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Find $h(t)$, Eigenvalue for function $e^{st}u(t)$.

impulse response

$$= \int_{-\infty}^{\infty} \frac{e^{-j\omega_0(t-\tau)} u(t-\tau) \cdot x(\tau) d\tau}{h(t-\tau) x(\tau)}$$

$$= h(t) * x(t)$$

\Rightarrow impulse response $h(t) = e^{-j\omega_0 t} u(t)$

* PROBLEMS FOR

The impulse this can be written as integral minus infinity to t e to the power of minus j omega naught, t minus tau u t minus tau. I can change the limits from minus infinity to infinity tau because I am incorporating u t minus tau times x tau d tau, now you can see this is your h t minus tau this is of course x tau is t minus tau into x tau this is h t convolve with x t.

The impulse response is h t times impulse response h t is e raised to minus j omega naught t times u t, that is the impulse response let us note that down.

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$$h(t) = e^{-j\omega_0 t} u(t)$$

$$h(t) * x(t) = \int_{-\infty}^{\infty} e^{-j\omega_0(t-\tau)} u(t-\tau) x(\tau) d\tau$$

$$= \int_{-\infty}^{\infty} e^{-j\omega_0 \tau} u(\tau) \cdot x(t-\tau) d\tau$$

Small correction we have the impulse response $h(t)$ equals $e^{-j\omega_0 t} u(t)$, now to find the Eigen value corresponding to e^{st} , we note that this can also be written as the impulse response $h(t)$ convolved with $x(t)$ that is $e^{-j\omega_0(t-\tau)} u(t-\tau)$ or $e^{-j\omega_0(t-\tau)} u(t-\tau)$ slightly this way it has to be $e^{-j\omega_0(t-\tau)} u(t-\tau) x(t-\tau) d\tau$.

Now, this I can also write this as $x(t)$ convolved with $h(t)$ which means this I can write it as $e^{st} u(t-\tau) x(t-\tau) d\tau$, now since $u(t-\tau)$ is nonzero only for $t \geq \tau$.

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$$= \int_0^{\infty} e^{-j\omega_0 z} x(t-z) dz$$

$$x(t) = e^{st}$$

$$= \int_0^{\infty} e^{-j\omega_0 z} e^{s(t-z)} dz$$

$$= e^{st} \int_0^{\infty} e^{-(s+j\omega_0)z} dz$$

This can be equivalently written as $\int_0^{\infty} e^{-j\omega_0(t-\tau)} x(t-\tau) d\tau$ just equivalently writing this as $\int_0^{\infty} e^{-j\omega_0(t-\tau)} x(t-\tau) d\tau$ and now I am going to substitute our function or signal $x(t)$ equals e^{st} and therefore, this becomes equal to $\int_0^{\infty} e^{-j\omega_0(t-\tau)} e^{s(t-\tau)} d\tau$. We bring the e^{st} common outside. This is $\int_0^{\infty} e^{-j\omega_0(t-\tau)} e^{s(t-\tau)} d\tau$ this is $e^{st} \int_0^{\infty} e^{-(s+j\omega_0)\tau} d\tau$.

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$$\begin{aligned}
 & \rightarrow = \int_0^{\infty} e^{st} e^{(s+j\omega_0)\tau} d\tau \\
 & = e^{st} \int_0^{\infty} e^{-(s+j\omega_0)\tau} d\tau \\
 & = e^{st} \cdot \left. \frac{e^{-(s+j\omega_0)\tau}}{-(s+j\omega_0)} \right|_0^{\infty} \\
 & = e^{st} \cdot \left(0 - \frac{1}{-(s+j\omega_0)} \right) \\
 & = e^{st} \times \frac{1}{(s+j\omega_0)} \quad s > 0
 \end{aligned}$$

Which is e raised to st times e raised to $-(s+j\omega_0)\tau$ evaluated between the limits 0 to ∞ . Now, at ∞ this is 0 if s is less than 0 if s is greater than 0 . This is equal to e^{st} and at 0 this; obviously, 1 , into s 0 minus 1 by $-(s+j\omega_0)$, but for this to be 0 , s has to be greater than 0 that if e to the power of $-(s+j\omega_0)\tau$ has to be a decaying exponential. So, this is e to the power of $-(s+j\omega_0)\tau$. e to the power of $-(s+j\omega_0)\tau$ has to be a decaying exponential which means s has to be greater than 0 .

