

Principles of Digital Communications
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Lecture - 51
Minimum Shift Keying - I

We will continue a study of Minimum Shift Key. Let us quickly recollect what we have done earlier. So, we said that minimum shift keying is a form of continuous phase shift keying modulation.

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MSK (Contd...)

$$s(t) = \sqrt{\frac{2E_b}{T_b}} \cos[2\pi f_c t + \theta(t)]$$

$$\theta(t) = \theta(0) \pm \frac{\pi}{2T_b} t \quad (h = \frac{1}{2})$$

$$s(t) = \sqrt{\frac{2E_b}{T_b}} \cos \theta(t) \cos 2\pi f_c t - \sqrt{\frac{2E_b}{T_b}} \sin \theta(t) \sin 2\pi f_c t$$

$$= s_I(t) \cos 2\pi f_c t - s_Q(t) \sin 2\pi f_c t$$

$$s_I(t) = \sqrt{\frac{2E_b}{T_b}} \cos \theta(t) = \pm \sqrt{\frac{2E_b}{T_b}} \cos\left(\frac{\pi t}{2T_b}\right) \quad -T_b \leq t \leq T_b$$

$$s_Q(t) = \sqrt{\frac{2E_b}{T_b}} \sin \theta(t) = \pm \sqrt{\frac{2E_b}{T_b}} \sin\left(\frac{\pi t}{2T_b}\right) \quad 0 \leq t \leq 2T_b$$

$+$: $\theta(0) = 0$
 $-$: $\theta(0) = \pi$
 $+$: $\theta(T_b) = \pi/2$
 $-$: $\theta(T_b) = -\pi/2$

Where your modulated signal $S(t)$ is given by this expression where $\theta(t)$ which is the phase of the modulated signal is specified by this expression, where the deviation ratio is chosen to be half. And on expansion of this expression we showed that it $S(t)$ can be written as shown here, which can be written in the form of in phase and quadrature component. And we had also seen that $s_I(t)$ which is defined by this part of the in phase component is equal to the expression written here.

And we said that this expression is valid over the range minus T_b to plus T_b . For our discussion without loss of generality, we are assuming the bit duration between time t equal to 0 and t equal to T_b . And $s_Q(t)$, similarly we showed that it is equal to the half sine wave and the duration of that half sine wave is over the period 0 to twice T_b .

So, $S_I(t)$ and $S_Q(t)$ wave forms are displaced or offset by T_b seconds and the polarity of this half cosine wave and half sine wave is decided by the phase state at t equal to 0 and at t equal to T_b . So, if $\theta(0)$ is equal to 0 this quantity will be positive and if $\theta(0)$ is π then this quantity becomes negative, and similarly here when $\theta(T_b)$ is equal to π by 2, this quantity is positive and when $\theta(T_b)$ is equal to minus π by 2 or 3 π by 2 modulo 2 π this quantity is same as that. So, then this quantity becomes minus.

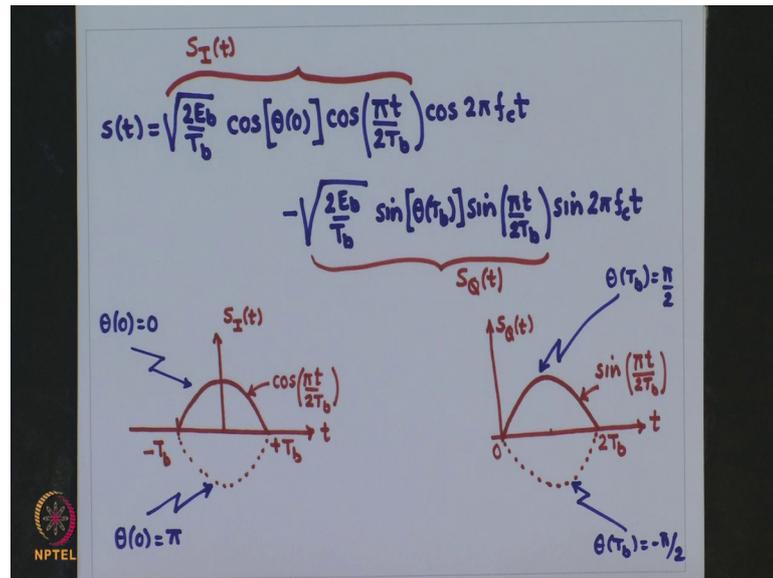
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$\theta(0)$	$\theta(T_b)$	Transmitted symbol
0	$\pi/2$	'1'
π	$\pi/2$	'0'
π	$-\pi/2$	'1'
0	$-\pi/2$	'0'

