

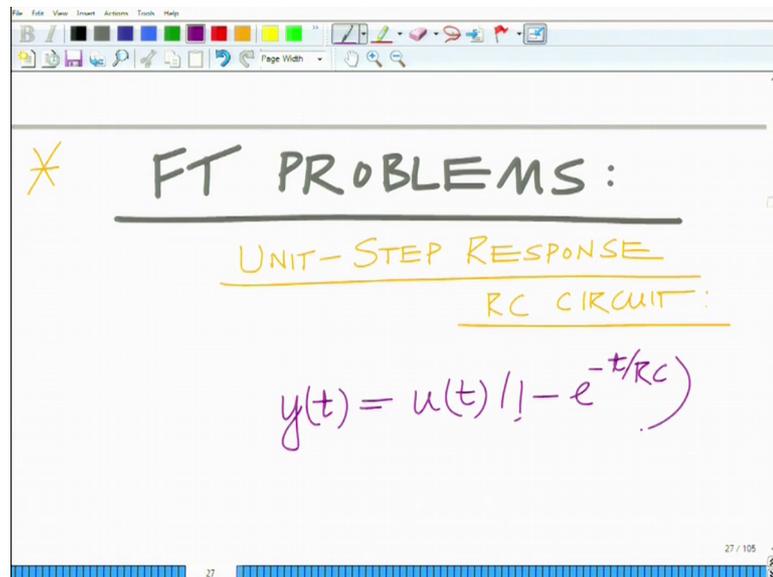
**Principles of Signals and Systems**  
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**Lecture - 51**

**Fourier Transform Problems: Unit Step Response of RC Circuit, Sampling of Continuous Signal**

Hello. Welcome to another module in this massive open online course. We are looking of problems in the Fourier transform and specifically the unit step response of the serial RC circuit all right.

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So, we have example problems in the Fourier for the Fourier transform and we are looking at unit response of the RC circuit, unit step response of the RC circuit ok. And what we have seen is the unit step response, is given us the output is y t equals. If you remember y t equals u t into 1 minus e raise to minus t over R C, this is the unit step response ok.

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$$y(t) = u(t) \left( 1 - e^{-t/RC} \right)$$

$$t \rightarrow \infty$$

$$y(\infty) = 1$$

Final value = 1

$$y(0) = 1 \times (1 - 1)$$

$$= 0$$

Now, you can see the final value that is if you look at  $y$   $t$  equal to infinity or  $t$  tends to infinity let us say. We can call that  $y$  of infinity equals well, as  $t$  tends to infinity  $e$  raise to minus  $t$  over  $RC$  that tends to zero. So, as  $t$  raise to  $t$  tends to infinity the final output is 1 all right. So, all the voltage will be across the capacitor ok. So,  $y$  infinity equals 1. Now if you look at the initial value.

Now  $y$  zero that is also fairly clear,  $y$  is zero equals well  $u$  of zero which is 1 or that is basically  $u$  of zero which is 1 into 1 minus  $e$  raise to minus  $t$  over  $RC$  that is 1. So, this is zero. So, it starts from zero and raises to 1; that is what we have seen. Now final value is 1. So, this is the final value  $y$  infinity, this is final value is 1. Now, remember the we define the raise time, as the time taken by the  $RC$  circuit or the capacitor voltage to go from 90 percent of the final value have to go from 10 percent of the final value to 90 percent of its final value.

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Handwritten mathematical derivation on a whiteboard:

$$\begin{aligned} &10\% \text{ of Final value} \\ &= 0.1 y(\infty) \\ &= 0.1 \\ (1 - e^{-t_{10}/RC}) &= 0.1 \\ \Rightarrow e^{-t_{10}/RC} &= 0.9 \end{aligned}$$

Now 10 percent of its final value that is 10 percent of 1; that is 0.1 into 1 y infinity equals 0.1. Now what is the time we have 1 minus e raise to minus? Let us call this as your  $t_{10}$ ; that is time taken for 10 percent of the final value 1 minus e raise to minus  $t_{10}$  over RC is point 1 implies e raise to minus  $t_{10}$  over RC equals 1 minus point 1; that is 0.9 implies your  $t_{10}$  equals minus RC log natural of 0.9.

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Handwritten mathematical derivation on a whiteboard:

$$\begin{aligned} &= 0.1 y(\infty) \\ &= 0.1 \\ (1 - e^{-t_{10}/RC}) &= 0.1 \\ \Rightarrow e^{-t_{10}/RC} &= 0.9 \\ \Rightarrow \boxed{t_{10} = -RC \ln 0.9} \end{aligned}$$

Time for 10% of Final val

So, this is your time taken for 10 percent of the final value ok, 10 percent of final value. Now, time taken for the, for 90 percent of the final value 0.9 into 1 that is 0.9 ok.