This is basically reduces to e to the power of st into $\frac{1}{s+j\omega_0}$ this is the Eigen value.

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$$= e^{st} \cdot \frac{e^{-(s+j\omega_0)t}}{-(s+j\omega_0)} \Big|_0^\infty$$

$$= e^{st} \cdot \left(0 - \frac{1}{-(s+j\omega_0)} \right)$$

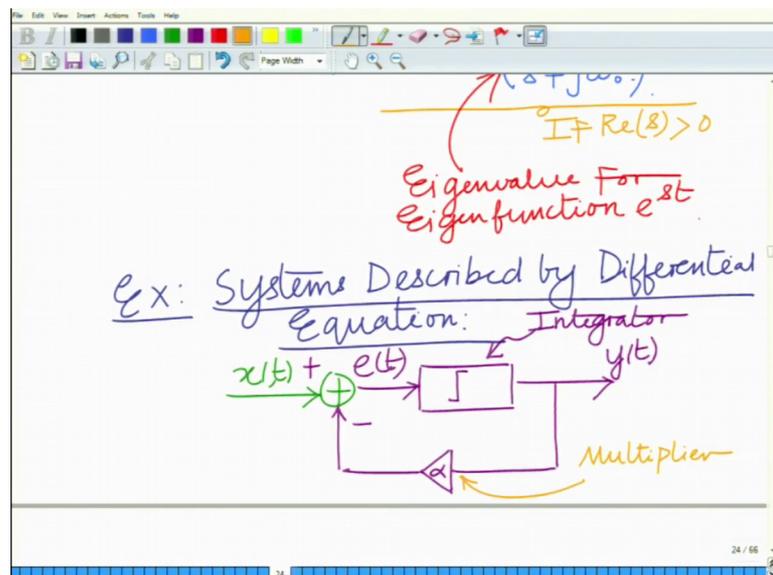
$s > 0$

$$= \frac{e^{st} \times \left(\frac{1}{s+j\omega_0} \right)}{I \nrightarrow \text{Re}(s) > 0}$$

Eigenvalue For Eigenfunction e^{st}

If real part of s is greater than 0, the condition required for this integral real part of f is greater than 0, this is your Eigen value corresponding to for the Eigen function e raised to $s t$. So, this is the Eigen value corresponding to the Eigen function e raised to $s t$ alright. We have constructed the impulses, we have derived the impulse response of this system the given LTI system and we also derived the Eigen value corresponding to the Eigen function e raised to $s t$.

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Let us look at another problem which is to determine the differential equation describing a system that is that is systems describe systems described by a differential equation for instance consider this system where I have let me draw a schematic of a system. This system is $x(t)$ and I have an error signal and this is an integrator and they and the output of the integrator is fed back and subtracted from the input to give the error signal and it is scaled by this is multiplied by a scaling constant this is a multiplier.

This is a schematic diagram of a system and let us try to determine now what we are asked to do we are asked to determine the differential equation corresponding differential equation representation of the system.

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Determine DE representation of system above.

$$x(t) - \alpha y(t) = e(t)$$

$$\int_0^t e(\tau) d\tau = y(t)$$

Differentiating both sides of above equation

$$e(t) = \frac{dy(t)}{dt}$$

We have to determine the differential equation the differential equation representation of the system above and you can see we have basically $x(t) - \alpha y(t) = e(t)$, we have that. Similarly, $e(t)$ is passed through an integrator to yield $y(t)$ which means $\int_0^t e(\tau) d\tau = y(t)$ now what we are going to do we are going to differentiate both sides of above equations. We are going to differentiating, differentiating both sides of above equation we have $e(t) = \frac{dy(t)}{dt}$ which means if I substitute this over here.

