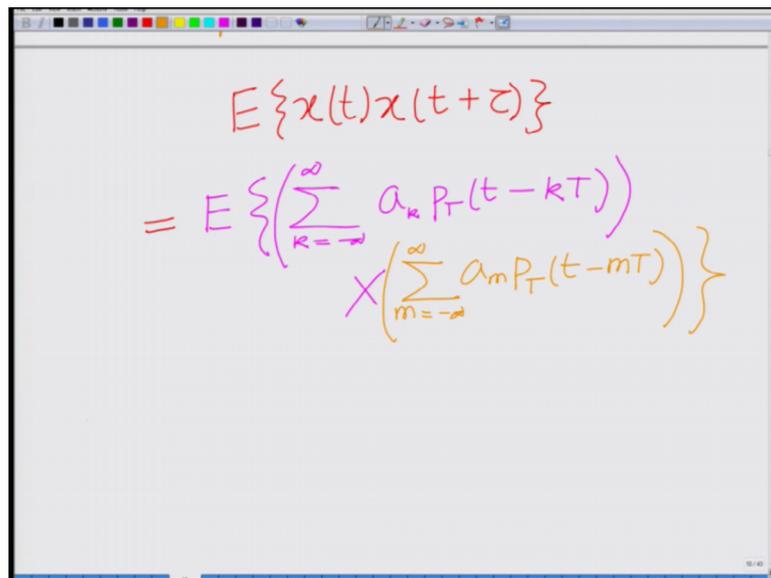


Principles of Communication Systems - Part II
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Lecture – 03
Spectrum of Transmitted Digital Communication Signal, Autocorrelation Function
and Power Spectral Density

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$$E\{x(t)x(t+\tau)\}$$
$$= E\left\{\left(\sum_{k=-\infty}^{\infty} a_k P_T(t-kT)\right) \times \left(\sum_{m=-\infty}^{\infty} a_m P_T(t-mT)\right)\right\}$$

Hello, welcome to another module in this massive open online course. So, we are looking at the autocorrelation function of the transmitted digital communication signal that is expected value of $x(t)$ into $x(t + \tau)$. So, we want to compute the autocorrelation function expected value of $x(t)$ into $x(t + \tau)$. Now, substituting the value of $x(t)$, I have expected value of well $x(t)$ is nothing but summation k equal to minus infinity to infinity $a_k P_T(t - kT)$ times well $x(t + \tau)$. So, summation m equals minus infinity to infinity $a_m P_T(t - mT)$ close the brackets.

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$$= E \left\{ \left(\sum_{k=-\infty}^{\infty} a_k P_T(t - kT) \right) \times \left(\sum_{m=-\infty}^{\infty} a_m P_T(t + \tau - mT) \right) \right\}$$

$$= E \left\{ \sum_{k=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} a_k a_m P_T(t - kT) P_T(t + \tau - mT) \right\}$$

Now, we expand the summation that is expected value of well summation k equal to minus infinity to infinity summation m equal to minus infinity to infinity a k a m P well of course, this is x t plus tau, so this has to be t plus tau minus m t. So, this is your x t this whole thing, and this is your x t plus also P T of t minus k T into P T of t plus tau minus m T.

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$$= E \left\{ \sum_{k=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} a_k a_m P_T(t - kT) P_T(t + \tau - mT) \right\}$$

Take expectation operator inside

$$= \sum_{k=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} E \left\{ a_k a_m \right\} P_T(t - kT) P_T(t + \tau - mT)$$

Now, if I take the expectation operator inside, so if I take the expectation operator inside the summation, we have taking the expectation operator inside, we have well summation

k equal to minus infinity to infinity summation m equal to minus infinity to infinity expected value of a k into a m times P T minus k T times P T t plus tau minus m T. Now, if you can look at this quantity expected value of a k into a m. Remember we already said the symbols are independent identically distributed.