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Handwritten mathematical derivation on a whiteboard:

$$\begin{aligned} &90\% \text{ of } y(\infty) \\ &= 0.9 \times 1 \\ &= 0.9 \\ &1 - e^{-t_{90}/RC} = 0.9 \\ &\Rightarrow e^{-t_{90}/RC} = 0.1 \\ &\Rightarrow \boxed{t_{90} = -RC \ln 0.1} \end{aligned}$$

Now time taken for 90 percent of the final value let us call that as  $t_{90}$  minus  $e$  raise to minus  $t_{90}$  by  $RC$  equals  $0.9$ . Now this implies that  $e$  raise to minus  $t_{90}$  over  $RC$  equals point  $1$ , which implies  $t_{90}$  equals minus  $RC \ln$  or log natural of point  $1$ . So, we have. So, this is basically your time taken for to reach 90 percent

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Handwritten mathematical derivation on a whiteboard, including an annotation:

$$\begin{aligned} &= 0.9 \\ &1 - e^{-t_{90}/RC} = 0.9 \\ &\Rightarrow e^{-t_{90}/RC} = 0.1 \\ &\Rightarrow \boxed{t_{90} = -RC \ln 0.1} \end{aligned}$$

Time taken for 90% of final value

So, this  $t_{90}$  is basically time taken to reach 90 percent of the final value ok. So, this is basically. So, this is the time taken to reach 90 percent of the final value and therefore the raise time is basically time taken to reach 90 percent of the final value from 10 percent.

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The image shows a screenshot of a presentation software window with a whiteboard background. The whiteboard contains the following handwritten mathematical derivation for the rise time of an RC circuit:

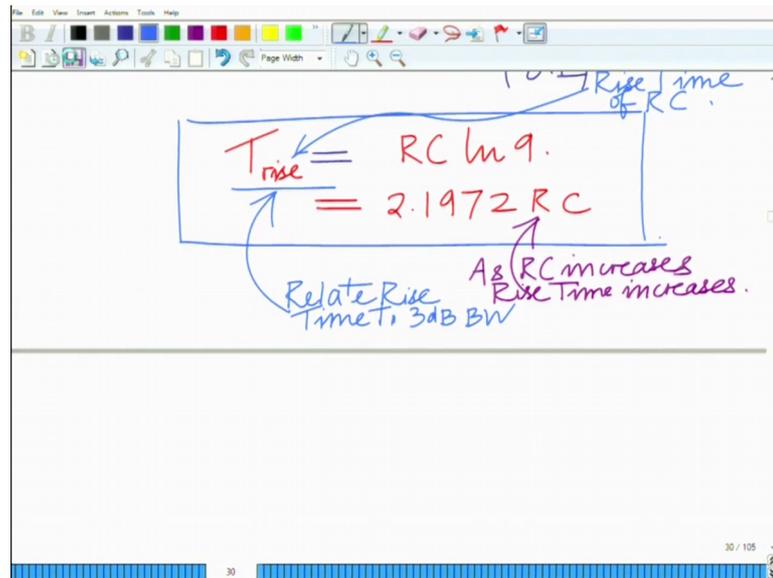
$$\begin{aligned} \text{Rise Time} &= t_{90} - t_{10} \\ &= -RC \ln 0.1 + RC \ln 0.9 \\ &= RC \ln \left( \frac{0.9}{0.1} \right) \\ &= RC \ln 9 \end{aligned}$$

The final result is boxed in blue and labeled as  $T_{\text{rise}} = RC \ln 9 = 2.1972 RC$ . A blue arrow points from the boxed result back to the  $t_{90} - t_{10}$  expression. A blue bracket on the right side of the boxed result is labeled "Rise Time of RC".

So, that is basically your  $t_{90}$  minus  $t_{10}$  which is basically minus  $RC \ln 0.1$  plus  $RC \ln 0.9$  which is basically  $RC \ln \left( \frac{0.9}{0.1} \right)$  which is basically  $RC \ln 9$ . So, this is your  $T_{\text{rise}}$  which is  $RC \ln 9$ , which is basically 2.1972 into  $RC$ .

So, this is the expression for the rise time. So, this is the rise time of  $RC$ ; that is 2.1972  $RC$  or  $\ln 9$  that is  $\ln 9$  to the base  $e$  times  $RC$ , this is the rise time all right. And you can clearly see as  $RC$  is increasing basically the time constant is increasing, basically the rise time is also increased ok, so no surprise is there. So, as and you can note that as  $RC$  is increasing rise time is also increasing as  $RC$  increases the rise time increases ok.