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$x(t) - \alpha y(t) = e(t)$
 $\int e(z) dz = y(t)$
Differentiating both sides of above equation
 $e(t) = \frac{dy(t)}{dt}$
 $x(t) - \alpha y(t) = \frac{dy(t)}{dt}$

Back over here that gives me x of t minus y of t equals $\frac{dy}{dt}$.

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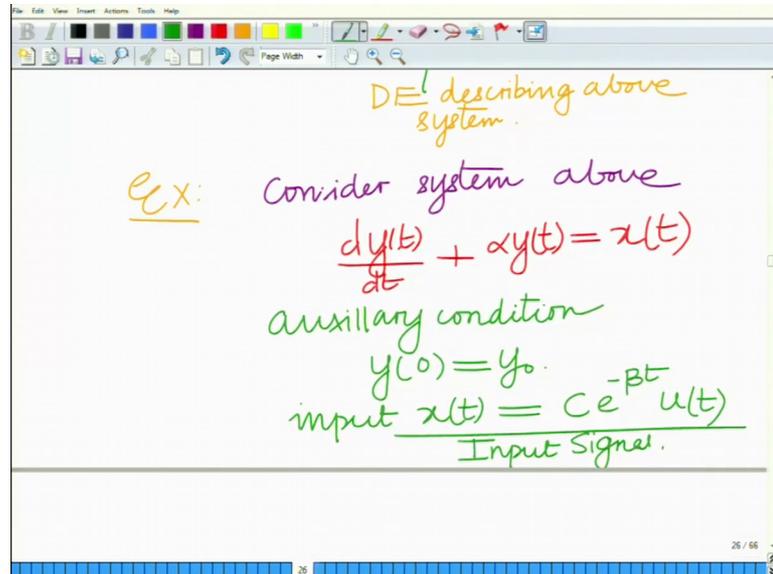
$\Rightarrow \frac{dy(t)}{dt} + \alpha y(t) = x(t)$
DE describing above system.
ex: Consider system above
 $\frac{dy(t)}{dt} + \alpha y(t) = x(t)$

Which basically implies that $\frac{dy}{dt} + \alpha y = x$, this is the differential equation? What we have is $\frac{dy}{dt} + \alpha y = x$ this is the differential equation describing the above system. This is the differential equation this is the differential equation describing above system.

Now, let us try to solve this to get the output response. Let us consider the as another example the same system that we have derived above consider system above which

remember is described as $\frac{dy}{dt} + \alpha y = x$ and with auxiliary condition remember to describe the system.

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The auxiliary condition auxiliary condition is $y(0) = y_0$ and corresponding to the input $x(t) = C e^{-\beta t} u(t)$.

This is basically your input signal this is basically your input signal where the auxiliary condition $y(0) = y_0$ and we have the system that is described by the differential equation $\frac{dy}{dt} + \alpha y = x$ and the way to approach this as we have already seen before is that the solution to any differential and their solution of the output $y(t)$ to any system described by this differential equation can be described as the sum of 2 components by $y_p(t)$ and $y_h(t)$ where $y_p(t)$ is a particular solution and $y_h(t)$ is the homogenous solution or $y_h(t)$ is the homogenous component.

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Solution: $y(t) = y_h(t) + y_p(t)$

Homogeneous Solution Particular Solution

For $y_p(t)$:

Let $y_p(t) = Ke^{-pt}$

$x(t) = Ce^{-pt}$

$\frac{dy_p(t)}{dt} + \alpha y_p(t) = x(t)$

$\Rightarrow -K\beta e^{-pt} + \alpha K e^{-pt} = C e^{-pt}$

$\Rightarrow \frac{K(\alpha - \beta)e^{-pt}}{K(\alpha - \beta)e^{-pt}} = \frac{C e^{-pt}}{e^{-pt}}$

$\Rightarrow K = \frac{C}{\alpha - \beta}$

$\Rightarrow y_p(t) = \frac{C}{\alpha - \beta} e^{-pt}$

Particular Solution

We have seen that $y(t)$ can be expressed as $y_h(t) + y_p(t)$ where this is the homogeneous solution and this is the particular solution and therefore, now for the particular solution now what we are going to do is to find the particular solution let we are going to start with the assumption that $y_p(t)$ has the structure same structure as the input signal that is $Ke^{-\beta t}$.

