

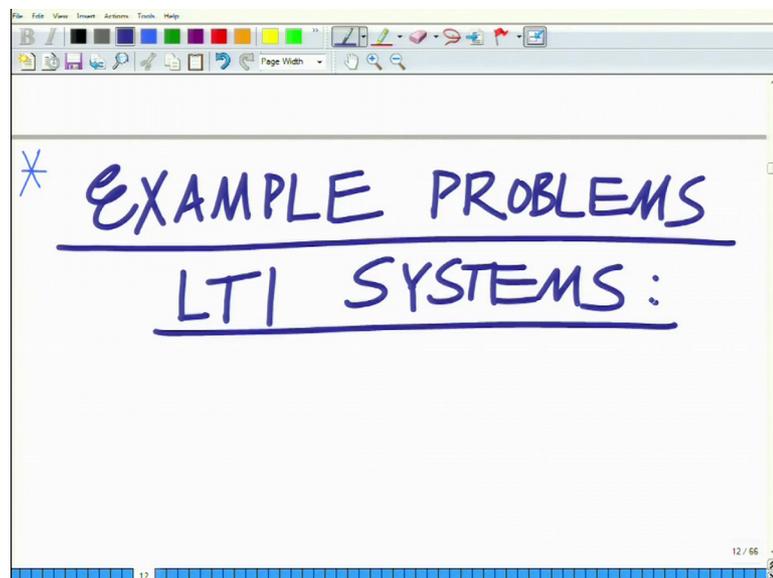
Principles of Signals and Systems
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Lecture - 14

Example Problems LTI Systems-Convolution, Periodic Convolution, BIBO Stability

Hello welcome to another module in this massive open online course. So, we are looking at example problems in the analysis of continuous time and discrete time LTI systems right linear time invariant systems let us continue looking at these examples.

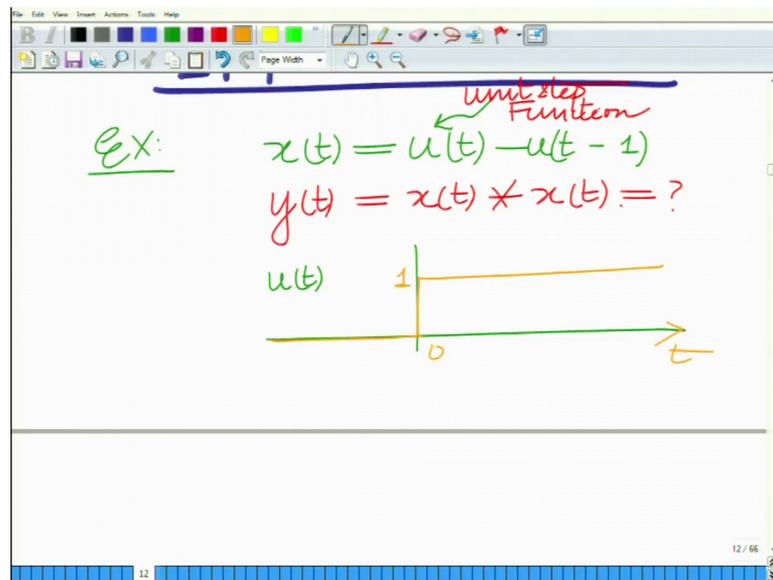
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So, we are looking at example problems for the analysis of with the analysis of LTI systems.