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A screenshot of a presentation slide showing handwritten mathematical derivations. The main equation is  $T_{rise} = RC \ln 9 = 2.1972 RC$ . A blue box encloses this equation. An arrow points from the text "Relate Rise Time to 3dB BW" to the boxed equation. Another arrow points from the text "As RC increases Rise Time increases." to the boxed equation. A third arrow points from the text "Rise time of RC" to the boxed equation. The slide number "30 / 105" is visible in the bottom right corner.

$$T_{rise} = RC \ln 9$$
$$= 2.1972 RC$$

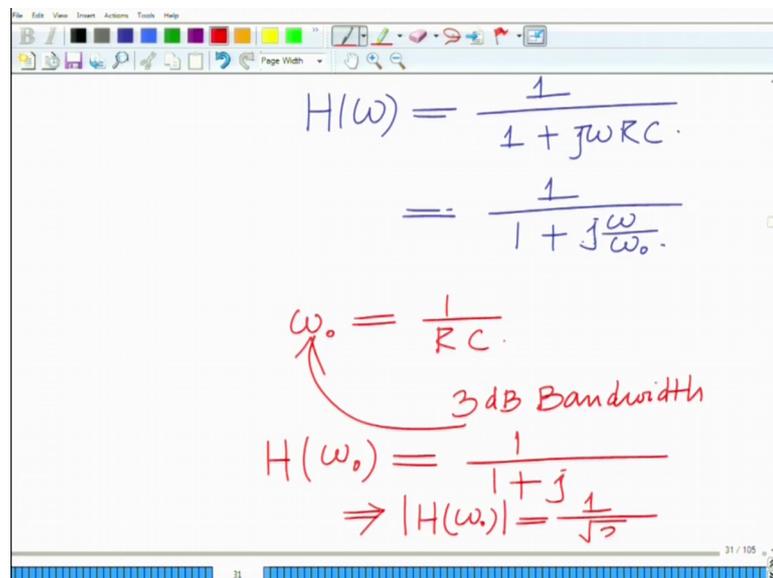
Relate Rise Time to 3dB BW

As RC increases Rise Time increases.

Rise time of RC

Now, observe that, now we want to relate rise time to the 3 dB bandwidth. So, relate the rise time to the 3 dB bandwidth. Now observe that the response of the RC circuit

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A screenshot of a presentation slide showing handwritten mathematical derivations. The transfer function is given as  $H(\omega) = \frac{1}{1 + j\omega RC} = \frac{1}{1 + j\frac{\omega}{\omega_0}}$ . Below this, the corner frequency is defined as  $\omega_0 = \frac{1}{RC}$ . An arrow points from this definition to the text "3dB Bandwidth". The magnitude of the transfer function at the corner frequency is then shown as  $H(\omega_0) = \frac{1}{1 + j} \Rightarrow |H(\omega)| = \frac{1}{\sqrt{2}}$ . The slide number "31 / 105" is visible in the bottom right corner.

$$H(\omega) = \frac{1}{1 + j\omega RC}$$
$$= \frac{1}{1 + j\frac{\omega}{\omega_0}}$$
$$\omega_0 = \frac{1}{RC}$$

3dB Bandwidth

$$H(\omega_0) = \frac{1}{1 + j}$$
$$\Rightarrow |H(\omega)| = \frac{1}{\sqrt{2}}$$

And this is what we already seen the transfer function is 1 over 1 plus j omega RC; this is the transfer function of the RC circuit, which is basically you can see, this is basically 1 over. Well you can write this as 1 plus j omega over 1 plus j omega over omega naught, where omega naught, now if you recall omega naught equals 1 over RC. Now this is termed as the 3 dB bandwidth.

Now, you remember, might also remember that this is equal to the 3 dB bandwidth, because  $H(j\omega)$ ; that is the transfer function is  $1/(1 + j\omega RC)$  which implies magnitude of  $H(j\omega)$  is  $1/\sqrt{2}$ , correct.

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The image shows a whiteboard with handwritten mathematical derivations. At the top, it states  $\Rightarrow |H(j\omega)| = \frac{1}{\sqrt{2}}$ . Below this, it shows  $\Rightarrow |H(j\omega_c)|^2 = \frac{1}{2}$ . A green arrow points from the  $\frac{1}{2}$  to the text "Output power suppressed by  $\frac{1}{2}$ " which is followed by "= -3dB". Finally, it concludes with "Therefore,  $\omega_c = \frac{1}{RC}$ " and "= 3 dB BW".

And therefore, the magnitude  $H(j\omega)$  squared equals half. Therefore the power that is transferred, the output power is half corresponding to that frequency all right, which basically corresponds to a 3 dB; 3 dB suppression in the power of the signal that is input to this RC ok. So, the output power is subdued or suppressed by the factor of half, so output power correct. So, magnitude  $|H(j\omega)|^2$  is the, which corresponds to, which equals to minus 3 dB reduction in power. And hence therefore  $\omega_c$  equals  $1/RC$  equals the 3 dB band value. And now that gives us does not interesting relation.

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The whiteboard shows the following handwritten equations:

$$T_{rise} = 2.1972 RC$$
$$\omega_{3dB} = \frac{1}{RC}$$
$$\Rightarrow F_{3dB} = \frac{\omega_{3dB}}{2\pi}$$

---

$$= \frac{1}{2\pi R C}$$

At the bottom right of the whiteboard, the page number "33 / 105" is visible.