So, we said that there are 4 states which it could have; the transmitted symbol information is contained in the phase states $\theta(0)$ and $\theta(T_b)$. So, we will have 4 phase states then it is 0 π by 2, we know that one has been transmitted when it is π π by 2 we know 0 is being transmitted and similarly for the other states.

Now, what we want to do is to represent this signal $S(t)$ in terms of orthonormal basis signals. And to do that let us try to understand how the information is being transmitted. The information is being transmitted using this signal $S(t)$ which has 2 parts or 2 components.

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One is a $S_I(t)$ component which modulates $\cos 2\pi f_c t$, and we have $S_Q(t)$ component which modulates $\sin 2\pi f_c t$. The information of the signals which has been transmitted is contained in this phase $\theta(t)$ and the waveform corresponding to this state are $S_I(t)$ and $S_Q(t)$ respectively. And we have seen that this waveform $S_I(t)$ is of the form shown here, which is half cosine and the polarity of this waveform is decided by the phase state $\theta(t)$.

So, if $\theta(t)$ is 0 I get this positive waveform and if it is π , I get it to be negative. So, what this means that we are transmitting 2 states of $\theta(t)$ either $\theta(t)$ can be 0 or it can be π . And to do that we are using these 2 waveforms of opposite polarity and similarly the argument can be extended to $S_Q(t)$. We are transmitting the phase information by using these 2 waveforms and what are the phase information $\theta(t)$ equal to either $\pi/2$ this waveform or $\theta(t)$ is equal to $-\pi/2$ given by this waveform.

So, now this $s(t)$ is composed of 2 waveforms this waveform modulates $\cos 2\pi f_c t$. So, there will be $\cos 2\pi f_c t$ in this and there is $\sin 2\pi f_c t$ in this. So, your $s(t)$ is composed of one waveform which goes from minus T_b to plus T_b and there is another waveform, which goes from 0 to $2T_b$. If you understand this then it is not difficult for us to see that our signal $s(t)$ can be represented in terms of 2 orthonormal basis signals as follows.

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$s(t) = s_1 \phi_1(t) + s_2 \phi_2(t)$

s_1 & s_2 are related to the phase states $\theta(0)$ and $\theta(T_b)$, respectively

→ Phase information at the MSK receiver is explored in $2T_b$ secs. intervals, so that the effective energy collected by this receiver corresponds to observations made during intervals of $2T_b$ secs.

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So, my task is to write $S(t)$ as $s_1 \phi_1(t) + s_2 \phi_2(t)$. s_1 and s_2 are related to the phase states $\theta(0)$ and $\theta(T_b)$ respectively.

So, phase information at the MSK receiver is explored in $2T_b$ seconds in terms. So, that the effective energy collected by this receiver corresponds to observations made during intervals of $2T_b$ seconds. The only difference is that those 2 waveforms which we have they are offset by T_b . So, given this argument it is not very difficult to see that we could choose 2 orthonormal basis signals as follows $\phi_1(t)$ is chosen as this signal.