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$\Rightarrow K(\alpha - \beta)e^{-pt} = Ce^{-pt}$

$\Rightarrow K(\alpha - \beta) = C$

$\Rightarrow \boxed{K = \frac{C}{\alpha - \beta}}$

$\Rightarrow y_p(t) = \frac{C}{\alpha - \beta} e^{-pt}$

Particular Solution

Now, we have $\frac{dy_p(t)}{dt} + \alpha y_p(t) = x(t)$ this implies $\frac{d}{dt} \left(\frac{C}{\alpha - \beta} e^{-pt} \right) + \alpha \left(\frac{C}{\alpha - \beta} e^{-pt} \right) = Ce^{-pt}$ as you can see minus $K\beta e^{-\beta t}$ plus $\alpha K e^{-\beta t}$ this is

equal to $x(t)$ this implies that $K \alpha^{-\beta} e^{-\beta t} = x(t)$ this implies or this is equal to you know $x(t)$ is given as $K e^{-\beta t}$.

This is equal to $K e^{-\beta t}$, this implies this or I am sorry this is C . We are given $x(t) = C e^{-\beta t}$. we have $x(t)$ the input signal is $C e^{-\beta t}$ correct and therefore, we have $\alpha^{-\beta} = C$ this implies or $K \alpha^{-\beta} = C$ which in turn implies $K = C \alpha^{\beta}$.

We have the particular solution implies $y_p(t) = C \alpha^{\beta} e^{-\beta t}$. This gives us the particular this gives rise this gives us the particular solution which is $C \alpha^{\beta} e^{-\beta t}$ we have founded by assuming the particular solution $y_p(t)$ to have the same form as the input signal that is it is the form some constant K times $e^{-\beta t}$ and this constant K has been determined subsequently determined as $C \alpha^{\beta}$.

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The image shows a handwritten derivation for the homogeneous solution of a differential equation. The text is written in various colors (purple, green, blue) on a white background. The derivation starts with the assumption of a homogeneous solution $y_h(t) = \tilde{K} e^{st}$ and leads to the characteristic equation $s = -\alpha$.

$$\text{For Homogeneous Solution:}$$

$$\text{Let } y_h(t) = \tilde{K} e^{st}$$

$$\frac{dy_h(t)}{dt} + \alpha y_h(t) = 0$$

$$\tilde{K} s e^{st} + \alpha \tilde{K} e^{st} = 0$$

$$\Rightarrow \tilde{K} (s + \alpha) e^{st} = 0$$

$$\Rightarrow \boxed{s = -\alpha}$$

Let us now find the homogenous solution. Now, to find the homogenous solution to find the homogenous solution now let us assume $y_h(t) = \sum K \tilde{K} e^{st}$ now remember the homogenous solution can be found as by equating $\frac{dy_h(t)}{dt} + \alpha y_h(t) = 0$, now, differentiating $y_h(t)$, we have $K \tilde{K} s e^{st} + \alpha K \tilde{K} e^{st} = 0$ which implies $K \tilde{K} (s + \alpha) e^{st} = 0$ which implies which implies $K \tilde{K} (s + \alpha) e^{st} = 0$ which implies $s = -\alpha$.