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$$= \sum_{k=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} E\{a_k a_m\} P_T(t - kT) \times P_T(t + \tau - mT)$$

$$E\{a_k a_m\} = \begin{cases} E\{a_k\} E\{a_m\} & \text{if } k \neq m \\ E\{a_k^2\} & \text{if } k = m \end{cases}$$

$= 0$
since symbols are zero mean IID.

Therefore expected value of a k a m, if k is not equal to m, we have seen this expected value of a k a m; if k not equal to m is expected value because symbols are independent this is expected value of a k into expected value of a m this is equal to 0. So, this is equal to 0. Since symbols are the digital symbols are IID 0 mean, since symbols are 0 mean IID, therefore, expected value of a k a m 0 but if only if k is not equal to; if k is equal to m then it becomes expected value of a k square and that is not equal to 0. So, only those terms will survive in this summation in which k is equal to m; all the other terms will vanish because expected value of a k times a m is 0.

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The whiteboard shows the following derivation:

$$= \sum_{k=-\infty}^{\infty} \underbrace{E\{a_k^2\}}_{A^2} P(t - kT) P(t + \tau - kT)$$

$$P_r(a_k = A) = \frac{1}{2}$$

$$P_r(a_k = -A) = \frac{1}{2}$$

$$E\{a_k^2\} = \frac{1}{2} \cdot A^2 + \frac{1}{2} \cdot (-A)^2$$

$$= A^2$$

So, therefore, the summation equivalently reduces to expected value or summation k equal to, so there is only now k equal to minus infinity to infinity because only terms corresponding to k equal to m will survive. This will become expected value of a k square right into P well t minus k T into P t plus τ minus k T . Now if you can look at this quantity expected value of a k square that can be simplified as follows, we know that probability a k equal to A is half; we have assumed symbols are equiprobable probability a k equals minus A is also half. So, expected value of a k square equals half into A square plus half into minus A square both of these are A square minus A square is also A square, so half into a square plus half into a square, which is simply A square.

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$R = -\infty$
 $A^2 = P_d$ — Data Symbol Power
 $Pr(a_k = A) = \frac{1}{2}$
 $Pr(a_k = -A) = \frac{1}{2}$
 $E\{a_k^2\} = \frac{1}{2} \cdot A^2 + \frac{1}{2} \cdot (-A)^2$
 $= A^2$

So, this quantity expected value of a k square is simply A square or we can also call this as P d which is the power of the data symbols this is P d, this is simply your data symbol, this is simply the power of the data symbols.

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$= P_d \sum_{R=-\infty}^{\infty} P_T(t - kT) P_T(t + \tau - kT)$
 Depends on t
 $E\{x(t)x(t+\tau)\}$
 Hence NOT WSS
 Wide Sense Stationary

Therefore, this reduces to P d times summation k equal to minus infinity to infinity P T t minus k T P T t plus tau minus k T. Now, if you can see this quantity here, this is our autocorrelation function. Now, if you can this see this quantity here, this quantity here depends on t. When we have expected value of x t into x t plus tau, we know that when

expected value of $x(t)$ into $x(t + \tau)$ depends not only on τ the time shift out, but also depends on t then the resulting random process is not wide sense stationary. Therefore what we can see here is that this digital communication signal is not wide stationary. Since, it depends on t hence it is not wide sense stationary that is basically wide sense. We would like to therefore convert this right because remember we wanted white sense stationary signals are very helpful to us because they are suitable to be studied, easy to be analyzed and characterized. And more importantly we can evaluate their power spectral density to characterize the spectral distribution of the power. So, we want to convert this signal which is not wide sense stationary this digital communication signal, it wide sense stationary signal.