So we have remember raise time equals, the raise time equals, we have just seen that 2.1972 times RC and 3 dB bandwidth omega 3 dB equals 1 over RC implies the 3 dB frequency equals omega 3 dB by 2 pi which is equal to 1 over 2 pi R C, which implies your RC equals 1 over 2 pi times the 3 dB frequency.

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The whiteboard shows the following handwritten equations and steps:

$$2\pi R C$$
$$\Rightarrow RC = \frac{1}{2\pi F_{dB}}$$

(2)

Substitute RC from (2) in (1)

$$\Rightarrow T_{rise} = \frac{2.1972}{2\pi F_{3dB}}$$
$$\Rightarrow T_{rise} = \frac{0.35}{F_{3dB}}$$

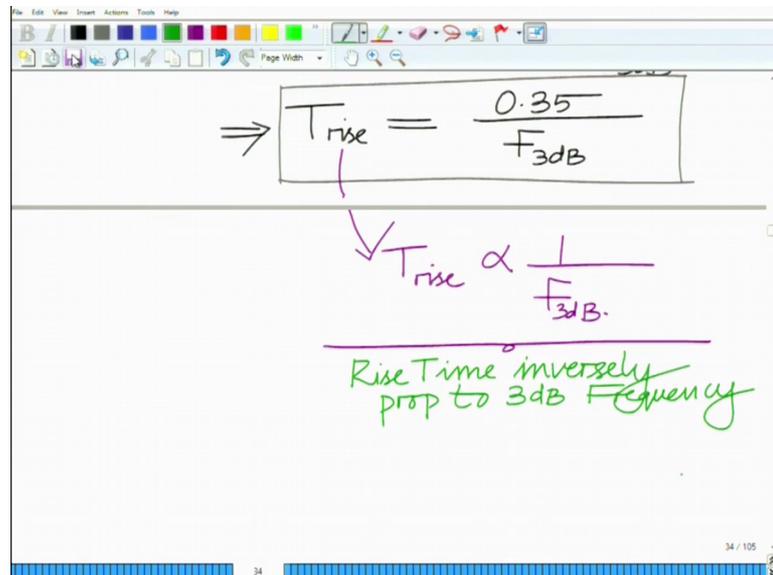
At the bottom right of the whiteboard, the page number "33 / 105" is visible.

Now what the only thing that is remaining is basically if you call this, let us say two and you call this let us say 1 and all that is remaining is substituting this value of RC. So, substituting RC from 2 in 1 this gives us the result that T raise equals 2.1972 times RC,

but RC is  $1 / (2.5 \times f_{3dB})$ , the 3 dB frequency implies the rise time is  $2.1972 / \pi$ ; that is basically 0.35 over the 3.

So, basically that gives us the relation that rise time is point. So, rise time is 0.35 over the 3 dB frequency (Refer Time: 13:32) which is very interesting what this tells us, is that the rise time is inversely proportional to the 3 dB frequency, which means of the 3 dB frequency is large right, the filter has the large bandwidth correct, the rise time is much smaller. On the other hand in the three different frequency is small, there is a filter has a very small band width; the rise time is much larger ok.

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The image shows a digital whiteboard with a toolbar at the top. The main content is handwritten in black and purple ink. It starts with an arrow pointing to a boxed equation:  $T_{rise} = \frac{0.35}{F_{3dB}}$ . Below this, a purple arrow points to the text  $T_{rise} \propto \frac{1}{F_{3dB}}$ . Underneath that, a green note reads "Rise Time inversely prop to 3dB Frequency". The bottom right corner of the whiteboard shows "34 / 105".

So, what this tells us is the rise time is, the rise time is basically inversely proportional to  $1 / 3$  dB frequency ok. So, rise time is inversely proportional, let us also note that the rise time is inversely proportional to the 3 dB frequency which is a very interesting thing. So, basically this problem illustrates the concept of this 3 dB frequency, how to derive the concept of rise time, how to derive the rise time all right and finally, its relation to the 3 dB frequency at the RC circuit all right.

Let us now look at another very important aspect, one of the very, one other very important application, one other very important concept of the Fourier transform is known as the sampling which is related to the sampling of a continuous time and a log signal. So, let us talk about.