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The image shows a whiteboard with handwritten mathematical work. At the top, the equation $\Rightarrow K(s + \alpha)e^{-st} = 0$ is written. Below it, the root $s = -\alpha$ is boxed. The homogeneous solution is given as $y_h(t) = \tilde{K}e^{-\alpha t}$, with a green arrow pointing to \tilde{K} and the text "Determine \tilde{K} ". The total solution is then expressed as $y(t) = y_h(t) + y_p(t) = \tilde{K}e^{-\alpha t} + \frac{C}{\alpha - \beta}e^{-\beta t}$. A green arrow points to \tilde{K} with the text "use auxiliary condition for \tilde{K} ". The whiteboard interface includes a toolbar at the top and a status bar at the bottom showing "29 / 66".

We have derived s equals minus alpha which implies $y_h(t)$ equals $\tilde{K}e^{-\alpha t}$, this is the homogeneous solution, but we have to remember still determine \tilde{K} . At this point we still have an unknown \tilde{K} and that can be done as follows we have $y(t)$ equals $y_h(t)$ plus $y_p(t)$ homogeneous solution plus particular solution. This is $\tilde{K}e^{-\alpha t}$ plus $\frac{C}{\alpha - \beta}e^{-\beta t}$. Now how do we determine the value of this unknown constant \tilde{K} and for that remember we use the auxiliary condition all right.

Determine this constant there is a single unknown \tilde{K} we have a single auxiliary condition that is the value of the signal output signal at time t equal to 0. I use the auxiliary condition to determine \tilde{K} . What we can do now is basically use auxiliary condition to determine \tilde{K} .

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use auxiliary condition for K

$$y(0) = y_0 \Rightarrow \tilde{K} + \frac{C}{\alpha - \beta} = y_0$$

$$\Rightarrow \tilde{K} = y_0 - \frac{C}{\alpha - \beta}$$

$$y(t) = \left(y_0 - \frac{C}{\alpha - \beta}\right)e^{-\alpha t} + \frac{C}{\alpha - \beta}e^{-\beta t}$$

We have $y(0)$ because $y(0)$ implies $\tilde{K} + \frac{C}{\alpha - \beta} = y_0$ implies $\tilde{K} = y_0 - \frac{C}{\alpha - \beta}$ implies $\tilde{K} = y_0 - \frac{C}{\alpha - \beta}$. Therefore, we have the solution $y(t) = \left(y_0 - \frac{C}{\alpha - \beta}\right)e^{-\alpha t} + \frac{C}{\alpha - \beta}e^{-\beta t}$. This determines the complete output signal.

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output signal for given system & auxiliary condition

For $t < 0$, $x(t) = 0$

$$\text{Let } y(t) = S e^{-\alpha t} \quad \frac{dy(t)}{dt} + \alpha y(t) = 0$$

$$\downarrow -\alpha S e^{-\alpha t}$$

Output signal for given system and also remember the auxiliary conditions are important in this case there is a single auxiliary condition that is a value of the output signal $y(0)$ at time $t = 0$ which is $y(0)$.

Now, for $t < 0$ since the input is 0. The only output response holds for $t < 0$ we can determine the output as follows for $t < 0$ we have $x(t) = 0$ which means $\frac{dy(t)}{dt} + \alpha y(t) = 0$ and the signal that satisfies this is again simply the homogeneous solution. This implies $y(t) = S e^{-\alpha t}$.

Substituting this in this differential equation we have $-\alpha S e^{-\alpha t} + \alpha S e^{-\alpha t} = 0$. This is now what satisfies this is basically we already know that what satisfies.

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The image shows a whiteboard with handwritten mathematical work. At the top, the equation $y(t) = S e^{-\alpha t}$ is written in green. An arrow points from the S to the text " $S = \text{constant}$ ". Below this, in orange, the initial condition $y(0) = y_0$ is written, followed by $\Rightarrow S = y_0$. A box contains the final result for $t < 0$: $y(t) = y_0 e^{-\alpha t}$. An arrow points from this box to the text "output signal for $t < 0$ ". The whiteboard interface includes a toolbar at the top and a status bar at the bottom showing "31 / 66".