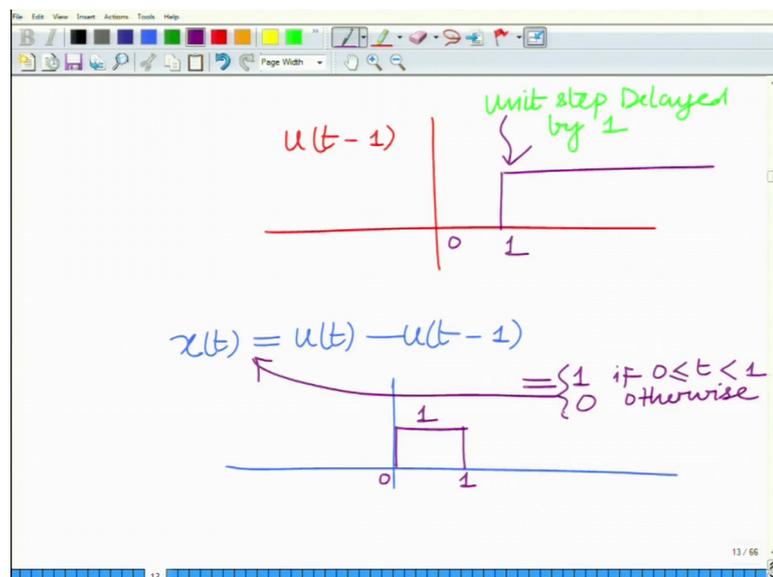
And let us start looking at another problem which is basically we are given that $x(t)$ equals $u(t) - u(t - 1)$.

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Where $u(t)$ remember, this is the unit step function. $u(t)$ is the unit step function. We want to find out what is $y(t)$ equal to $x(t)$ convolved with $x(t)$. What this? First let us try plotting $u(t)$ $x(t)$ now we have we know $u(t)$ is the unit step function. This is given by 1 if t is greater than or equal to 0 and is 0 otherwise this is the unit step function, which is equal to 1 if t is greater than equal to 0 and 0 otherwise.

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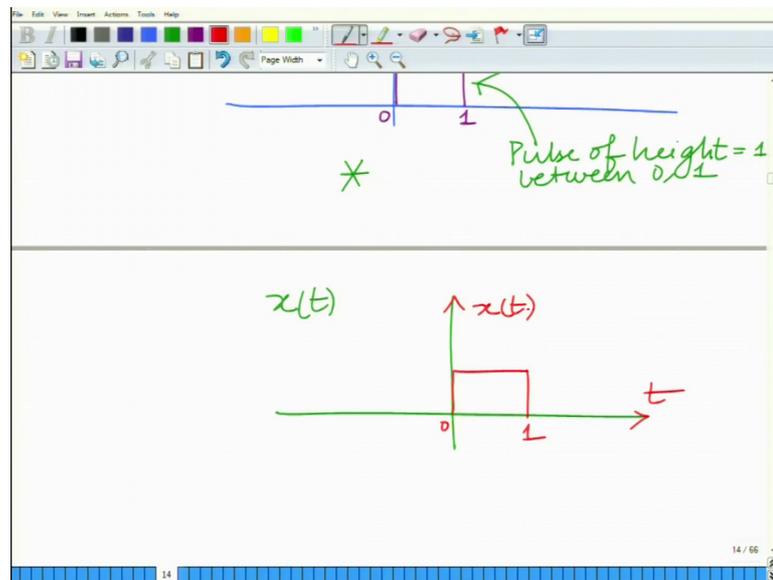


Now, $u(t-1)$ is the unit step function delayed by 1. So, this will be take the unit step function. And delay it by 1 unit. So, this is unit step delayed by unit step delayed by 1.

And therefore, $x(t)$ you can see which is equal to $u(t) - u(t - 1)$. This will be given as this will be given by the function, which is equal to 1 when $0 \leq t < 1$ and 0 otherwise. So, this is $x(t)$ which is $u(t) - u(t - 1) = 1$ if $0 \leq t < 1$, and this is 0 otherwise. So, this is your $x(t)$.

