test, winding current is pretty small; maybe 2 to 5 percent of the rated current. But applied voltage is rated at rated frequency level of flux in the core is rated and it is the level of flux that decides the core loss. So, core loss will be much higher than these very small current causing a power loss in r_1 .

So, we neglected the copper loss note that W_1 . Practically records the core loss only; core loss. The copper loss which is also called copper loss the winding power loss the copper loss the loss $I_1^2 r_1$ into $I_{naught}^2 r_1$ is much less compared to core loss, why? Because flux is rated as flux is rated and I_{naught} is only about 2 to 5 percent of the rated current. Anyway please go through this part and try to understand each bit of it because this is the approximate equivalent circuit ok. If you forget about the exact thing from your mind it is it, then what meter this copper loss; I mean core loss only.

But that is not the case, there will be a little bit of copper loss in the winding which we are neglecting. We are attributing the wattmeter reading to be fully the core loss. We will continue with this with the SC test to determine these this parameters in the next class.

And remember that generally the reason, I have not told yet the open circuit test is carried out from LV side; from LV side means there you connect the supply HV side is kept open; anyway.

Thank you.