This is the homogeneous solution $S e^{-\alpha t}$ for you know the form $S e^{-\alpha t}$ where S equals some constant and now that you determine that by using once again the auxiliary condition $y(0) = y_0$ implies $S = y_0$. For $t < 0$, we have the output $y(t) = y_0 e^{-\alpha t}$. This is the output signal; this is the output signal for at time $t < 0$ when the input is 0.

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Handwritten notes on a whiteboard:

- At the top, it says $\Rightarrow s = y_0$.
- Below that, it says "For $t < 0$, $y(t) = y_0 e^{-\alpha t}$ ". The equation is enclosed in a green box.
- An arrow points from the text "output signal for $t < 0$ " to the boxed equation.
- Below that, it says "Zero-input signal. i.e. for $x(t) = 0$ ".

The whiteboard interface shows a toolbar at the top and a page number "31 / 66" at the bottom.

You can think of this also as the 0 input signals this is basically your 0 input signals. You can think of this as basically this is the output of the system when the input x equal to 0.

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Handwritten notes on a whiteboard:

- At the top, it says "signal".
- The main equation is
$$y(t) = \left(y_0 - \frac{c}{\alpha - \beta}\right) e^{-\alpha t} + \frac{c}{\alpha - \beta} e^{-\beta t}$$

$$= \underbrace{y_0 e^{-\alpha t}}_{y_{zi}(t)} + \underbrace{\frac{c}{\alpha - \beta} (e^{-\beta t} - e^{-\alpha t})}_{y_{zs}(t)}$$

The whiteboard interface shows a toolbar at the top and a page number "32 / 66" at the bottom.

Now, if you look at the output for t greater than 0 that can be expressed as y naught e power minus α t plus well, let me write it down I have that y naught minus C alpha minus β C over α minus β e raised to minus α t plus 0 α minus β e raised to minus α β minus e raised to minus β t for t greater than 0 which can be written as y naught e raised to minus α t plus C over α minus β C over

alpha minus beta times C raised to minus beta t minus e raised to minus alpha t and you can now see that this is basically the y naught e power minus alpha t is the 0 input signal that is output to when the input is 0 and this is basically known as the 0 state signal.

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$$= y_0 e^{-\alpha t} + \frac{c}{\alpha - \beta} (e^{-\beta t} - e^{-\alpha t})$$

$y_{zi}(t)$ $y_{zs}(t)$
 output to zero input
 Response to auxiliary conditions
 Response with zero auxiliary condition
 zero state signal.

I can express the solution as $y_{zi}(t)$ which is input which is output to 0 input and this is basically 0 state signal which is the z which is basically the response. This is the output to 0 inputs or basically response to auxiliary conditions and these y_0 states is basically the response with response with 0 auxiliary conditions.

Basically you can see you can express the solution $y(t)$ also as $y_{zi}(t)$ which is basically the output to a 0, input that is output signal when the input is 0 plus $y_{zs}(t)$ right where z denotes 0 state. The 0 state signal 0 state output is basically output to 0 auxiliary condition and you can see here that if y_0 is 0 the auxiliary condition is 0 y_0 that is the output at time 0 that is $y(0)$ is 0 then $y_{zs}(0)$ is 0 $y_{zi}(t)$ is 0, it reduces to the 0 state signal that is the output to 0 auxiliary condition.

Similarly, if the input is 0 then we already seen the output is basically $y_{zi}(t)$ which is basically the output simply to the thus simply the output to the 0 input that is response simply to auxiliary condition y at time equal to 0 is $y(0)$. Basically that completes this thing. I can express this as the 0 input $z_i(t)$ which is the 0 input signal plus $y_{zs}(t)$ which is the 0 state.

This z_s stands for 0 state signals, 0 state this is basically your 0 state signals. We will stop this module here. We have looked at a couple of interesting example problems all right, pertaining to basically the Eigen function of a LTI system, the differential equation representation of a given system and also the solution or the output signal to a system represented by represented by a differential equation given the input signal and also the auxiliary controls all right. We will stop here and continue with other aspects in subsequent modules.

Thank you very much.