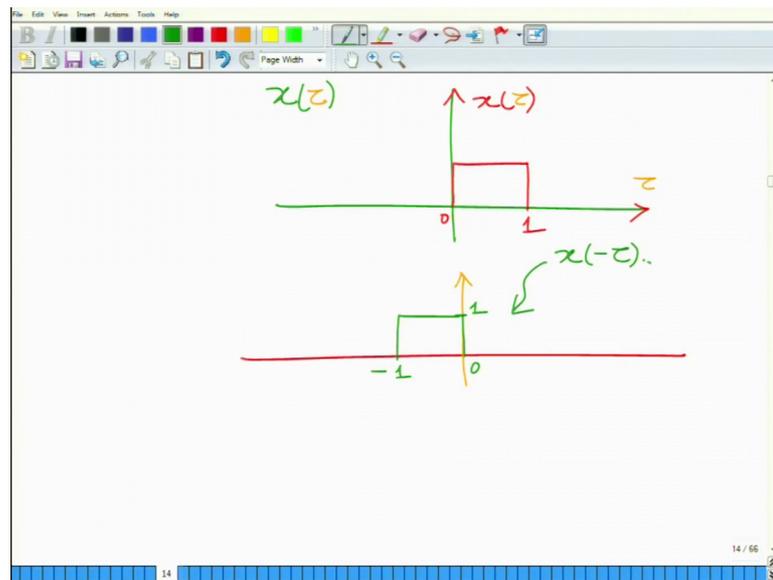
Basically, it is a pulse of height one between 0 and 1, you can clearly see this is a pulse of height equal to 1 between 0 and 1 that is from $t = 0$ to $t = 1$.

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And now we want to convolve $x(t)$ with itself convolve $x(t)$ with another $x(t)$ which is again; obviously, this is going to be a pulse of height one between 0 and between 0 and 1.

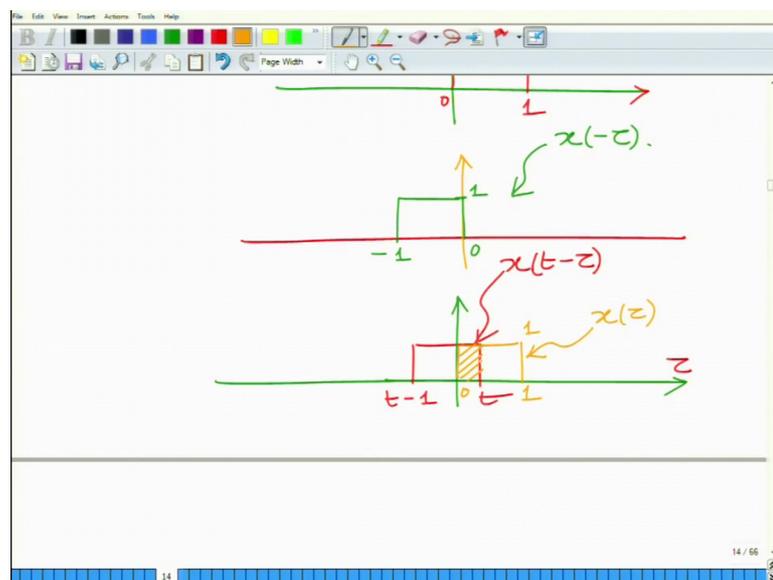
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Now to convolve remember we need to flip first we are convolving $x(t)$ with $x(t)$. So, we need to first flip it about 0.

So, let us say this we make it as the tau axis rather than the t axis. So, we make this as x of tau and this becomes your x of tau. So, x of minus tau will be flip about the y axis. So, this will be pulse of height one this will be pulse of height one, and between minus 1 and 0. So, this is your x of minus tau.

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And x of t minus τ will be delayed version of this by t . So, x of t minus τ will be shifted to the right by t as you see we have. So, this is t this is t minus 1 and this is x of t minus τ . So, this is your τ axis, this is x of t minus τ . And now multiply this with x of τ remember x of τ is still a pulse of height one between 0 and 1.