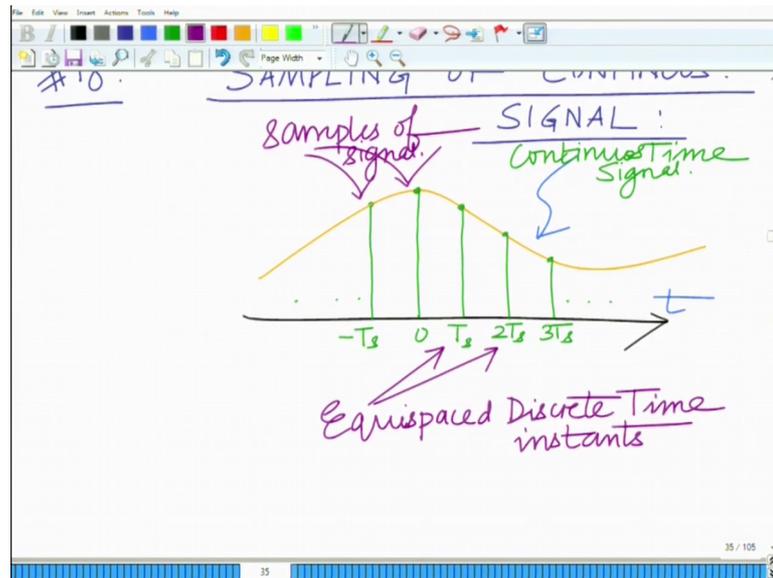
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Now, let us come to another yet another important concept; that is the concept of sampling ok. So, this is let us treat this as problem number 18 and this is although its an example problem, this is a very important concept by itself, this is sampling of continues signal. Now consider a signal, a continues time signal which looks something like this ok. And now this is a continues time signal ok, this is  $t$ , this is the continues time signal, this is a continues time signal.

Now before converting it to digital signal we would like to convert it to a discrete time signal. So, to convert this continues time signal into a discrete time signal, we have to look at the values of the signal at discrete time instants or we have to extract the values of this signal values at discrete time signal, discrete time instants which are equispaced. This process is termed as sampling; that is converting this continues time analogue signal into a discrete time signal, this is termed as sampling.

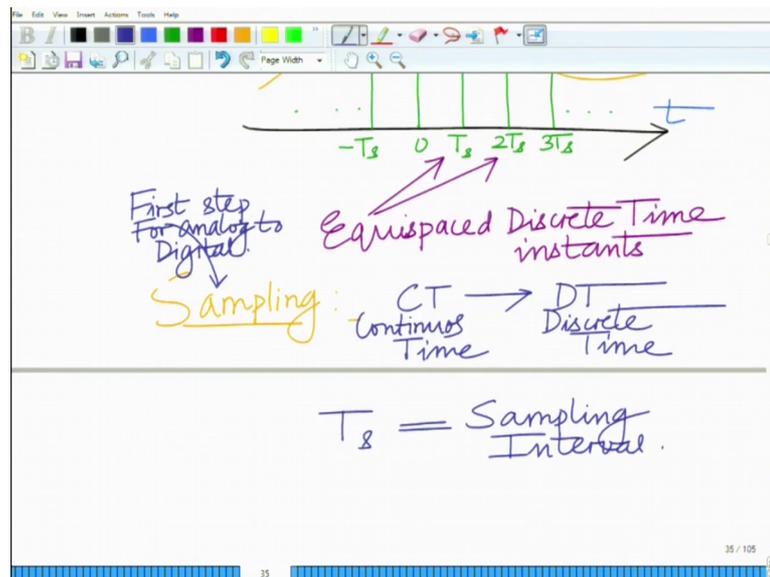
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So, we place different time instants at which we extract the values of the. So, let us say this is zero, this is  $T_s$ , this is  $2T_s$ ,  $3T_s$  minus  $T_s$  and so on. So, these are the discrete time instants ok.

So, these are the various equispaced; equi-space discrete time instants and these are the samples, samples and extracting the samples of the signal at this discrete time instances all right, this process correct is termed as sampling ok. So, sampling is a very interesting concept and one of the most fundamental concepts which has a very interesting theoretical framework, which is basically analogue to or continuous time, continues time; that is C T to D T; that is continuous time.

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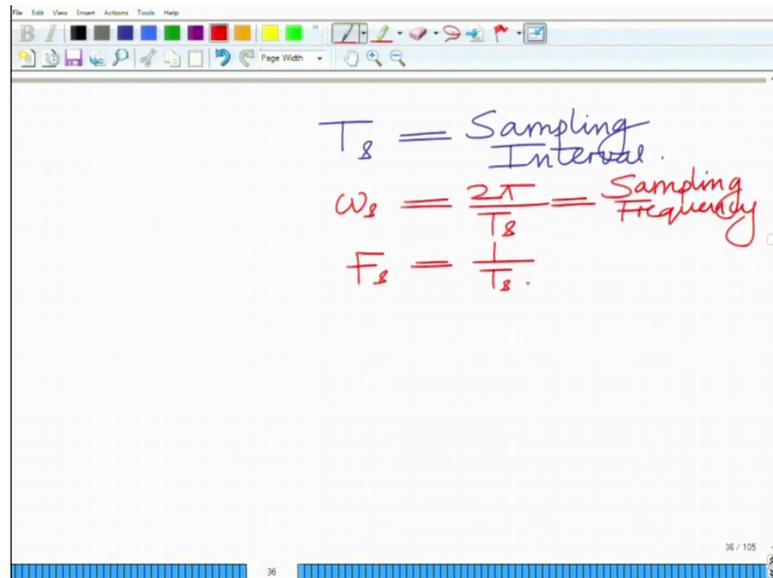


This is what sampling does, you have a continuous time signal converted into a discrete time signal by extracting samples at equispaced instance or equispaced intervals of time and this interval of time which is of duration  $T_s$ . So, we are extracting samples at zero  $T_s$ , twice  $T_s$  and minus  $T_s$  minus 2. So, we are extracting samples at multiples of  $T_s$ , this quantity  $T_s$  is known as the sampling interval ok.