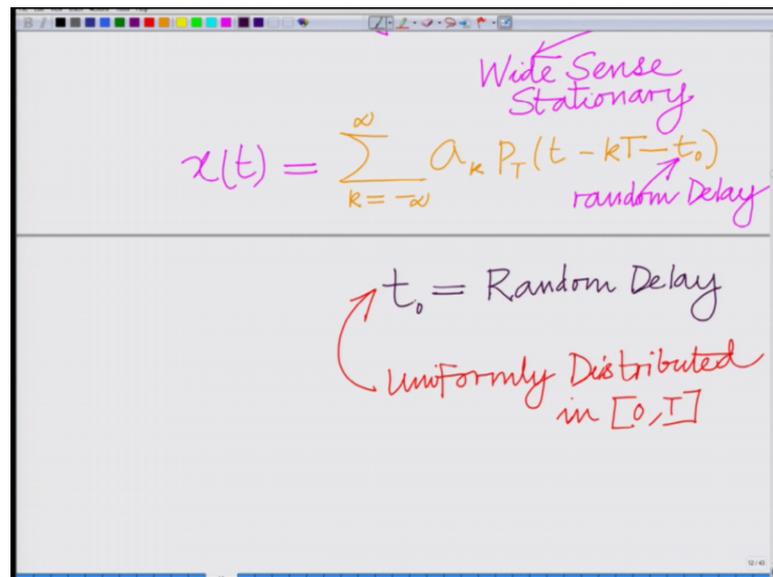
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$$x(t) = \sum_{k=-\infty}^{\infty} a_k P_T(t - kT - \tau_0)$$

Hence NOT WSS
Wide Sense Stationary
random Delay
 $\tau_0 = \text{Random Delay}$

Therefore, what I am going to do is I am going to do the following is I am going to take my signal $x(t)$ and I am going to move like a slight modification k equal to minus infinity to infinity $a_k P_T(t - kT - \tau_0)$ that is same as before. But I am also going to introduce this τ_0 which is a random delay. So, this quantity τ_0 is a random delay of the signal and this is uniformly distributed.

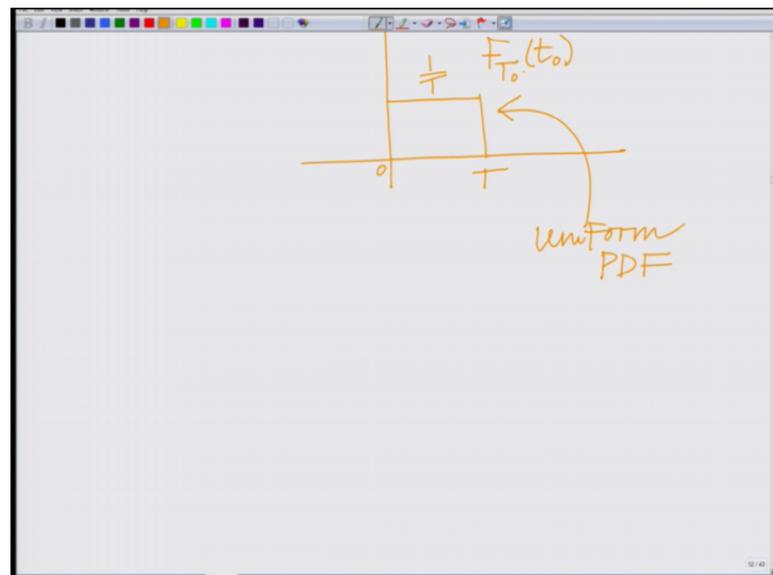
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The whiteboard shows the equation $x(t) = \sum_{k=-\infty}^{\infty} a_k P_T(t - kT - t_0)$. Above the equation, it says "Wide Sense Stationary" with an arrow pointing to the equation. Below the equation, it says "random Delay" with an arrow pointing to t_0 . Below the equation, it says " $t_0 =$ Random Delay" and "Uniformly Distributed in $[0, T]$ " with an arrow pointing to t_0 .

We are going to assume this is uniformly distributed in 0 to that is basically a uniform distribution. Let us so this is t_0 sorry not τ_0 t_0 and it is uniformly distributed in 0 to T and you can see a uniform distribution in 0 to T is given as follows it is of height 1 over T .

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So, this is T , the distribution is 1 over T in 0 to T . So, this is your F of T naught of t_0 naught, this is the uniform probability density function of T naught, this is your uniform probability density function. So, now what we are assuming is that this digital

communication signal has a random delay T naught, which is basically distributed in uniformly in 0 to capital T , where capital T is the symbol time, 1 over T is the 1 or capital T is the symbol rate.