So, this is a pulse of height one between 0 and 1, and this is your x of τ . So, multiply x of t minus τ by x of τ integrate from 0 to infinity. So, this is basically the area of overlap between these 2 pulses which can be seen to be; so, you can see integral minus infinity to infinity x of τ x of t minus τ $d\tau$.

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$$\int x(\tau) \cdot x(t-\tau) d\tau$$

$$\rightarrow = |x \cap t| = t$$

increases linearly with t
For $0 \leq t < 1$

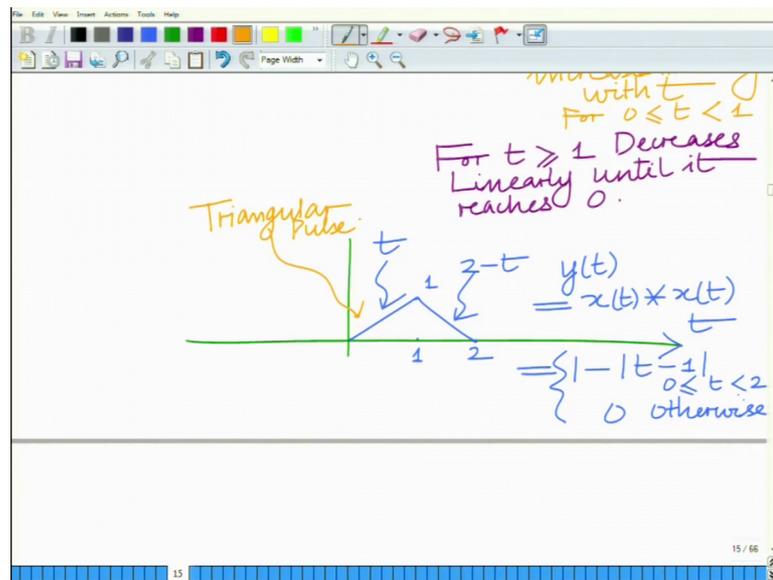
For $t \geq 1$ Decreases linearly until it reaches 0.

Remember for x of t minus τ we flip about the y axis; that is, look at x of minus τ and then delay it by t shifted to the right by t . And now you can see the overlap corresponding overlap is of t height 1.

So, this will be equal to 1 into t equals t . And therefore, this increases linearly with t increases linearly with t for $0 \leq t < 1$. Now once t greater than 1, from t greater than 1 it decreases again if you can see right for t greater than 1. You can see it again decreases linearly greater than 1 or greater than equal to 1 decreases linearly because overlap progressively decreases until it reaches 0.

So, which means this will look something like this will look something like from 0 to one it increases.

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So, this is at one from one it decreases until it hits 2. So, this is t this height corresponds to the maximum overlap which is 1 and this is therefore, this portion will be 2 minus t . So, this is your $y(t) = x(t) * x(t)$. And you can see this can be described as this can be described as 1 minus modulus of t minus 1 or for $0 \leq t < 2$, and this is 0 and this is 0 otherwise.

So, this is basically your triangular waveform or not a wave form a triangular pulse basically. So, the convolution of 2 square pulses of equal width character the convolution of 2 square pulses of equal width, and height there is convolution of square pulse with itself yields the triangular pulse. This is 1 minus modulus of t minus 1 for $0 \leq t < 2$, and this is 0 otherwise. So, this is basically the result of the convolution.

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A whiteboard screenshot showing a handwritten solution. The function is defined as $y(t) = \begin{cases} 1 - |t - 1| & 0 \leq t < 2 \\ 0 & \text{otherwise} \end{cases}$. The word "Solution" is written in red below the function, with an arrow pointing to it. Below that, "EX: Periodic Convolution" is written in green.

Let me just write it over here once again this is 1 minus magnitude t minus 1 0 less than equal to t less than 2 and this is 0 otherwise. So, this is the solution. This is the final solution.

Let us now look at the next problem which is that of a periodic convolution between 2 periodic signals. Periodic convolution is defined thus let x_1, x_2 be periodic with a common period T naught let there be let x_1 to x_2 b 2.