So, this is the sampling duration  $T_s$ . This quantity  $T_s$  is termed as a sampling interval and this is the first step. So, sampling is basically to convert, this is first step conversion of analogue to for analogue to digital. Eventually to convert this analogue signal into a digital signal, sampling is the first step and it is. Therefore, one of the, one of the most important term, because frequently the natural signals are analogue and to store and transmit signals, we convert them into digital signals.

Alright, which is the most popular technology both for storage and communications of sampling plays a key role in that and it is very important to have understand the concepts, various concepts and the various aspects of sampling in good detail ok. Now if you look at the sampling. Now this  $T_s$  is as we said is sampling interval. Now this quantity  $\omega_s$  equals  $2\pi$  over  $T_s$ , you can call it call this as a sampling frequency or basically the angular frequency of sampling.

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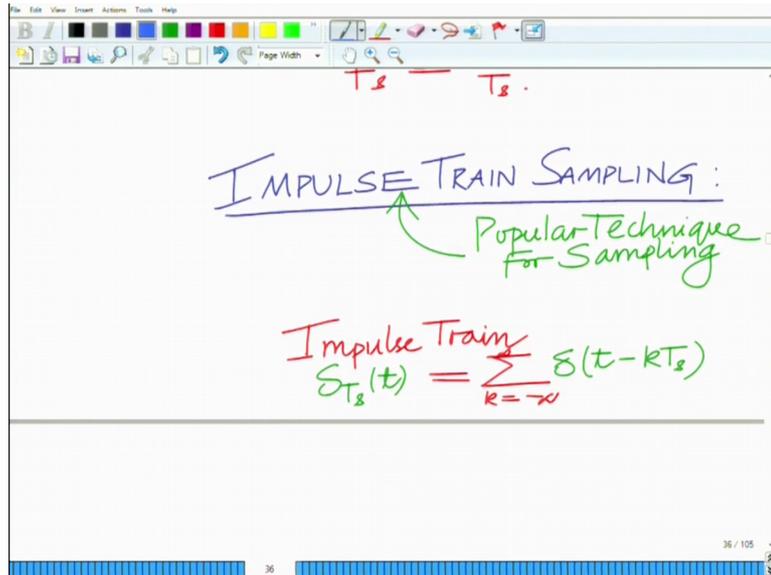
$$T_s = \text{Sampling Interval.}$$
$$\omega_s = \frac{2\pi}{T_s} = \text{Sampling Frequency}$$
$$F_s = \frac{1}{T_s}.$$

The whiteboard interface includes a toolbar at the top with various drawing tools and a status bar at the bottom showing '36 / 105'.

So, this is equal to sampling frequency. This is the angular sampling frequency domain. The sampling frequency can also be denoted by  $F_s$  equals  $1$  over  $T_s$  ok. So, one is the frequency, the other is the angular frequency all right  $2\pi$  over  $T_s$   $1$  over  $T_s$  all right, the frequency with which samples are being taken for this continuous time signal ok.

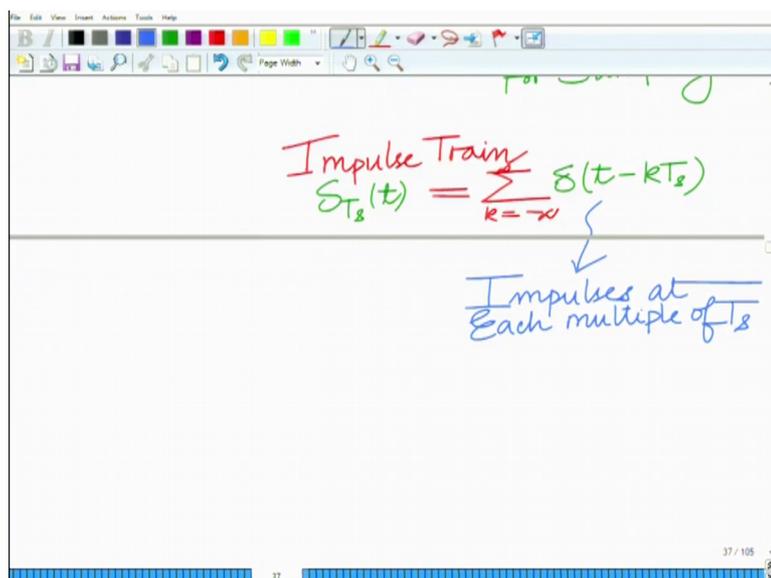
And now there are various techniques of sampling, one of the most simplest methods of sampling, is basically have an impulse scaled by the value of the signal at each multiple of  $T_s$ ; that is basically what you do is you have impulses which are periodically spaced or equispaced at  $T_s$  at multiples of  $T_s$ , and you scale the amplitude or you scale each impulse by the value of the signal at that particular point. This is known as impulse train sampling. In other words you have this impulse train of impulses right which are of unit scaling and you multiply it with the signal ok.