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$$\phi_1(t) = \sqrt{\frac{2}{T_b}} \cos\left(\frac{\pi t}{2T_b}\right) \cos 2\pi f_c t$$

$$\phi_2(t) = \sqrt{\frac{2}{T_b}} \sin\left(\frac{\pi t}{2T_b}\right) \sin 2\pi f_c t$$

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And the duration of this $\phi_1(t)$ will be between minus T_b to plus T_b correct.

And we will choose another orthonormal basis signal, $\phi_2(t)$ and its duration is going to be from 0 to $2T_b$. Given this 2 basis signal of this duration I should be able to write my $S(t)$ as this form and consequently what I am doing is I am trying to represent my signal $S(t)$, which is given by this expression in terms of orthonormal basis signal. If we choose this has the orthonormal basis signal let us see whether these 2 signals are orthogonal and let us try to find out the energy of each of this signal.

So, first let us try to find out the orthogonality and the common interval between these 2 signal is 0 to $2T_b$ whereas, between minus T_b to 0 $\phi_1(t)$ exists, but $\phi_2(t)$ does not exist and similarly $\phi_2(t)$ exists between T_b to $2T_b$, but $\phi_1(t)$ does not exist over that duration. So, the only common interval for these 2 signals is 0 to T_b let us see what happens between that duration. So, let us try to calculate this integral $\phi_1(t)\phi_2(t)$ over the duration 0 to T_b .

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$$\int_0^{T_b} \phi_1(t)\phi_2(t)dt = ?$$

$$\int_0^{T_b} \cos\left(\frac{\pi t}{2T_b}\right)\sin\left(\frac{\pi t}{2T_b}\right)\cos 2\pi f_c t \sin 2\pi f_c t dt$$

$$= \frac{1}{4} \int_0^{T_b} \sin\left(\frac{\pi t}{T_b}\right) \sin(4\pi f_c t) dt$$

$$= \frac{1}{8} \int_0^{T_b} \left[\cos\left(4\pi f_c - \frac{\pi}{T_b}\right)t - \cos\left(4\pi f_c + \frac{\pi}{T_b}\right)t \right] dt$$

So, I will ignore the factor of root 2 by T_b associated be each of this waveforms without loss of generality and if we write this expression using the trigonometric identities, I can rewrite this expression as 1 by 2 of this and similarly this I can write it as 1 by 2 of sin of $4\pi f_c t$ that is why we get 1 by 4.

Now, again we use the trigonometric identity and simplify this product to this summation of 2 cos terms and if we integrate each of this individually a straightforward integration.

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$$\int_0^{T_b} \phi_1(t) \phi_2(t) dt$$

$$= \frac{1}{8} \left\{ \frac{\sin[(4\pi f_c - \pi/T_b)t]}{(4\pi f_c - \pi/T_b)} \Big|_0^{T_b} - \frac{\sin[(4\pi f_c + \pi/T_b)t]}{(4\pi f_c + \pi/T_b)} \Big|_0^{T_b} \right\}$$

$$= 0$$

f_c , as usual, is an integer multiple of $1/T_b$.

We get these 2 terms and they have to be evaluated over the lower limit and upper limit of the integral and its simple to see that, when I do this both of this terms go to 0 and I will get to 0. Here when I am doing this it is assumed as usual that f_c is an integer multiple of $1/T_b$ ok. So, we can say that $\phi_1(t)$ and $\phi_2(t)$ are orthogonal over the periods of interest.

Now, let us calculate the energy of $\phi_1(t)$ and $\phi_1(t)$ exists between minus T_b to plus T_b .