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$$E\{x(t)x(t+\tau)\}$$

$$= P_d \sum_{k=-\infty}^{\infty} E\{P_T(t-kT-t_0) \times P_T(t+\tau-kT-t_0)\}$$

Average or Expected value wrto t_0 :

So, now we have again let us go back to our expected value of x t into x t plus τ , this is going to be well the same thing. Let us start with P d summation k equal to minus infinity to infinity P T t minus k T , now, in addition there is this delay t naught times P T t plus τ minus k T minus t naught. Now, I have to take the expected value of this because t naught is random, I have to take the expected value average with respect to average or expected value with respect to average or the expected value with respect to t naught.

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$$= P_d \sum_{k=-\infty}^{\infty} \int_{-\infty}^{\infty} F_{T_0}(t_0) P_T(t - kT - t_0) \times P_T(t + \tau - kT - t_0) \times dt_0.$$

Averaging wrto t_0
random Delay

And to take the average with respect to t naught, all I am going to do is I am going to multiply by the probability of density function that is well. Let me write the complete expression I have to multiply with the probability density function of t naught that is $P_T(t - kT - t_0)$ into $P_T(t + \tau - kT - t_0)$ into dt_0 . So, multiplied by the probability density functions and integrates, so this is basically averaging with respect to which is basically your random delay, this is your random delay.

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$$F_{T_0}(t_0) = \begin{cases} \frac{1}{T} & 0 < t_0 \leq T \\ 0 & \text{otherwise} \end{cases}$$

Delay

$$= \frac{P_d}{T} \sum_{k=-\infty}^{\infty} \int_0^T P_T(t - kT - t_0) \times P_T(t + \tau - kT - t_0) \times dt_0.$$

Now, $F(t)$ is the uniform probability density function, so $F(t)$ equals $\frac{1}{T}$ for $0 \leq t < T$, it is 0 outside that interval. Therefore, I can write this thing as follows therefore, I can write expected value of $x(t)$ into $x(t) + \tau$ is summation k equal to minus infinity to infinity integral 0 to T well $\frac{1}{T}$ over T , $F(t)$ is $\frac{1}{T}$ or a constant. So, I can bring that outside $\frac{1}{T}$ $P(t - kT - t_0)$ times $P(t + \tau - kT - t_0)$ into dt_0 equals $\frac{P_d}{T}$ summation k equal to minus infinity to infinity.

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The whiteboard shows the following derivation:

$$= \frac{P_d}{T} \sum_{k=-\infty}^{\infty} \int_0^T P_T(t - kT - t_0) \times P_T(t + \tau - kT - t_0) dt_0$$

$t_0 + kT = \tilde{t}_0 \quad dt_0 = d\tilde{t}_0$

$$= \frac{P_d}{T} \sum_{k=-\infty}^{\infty} \int_{kT}^{(k+1)T} p(t - \tilde{t}_0) p(t + \tau - \tilde{t}_0) d\tilde{t}_0$$

Now, what I am going to do is I am going to employ a substitution here that is let us employ the substitution $t_0 + kT = \tilde{t}_0$. So, we have $dt_0 = d\tilde{t}_0$ and this becomes well $\frac{P_d}{T}$ times summation k equal to minus infinity to infinity well integral limits become well $t_0 + kT$ equal to \tilde{t}_0 . So, limits become kT to $(k+1)T$ $t_0 + kT$ is \tilde{t}_0 , so this is $P(t - \tilde{t}_0)$ into $P(t + \tau - \tilde{t}_0)$ into $d\tilde{t}_0$.

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$$= \frac{P_d}{T} \sum_{k=-\infty}^{\infty} \int_{kT}^{(k+1)T} p(t - \tilde{t}_0) p(t + z - \tilde{t}_0) d\tilde{t}_0$$

each integral from kT to $(k+1)T$
 k from $-\infty$ to ∞

$t_0 + kT = \tilde{t}_0$
 $d\tilde{t}_0 = \dots$

Now, if you can look at this integral each integral is from kT to $k + 1T$ each integral is from kT to $k + 1T$ k is from infinity minus infinity to infinity. Therefore, the entire this can be also. So, we have a sum of several integrals each integral is going from kT to $k + 1T$ and k itself is going from minus infinity to infinity. Therefore, I can replace it by 1 integral going from minus infinity to plus infinity that is the whole point.