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A whiteboard screenshot showing the definition of periodic convolution. It starts with "EX: Periodic Convolution" in green. Below it, "Let $x_1(t), x_2(t) =$ Periodic common Period = T_0 " is written in blue. The convolution integral is given as $F(t) = \int_0^{T_0} x_1(\tau) x_2(t - \tau) d\tau$. Below the integral, it is equated to $x_1(t) \otimes x_2(t)$, with the word "Periodic Convolution" written in yellow below the symbol.

2 signals which are periodic with common period T naught capital T naught the periodic convolution f of t is defined as 0 to T naught x 1 tau that is convolution over one period x 1 tau x 2 t minus tau d tau and it is represented as x 1 t with this kind of a symbol.

And this is termed as a periodic convolution.

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The image shows a whiteboard with handwritten mathematical notes. At the top, it says "Periodic convolution" in yellow, followed by an equation: $x_1(t) \otimes x_2(t)$. Below this, it says "Show:" and then an integral equation: $\tilde{f}(t) = \int_{t_0}^{t_0+T_0} x_1(\tau) x_2(t-\tau) d\tau$. An arrow points from the lower limit t_0 to the text "= f(t)". Below that, it says "Shifting integral by t_0 ." At the bottom, it says "Solution: First, observe,".

And it is desired to show that this periodic convolution remains unchanged if it if you shift it by small T naught. So, show that f of t if you shift it if you shift the integral by small T naught that is x 1 t x 2 that is if you denote this by f tilde at t x 2 t minus tau d tau, we need to show that this is equal to f t ; that is, you are shifting the integral by T naught. Shifting integral by small T naught any arbitrary value small T naught. Now the solution can proceed as follows.

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= $f(z)$
Shifting integral by t_0 .

Solution: Let $\phi_t(z) = x_1(z) x_2(t-z)$
for given t

$$\phi_t(z+T_0) = x_1(z+T_0) \times x_2(t-(z+T_0))$$

$$= x_1(z+T_0) \cdot x_2(t-z-T_0)$$

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First observe that now, let us set phi of t now let phi of tau for any given value of t equals x_1 of tau \times x_2 of t minus tau; that is, phi of tau for now let us subscript this by t to show the dependence also on t for given for given t. Now we will show that phi of t of tau, now let us look at this phi of t of tau plus t equals x_1 of tau plus T naught in to x_2 of t minus tau plus T naught, which is equal to x_1 of tau plus T naught into x_2 of t minus tau minus T naught.

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$$= x_1(z+T_0) \cdot x_2(t-z-T_0)$$

$$= x_1(z) \cdot x_2(t-z)$$

$$\phi_t(z+T_0) = \phi_t(z)$$

$\phi_t(z) = \text{Periodic}$

$$\int_0^{T_0} \phi_t(z) dz = \int_{t_0}^{t_0+T_0} \phi_t(z) dz$$

$$= \int_{t_0}^{t_0+T_0} x_1(z) x_2(t-z) dz$$

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Now, look at this x_1 of τ plus T naught that is shifting by period this is equal to x_1 tau times x_2 of t minus τ minus T naught is also equal is again is equal to x_2 of t minus τ , which means this is equal to ϕ t of τ . So, what we have shown is ϕ t of t plus τ ϕ t of t plus T naught is equal to ϕ t of τ . So, we have shown that ϕ sub t of ϕ sub t I am sorry, ϕ sub t of τ plus T naught ϕ sub t of τ plus T naught equals ϕ sub t after which means ϕ sub t of τ is periodic. Which means this implies ϕ sub t of τ this is periodic, which implies that basically integral over any period is same.

So, you have integral 0 to T naught ϕ sub t of τ d t is integral, you can shift it by any constant quantity small T naught. So, integral this is equal to ϕ . So, t of τ d t which is equal to T naught to T naught plus τ naught to a small T naught plus capital T naught x of τ x 2 of t minus τ d τ which is equal to f tilde of f tilde of t .