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$$\begin{aligned} & \frac{2}{T_b} \int_{-T_b}^{T_b} \cos^2\left(\frac{\pi t}{2T_b}\right) \cos^2(2\pi f_c t) dt \\ &= \frac{2}{T_b} \int_{-T_b}^{T_b} \left[\frac{1}{2} + \frac{1}{2} \cos\left(\frac{\pi t}{T_b}\right) \right] \left[\frac{1}{2} + \frac{1}{2} \cos(4\pi f_c t) \right] dt \\ &= \frac{2}{T_b} \times \frac{1}{4} \int_{-T_b}^{T_b} \left[1 + \cos\left(\frac{\pi t}{T_b}\right) + \cos(4\pi f_c t) + \cos\left(\frac{\pi t}{T_b}\right) \cos(4\pi f_c t) \right] dt \\ &= \frac{2}{T_b} \times \frac{1}{4} \times 2T_b = 1 \end{aligned}$$

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So, we will evaluate this expression this expression can be rewritten using the trigonometric identity as this expression and similarly cos squared 2 pi fct can be written using trigonometric identity has this expression, and then we take half out of both. So, I get one fourth and then multiply these 2 terms, I will get all these terms

Now, when we integrate each of this terms these three terms over this period minus 3 be sorry I repeat when we integrate this terms over the period minus T b to plus T b all these three terms will go to 0 correct. And we will get the integration of only this the constant and that will be equal to 2 T b and that is how I get 1. And similarly now it is very easy to show that even phi 2 t has unit energy over the duration 0 to 2 T b correct using the same similar procedure you will get this value ok.

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Similarly, $\int_0^{2T_b} \phi_2^2(t) dt = ?$

$$\frac{2}{T_b} \int_0^{2T_b} \sin^2\left(\frac{\pi t}{2T_b}\right) \sin^2 2\pi f_c t dt = 1$$


So, now we can use this $\phi_1(t)$ and $\phi_2(t)$ as our orthonormal basis signal. So, we will get the coefficients S_1 and S_2 they can be evaluated easily. So, S_1 would be obtained by taking the projection of $s(t)$ over $\phi_1(t)$ and it's very straightforward to see that, your $s(t)$ is given by this quantity.

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Similarly,

$$S_2 = \int_0^{2T_b} s(t) \phi_2(t) dt$$

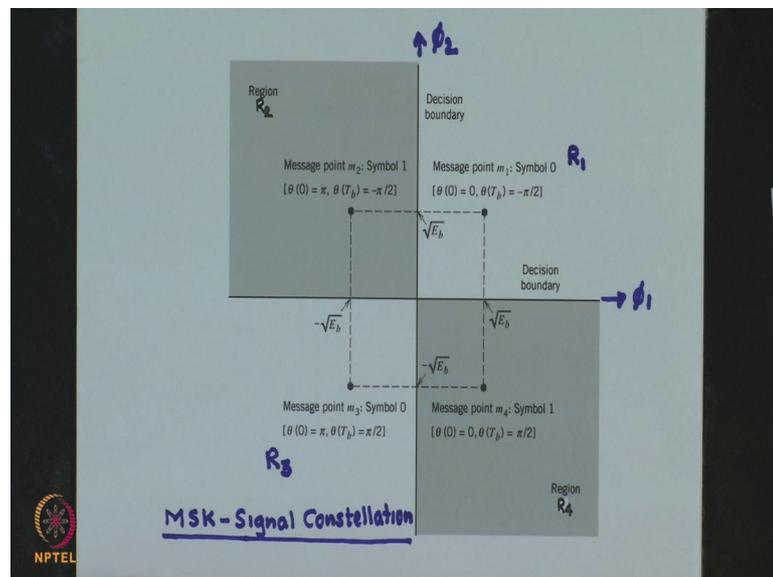
$$= -\sqrt{E_b} \sin[\theta(T_b)]$$


And if I integrate this quantity completely after multiplying by $\phi_1(t)$ it's simple to see that the $\phi_1(t)$ is orthogonal to this term out here, it will go to 0. So, you will get the contribution only from this term, and you will get $\sqrt{E_b} \cos \theta$. So, this is what we get that will be my projection of $s(t)$ or $\phi_1(t)$.