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$$= \frac{P_d}{T} \int_{-\infty}^{\infty} P_T(t - \tilde{t}_0) p(t + z - \tilde{t}_0) d\tilde{t}_0$$

k from $-\infty$ to ∞

$t + z - \tilde{t}_0 = t'$
 $-d\tilde{t}_0 = dt'$

So, I get P_d over T integral, I can combine the summation and integral as minus infinity to infinity $P_T(t - \tilde{t}_0) p(t + z - \tilde{t}_0) d\tilde{t}_0$ into $P_T(t) p(t + z - t)$

tilde into dt , so that is my resulting integral. Now, what I can do is I can further employ a substitution $t + \tau - t_{\text{naught tilde}} = t_{\text{naught prime}}$, which implies $-dt_{\text{naught tilde}} = dt_{\text{naught prime}}$.

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The image shows a whiteboard with handwritten mathematical work. At the top, there is an integral expression:
$$= \frac{1}{T} \int_{-\infty}^{\infty} P_T(t - t_{\text{naught tilde}}) P_T(t + \tau - t_{\text{naught tilde}}) dt_{\text{naught tilde}}$$
 Below this, a substitution is defined:
$$t + \tau - \tilde{t}_0 = t'_0$$

$$-d\tilde{t}_0 = dt'_0$$
 A green arrow labeled "reverses order of integration" points from the original integral to the transformed one. The transformed integral is:
$$= \frac{1}{T} \int_{\infty}^{-\infty} P_T(t'_0 - \tau) P_T(t'_0) (-dt'_0)$$
 The minus sign in the differential is moved outside the integral, and the limits are swapped.

So, therefore, this integral further becomes $\frac{1}{T} \int_{-\infty}^{\infty} P_T(t + \tau - t_{\text{naught tilde}}) P_T(t_{\text{naught tilde}}) dt_{\text{naught tilde}}$ where $t_{\text{naught tilde}}$ is minus infinity this becomes plus infinity when $t_{\text{naught tilde}}$ is infinity $t_{\text{naught prime}}$ becomes minus infinity times well $P_T(t_{\text{naught tilde}} - \tau) = P_T(t_{\text{naught prime}} - \tau)$ into $P_T(t_{\text{naught prime}})$ that is $t_{\text{naught prime}} dt_{\text{naught prime}}$ or $-dt_{\text{naught prime}}$, because $dt_{\text{naught tilde}} - dt_{\text{naught tilde}} = dt_{\text{naught prime}}$. So, this becomes $-dt_{\text{naught prime}}$ and this minus reverses the order of integration. The integral to bring the minus sign outside integral from infinity to minus infinity becomes integral in minus infinity to infinity.

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The whiteboard shows the following derivation:

$$= \frac{P_d}{T} \int_{-\infty}^{\infty} P_T(t'_0) P_T(t'_0 - \tau) dt'_0$$

Does NOT depend on t
only depends on τ
Hence, $x(t)$ is WSS

$$E\{x(t)x(t+\tau)\} = R_{xx}(\tau)$$

So, therefore, we have something that is P_d over capital T integral minus infinity to infinity $P_T t$ naught prime minus tau into P_T or let me write it the other way P_T into t naught prime, t naught prime is simply a dummy p t into t naught prime minus tau. Remember t naught prime is simply a dummy variable, this is t naught. And now if you look at this you will observe something remarkable. If you look at this you will observe that this whole thing now we have successfully avoided or successfully removed, it is dependence on t which is the specific time instant. So, this does not depend on T , it only depends on tau. So, it does not depend on T only depends on tau. Hence, now x t is wide sense stationary. In fact, this is your expected value of x t into x t plus tau which is basically you can call this as R_{xx} of tau depends only on tau.