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The image shows a whiteboard with the following handwritten content:

$$\phi_T(z) = \text{Periodic}$$

$$F(t) = \int_0^{T_0} \phi_T(z) dz = \int_{t_0}^{t_0+T_0} \phi_T(z) dz$$

$$= \int_{t_0}^{t_0+T_0} x_1(z) x_2(t-z) dz$$

$$= \tilde{F}(t)$$

For a periodic function integral over any period = same.

And this is f t and this follows because for a periodic function, for a periodic function integral over any period is the same that is if I have a periodic function with period T naught. Now if I consider any duration right any time interval of duration capital T naught, then the integral will be same therefore, integral 0 to capital T naught will be the same as integral any small T naught to small T naught plus capital T naught.

So, integral over any interval of duration capital T naught where T naught is the capital T naught is the period will be the same for a periodic function. So, we are using that property. So, basically shows that which shows that the periodic convolution also

remains unchanged when you consider any time interval of duration capital T naught, that is the basic idea behind this a periodic convolution, which is defined for the convolution of 2 periodic functions with a common period capital T.

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For a periodic function
integral over any period = same.

Ex: Consider system with impulse
response
$$h(t) = e^{-t^2/2} \quad -\infty < t < \infty$$

BIBO stable?

Solution: For BIBO stability
$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

Let us look at the next example, which is consider the system with the impulse response h of t equals e to the power of minus t square by 2 minus infinity less than t less than infinity. So, we want to ask the question is this BIBO stable, is this system BIBO stable and the solution can proceed as follows. For BIBO stability remember, the condition for BIBO stability we have to have minus infinity to infinity must be finite that is the impulse response must be absolutely integrable; that is, integral minus infinity to infinity magnitude h t d t should be a finite quantity.

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response

$$h(t) = e^{-t^2/2} \quad -\infty < t < \infty$$

BIBO stable?

Solution: For BIBO stability

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

"Bell Shaped Function" Finite Gaussian Pulse $e^{-t^2/2}$

So, this quantity should be finite. Now if you look at e to the power of minus t square by 2, you will realize that it is given basically by a Gaussian pulse. Which is this is a standard signal, this is known as the Gaussian pulse. Due to the power of minus t , it has a bell shaped also known as a bell-shaped function. This is also known as a bell-shaped looks like a bell. Therefore, also known as a bell shape function. Now this is always positive e to the power of minus t square by 2.

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$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} e^{-t^2/2} dt$$
$$= \sqrt{2\pi} \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-t^2/2} dt$$

Gaussian Prob Density Function $\frac{1}{\sqrt{2\pi}} e^{-t^2/2}$

var $\sigma^2 = 1$

So, therefore, integral minus infinity to infinity to check BIBO stability integral minus infinity to infinity equals integral minus infinity to infinity magnitude, but this is always positive. So, this is simply h of t . Now this is also e to the power of minus t square by 2 is also you can see, now what I am going to do is I am going to multiply by square root of 2π and this is integral minus to infinity to infinity, 1 over square root of 2π , this is 1 over square root of 2π e to the power of minus t square by 2.

Now, you can show that this integral is equal to 1, because this integral is minus infinity to infinity of a Gaussian probability density function. So, each 1 over square root of 2π e to the power of minus t square by 2 this you can show is a this you can show is a Gaussian probability. This is a Gaussian probability density function that is 1 over square root of 2π σ square e power minus t square by 2 σ square is a Gaussian probability density function with mean μ , and variance equal to σ square. And integral minus infinity to infinity Gaussian probability density function is 1. Here you can see the variance is σ square is equal to here you can see the variance is σ square equal to 1.