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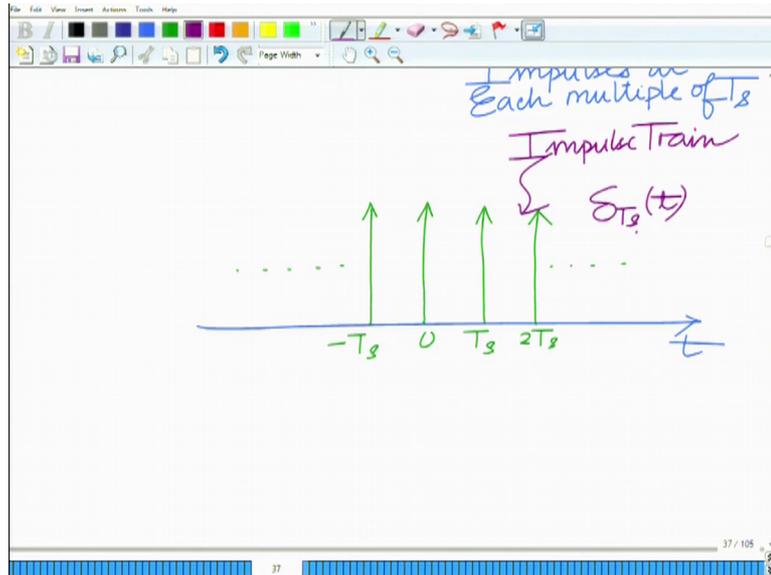


So, what you have, is one of the most popular techniques for sampling is what is known as impulse train. So, this is one of the most popular technique; one of the most popular technique for sampling is impulse train sampling. Now the impulse train is basically you can write this as summation  $k$  equals minus infinity to infinity delta  $t$  minus  $k T_s$ ; that is delta  $t$  minus  $k T_s$ ; that is impulses of unit scaling at each multiple of  $T_s$ . Unit impulses at, unit impulses at each multiple of  $T_s$ . So, this is basically termed as your impulse train.

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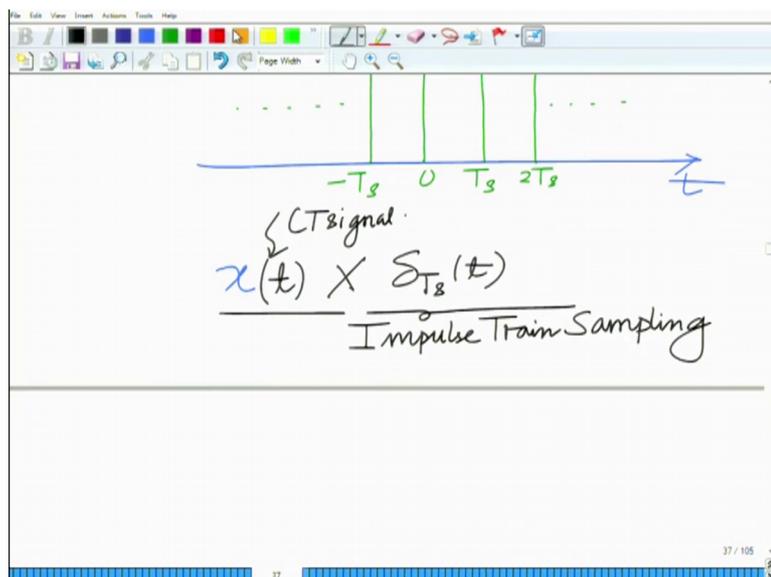
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So, what you have is, you have that time axis and at each multiple of  $T_s$  you have unit impulses. So, this is at zero  $T_s$  twice  $T_s$  minus  $T_s$ . So, this is basically your impulse train or your train of impulses, ok.

. So, this is your impulse train, that is impulses spaced at equal intervals on the time axes in multiples of  $T_s$ , where  $T_s$  is the sampling interval ok, and this is your  $\delta_{T_s}(t)$  ok, this is your impulse train. Now what we are doing is we have our continuous time signal.