And similarly we can obtain the projection of $s(t)$ over $\phi_2(t)$ and I will get this quantity out here. Please note this presence of minus sign because your $S_2(t)$ was without minus sign and that is why that minus sign comes with S_2 there is only difference, and important to note that both the integrals for S_1 and S_2 are evaluated over the duration of $2T_b$, but the lower limit and the upper limit are offset by duration T_b correct.

So, we see that the signal constellation for this is going to be the 2 dimensional one and it will be as shown in this figure out here.

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. So, in this case n is equal to capital N , which is the dimension of the signal space will be equal to 2, we have 4 points here out here. So, there are 4 message points. So, this signal constellation looks very similar to QPSK case, but the difference is that in the QPSK case, all the 4 message points correspond to 4 different signals. Whereas, for MSK signal constellation though we have 4 message points only there are 2 messages to be transmitted because this 2 message points correspond to symbol 1 and this 2 message point correspond to symbol 0.

Now, given this based on our definition for S_1 and S_2 we can calculate what are this values out here correct. By choosing appropriate θT_b which can be plus $\pi/2$ and minus $\pi/2$ I know S_2 can be either minus root E_b or plus root E_b and similarly for your S_1 plus root E_b or minus root E_b . So, we will get this 4 message points and

we can have the signal space characterization of the MSK has given in this table shown here.

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Signal-space characterization of MSK

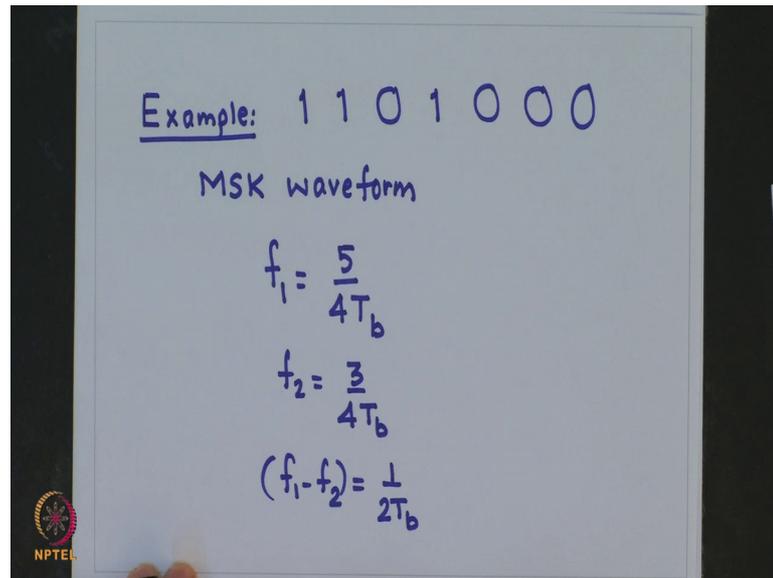
Transmitted Binary Symbol	Phase States (radians)		Coordinates of message points	
	$\theta(0)$	$\theta(T_b)$	s_1	s_2
0	0	$-\pi/2$	$+\sqrt{E_b}$	$+\sqrt{E_b}$
1	π	$-\pi/2$	$-\sqrt{E_b}$	$+\sqrt{E_b}$
0	π	$+\pi/2$	$-\sqrt{E_b}$	$-\sqrt{E_b}$
1	0	$\pi/2$	$+\sqrt{E_b}$	$-\sqrt{E_b}$

So, we have 4 phase states 0 minus pi by 2 pi minus pi by 2 and corresponding to this 4 phase states, we have the corresponding transmitted binary signals. So, see 2 states this and this correspond to one binary signal that is 0 and this 2 states correspond to another binary signal that is 1 and corresponding to this phase states, we get the coordinates of message points S 1 and S 2 as shown in this table.