If you simply substitute σ square equal to 1, all right. Do you get the Gaussian probability density function the σ square equal to 1 and mean is 0. So, that gives the Gaussian probability density function 1 over square root of 2π e raise to minus t square by 2 all right. So, that is the Gaussian probability density function for the range t lying between minus infinity to infinity. So, this integral is one implies the net integral is square root of 2π which is a finite quantity.

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$$= \sqrt{2\pi} \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}} e^{-\frac{t^2}{2\sigma^2}} dt$$

Gaussian Probability Density Function
 $\text{var } \sigma^2 = 1$
 $\frac{1}{\sqrt{2\pi\sigma^2}} e^{-\frac{t^2}{2\sigma^2}}$
 \uparrow
 $\text{var} = \sigma^2$

$$= \sqrt{2\pi} < \infty$$

\Rightarrow System is indeed BIBO Stable.

So, this is a finite quantity. Hence, the system implies a system is indeed the system with impulse response given by the Gaussian pulse is indeed BIBO stable. Because it is the integral absolute it is that it is absolutely integrable there is integral minus infinity to infinity magnitude $h(t)$ is a finite quantity, all right. Let us now proceed to the next example. Which is given by the consider the LTI system given by consider the LTI system where the output $y(t)$ equals minus infinity to infinity $e^{j\omega(t-\tau)} x(\tau) d\tau$, this is the output.

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Ex: Consider LTI system

$$y(t) = \int_{-\infty}^{\infty} e^{-j\omega(t-\tau)} x(\tau) d\tau$$

output \uparrow input \uparrow

Find $h(t)$, Eigenvalue for function $e^{j\omega t} u(t)$.

impulse response

$$= \int_{-\infty}^{\infty} e^{-j\omega(t-\tau)} x(\tau) d\tau$$

$$= \int_{-\infty}^{\infty} \frac{h(t-\tau)}{x(\tau)} x(\tau) d\tau$$

So, for input x of t , this output y of t is given by this relation. What we need to do is we need to find the impulse response. And also, Eigen value for the function e to the power of $s t u t$. Or let us put it this way, this is let us make it e to the power of $\text{minus } t \text{ minus } e$ to the power of $\text{minus } t \text{ minus } \tau$ times. Or let us keep it this way e to the power of $\text{minus } j \text{ omega naught } t \text{ minus } \tau$ $x \tau d \tau$. And find the Eigen value corresponding to Eigen value for function e to the power of $s t u t$.

Now, it is easy to see what the impulse response is; the impulse response of this is equals well you can write this as a this is integral $\text{minus } \infty$ to ∞ $\text{minus } j \text{ omega } t \text{ minus } \tau$ $x \tau d \tau$. So, this is your h of $t \text{ minus } \tau$, and this is your x of τ . So, this is basically $h t$ convolved with $x t$ impala implies that the impulse responses; implies, that the impulse responses e power $\text{minus } j \text{ omega naught } t$.

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The image shows a whiteboard with handwritten mathematical derivations. At the top, it says "Find $h(t)$, Eigenvalue for function $e^{st}u(t)$ ". A blue arrow points from "Find $h(t)$ " to the first equation. A red arrow points from "Eigenvalue for function $e^{st}u(t)$ " to the second equation. The equations are:

$$\begin{aligned} & \text{impulse response} \\ &= \int_{-\infty}^{\infty} \frac{e^{-j\omega_0(t-\tau)} x(\tau) d\tau}{h(t-\tau) x(\tau)} \\ &= h(t) * x(t) \\ &\Rightarrow \text{impulse response } h(t) = e^{-j\omega_0 t} \end{aligned}$$

The whiteboard also has a toolbar at the top and a status bar at the bottom showing "21 / 66".

So, the impulse response of this system is e power $\text{minus } j \text{ omega naught}$. So, all right so, we will stop this module here, we are looking at some examples of the analysis of the LTI systems. So, we will stop this module here and continue with other similar examples in the subsequent modules.

Thank you very much.