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So, impulse train sampling is simply you take the continuous time signal  $x(t)$ , this is your C T signal and you multiply it with the impulse train. This is termed as your impulse train sampling. This is your impulse train sampling which is basically

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$$= x(t) \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$$

$$= \sum_{k=-\infty}^{\infty} x(t) \delta(t - kT_s)$$

$$x_{T_s}(t) = \sum_{k=-\infty}^{\infty} x(kT_s) \delta(t - kT_s)$$

SAMPLED SIGNAL

Now if you look at this, this is  $x(t)$  summation  $k$  equal to minus infinity to infinity  $\delta(t - kT_s)$  which is equal to summation  $k$  equals minus infinity to infinity  $x(t)$  into  $\delta(t - kT_s)$ . And now you can see  $\delta(t - kT_s)$  multiplies by  $x(t)$  the property of the impulse function, basically extracts the value of the signal at the sampling instant corresponding sampling instant is  $kT_s$  ok.

So, for instances you use the property that  $x(t)$  multiplied by  $\delta(t - t_0)$  is equal to essentially  $x(t_0)$  into  $\delta(t - t_0)$ . This is the impulse space  $t_0$  and multiplied by  $x(t)$ , extracts the samples, extract the samples of that value  $t_0$  ok. So, this is equal to your  $k$  equals minus infinity to infinity  $x(kT_s) \delta(t - kT_s)$  and, and this you can denote this as the sampled signal ok. So, this is basically your  $x(t)$ , this is basically your. So, this is basically the sampled signal ok. And this is basically you can see the sample value of the signal at sampling instant  $kT_s$ , sample value, this is the sample value of the signal at the sampling instant  $kT_s$ .

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$$x_{T_s}(t) = \sum_{k=-\infty}^{\infty} x(kT_s) \delta(t - kT_s)$$

Spectrum of sampled signal.

Sample value of signal at  $kT_s$ .

SAMPLED SIGNAL

Now, what you want to do as basically we want to study this sampling process in detail, and a very interesting view point or a very int and lot of insides what the sampling process can be gained by looking into the Fourier transform, as what happens to the spectrum of the signal that is sampled. So, what you want to do to understand this twice process better, is to look at the spectrum, we want to look at the spectrum of the sampled signal; that is if we want to ask the question, if  $x(t)$  has the spectrum  $X(\omega)$  then what is the spectrum of this quantity  $x_{T_s}(t)$  that is the sampled signal.

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$$x(t) \longleftrightarrow X(\omega)$$

$$x_{T_s}(t) \longleftrightarrow ?$$

$$S_{T_s} = \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$$

PERIODIC SIGNAL  
 Period =  $T_s$ .

What is the spectrum of the sampled signal that is the question that we would like to address. Now so we would likely find the spectrum of the resulting sampled signal to do that first let us start with the spectrum of the impulse train ok.

So, you know the impulse train, now if you take a look at this impulse train equals summation k equals minus infinity to infinity unit impulse functions at each multiple of T s, each integer multiple of T s. Now you can see this is the periodic signal. This is the periodic signal and the period is nothing, but T s, the period is basically T s and you can see that this is basically. If you look at minus T s if you look at this here, you will see that this is basically periodic signal.

So, you have one impulse in every interval of T s, at every duration of after every duration of T s all right; so impulse at 0 T s 2 T s minus T s minus 2 T s of. So, this is periodic and the time period, fundamental time period is basically T s and. In fact, the fundamental frequency of this is 1 over T s the fundamental angular frequency is 2 pi over T s ok. So, that is the first thing that we note ok. So, this is periodic period is T s fundamental frequency was 2 pi over T s which is equal to, which is equal to omega s.

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$$S_{T_s} = \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$$

PERIODIC SIGNAL  
 Period =  $T_s$   
 Fund Freq =  $\frac{2\pi}{T_s}$   
 $= \omega_s$

Therefore, we can derive the CEF S

Now there for this has a complex exponential Fourier series, because it is a periodic signal. Therefore, we can develop or derive the C E F S; that is the complex exponential Fourier series that is. I can express this as delta T s of t equals summation k equals minus infinity to infinity C k e raise to j 2 pi k or e raise to j, simply e raise to j k omega s t,

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The image shows a digital whiteboard with handwritten notes. At the top, it says "period = T\_s". Below that, the fundamental frequency is calculated as  $\text{Fund Freq} = \frac{2\pi}{T_s} = \omega_s$ . A blue arrow points from this result to the text "Therefore, we can derive the CEFS". Below a horizontal line, the CEFS equation is written as 
$$S_{T_s}(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_s t}$$

Where  $\omega_s$  is the fundamental frequency; that is basically linear combination, there is a combination of an infinite number of basically sinusoids, harmonics, corresponding to the fundamental frequency  $\omega_s$  and multiples of the fundamental frequency; that is harmonics right multi, harmonics at multiples of the fundamental frequencies that is  $k\omega_s$  ok. And we can derive this coefficients of the CEFS as the complex exponential Fourier series  $C_k$  and from that we can basically one can derive where is the resulting spectrum of the sampled signal.

So, I will stop this module here and continue in the subsequent module.

Thank you very much.