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$$\int_{-\infty}^{\infty} P_T(t') P_T(t' - \tau) dt'$$

$R_{P_T P_T}(\tau)$
Auto correlation of pulse $P_T(t)$

And in fact, you will observe something very interesting, if you look at this integral minus infinity P T let us write down this integral, integral minus infinity to infinity P T minus or P T or P t naught this is simply a dummy variable. So, this is t naught prime minus t naught prime minus tau or P t naught prime into t naught prime minus tau. If you remember this is nothing but the autocorrelation of the pulse, this is R P T P T of tau, this function is indeed the autocorrelation function of the pulse P T of P T. So, this is the autocorrelation of the pulse P T autocorrelation of the pulse P T. And therefore, this is nothing but P d over T times autocorrelation of the pulse P T.

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$$R_{xx}(\tau) = E\{x(t)x(t+\tau)\}$$
$$= \frac{P_d}{T} R_{P_T P_T}(\tau)$$

Autocorrelation of signal depends on auto-correlation of pulse $P_T(t)$.

And therefore, we finally, have a very elegant result for this that is expected value of $x(t)$ that is $R_{xx}(\tau)$ equals expected value of $x(t)$ into $x(t + \tau)$ which is equal to P_d over τ times simply which is equal to P_d over τ which is simply equal to P_d over T times autocorrelation function $R_{pp}(\tau)$ of the pulse this is the very interesting result. So, what we have seen is that the autocorrelation of this thing is indeed related to the pulse and it depends on the autocorrelation of the pulse it depends on the. So, what we are able to see is something very interesting autocorrelation of the signal depends on autocorrelation function of the pulse $R_{pp}(\tau)$ it depends on the autocorrelation function of the pulse $R_{pp}(\tau)$.

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The image shows a whiteboard with two equations. The top equation is $R_{xx}(z) = \frac{P_d}{T} R_{pp}(z)$. Below it, a double-headed orange arrow points down to the equation $S_{xx}(F) = \frac{P_d}{T} S_{pp}(F)$. A double-headed purple arrow points down from $R_{pp}(z)$ to $S_{pp}(F)$.

In fact, if you can now take the Fourier transform of this autocorrelation function, so we have $R_{xx}(\tau)$ equals P_d over T , P_d over T is a constant times $R_{pp}(\tau)$ of τ depends on the autocorrelation of the pulse. If you take the Fourier transform of this you will get the power spectral density that is you will get $S_{xx}(F)$ which is equal to P_d over T times the Fourier transform of the pulse which is nothing but $S_{pp}(F)$ this is nothing but the autocorrelation. This is nothing, but the energy spectral density of the pulse.

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$$S_{xx}(F) = \frac{P_d}{T} \frac{S_{FF}(F)}{\sigma}$$

PSD
Power Spectral
Density of Transmitted
Signal $x(t)$.

Energy Spectral
Density

Remember if you take the Fourier transform the autocorrelation function of a deterministic signal such as this pulse, you will get the energy spectral density. For a random signal it gives the power spectral density; for a deterministic signal, the Fourier transform of the autocorrelation is the energy spectrum, it is the energy spectral density. And we have this very beautiful relation where the Fourier transform of the pulse that is we have the power spectral density of the transmitted signal the PSD or power spectral density of the transmitted signal that is $x(t)$.

So, what we have able to show that the autocorrelation function that is first we have converted this signal into a wide sense stationary signal. And then we will be able to derive the expression of autocorrelation, we have seen that as a very good elegant expression the autocorrelation function of the digital communication signal simply depends on the autocorrelation function of the pulse. And moreover the power spectral density of the transmitted signal is simply therefore proportional to the energy spectral density of the pulse which is nothing but the Fourier transform of the autocorrelation function of the pulse and that is a very interesting result. So, we have been able to derive now the power spectral density of the signal which gives the distribution of the spectral distribution of the power of the transmitter digital communication signal. So, we stop here and look at this further in the subsequent modules.

Thank you very much.