So, this table summarizes your MSK signal constellation correct. So, given the input data sequence, we can use the entries of the table to derive on a bit by bit basis there are 2 sequences of coefficients required to scale phi 1 t and phi 2 t and once I know how to scale them; that means, I am obtaining the coefficient S 1 and S 2 then I can get my MSK signal as S 1 phi 1 t plus S 2 phi 2 t ok.

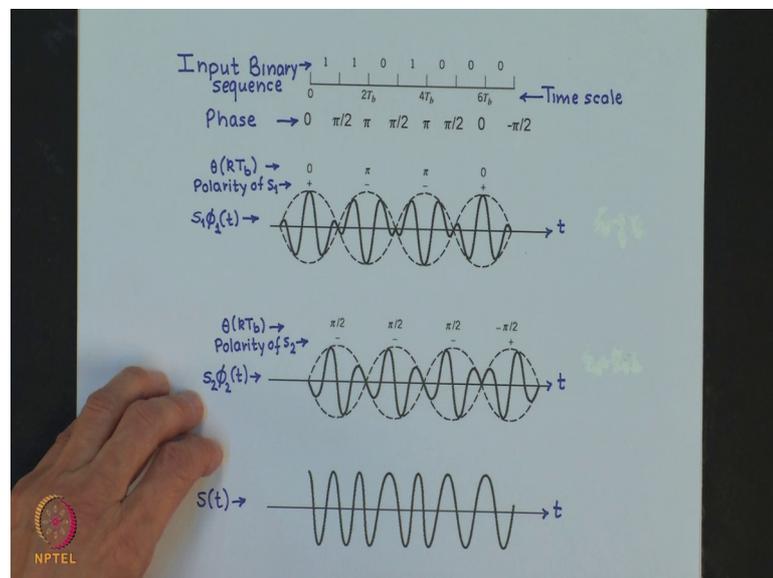
Let us take one example to help us to understand the concepts which we have studied so far. So, I will take one example of binary sequence transmission as 1 1 0 1 0 00 and we want to generate the MSK waveform for this.

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f_1 has been assumed to be $\frac{5}{4T_b}$ and f_2 is $\frac{3}{4T_b}$. So, the difference between the two is $\frac{1}{2T_b}$, which is the condition to be satisfied by an MSK signal and for this case we will generate the MSK signal as follows.

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I have my put binary signal corresponding to this input binary signal I have the phases specified here.

So, when I transmit I will assume that at time T equal to 0 of my phase value is zero. So, at this duration T equal to T_b since I am transmitting 1 my phase will increase by π by

2, again I am transmitting 1. So, my phase will increase by π by 2 it becomes π , I transmit 0 during $2 T_b$ to $3 T_b$. So, my phase decreases by π by 2 and so on we will get the phase states of the sequence at each time instances.

Once we know this phases we can calculate the polarity for S_1 and S_2 as follows. So, we require to the theta for finding out the polarity of S_1 we need to know the phase state θ_0 , $\theta_{2 T_b}$, $\theta_{4 T_b}$ and $\theta_{6 T_b}$. So, if we look at θ_0 , phase is assume to be 0; $\cos 0$ is 1. So, the polarity of S_1 is going to be positive. Then at $2 T_b$ we have $\cos \pi$ is minus 1. So, the polarity of S_1 is going to be negative and similarly for $4 T_b$ and $6 T_b$ we can find out the polarity to be minus and plus respectively.

Given this polarity remember now we can find out our $S_{\phi 1}$, I know my S_1 polarity I calculate my S_1 value the magnitude of S_1 is $\sqrt{E_b}$ correct fine. So, only the polarity has to be known. So, here it is positive. So, we choose this waveform; so your \cos function starts from here. So, this is positive like this, here in this case your polarity has become negative; so your \cos function which we choose half cosine is negative. So, your $\cos 2\pi fct$ becomes negative correct and similarly this is negative and then again this becomes positive fine.

Now, we find out similarly for the polarity of S_2 . So, we have to find out the phase state at θ_{T_b} , $\theta_{3 T_b}$, $\theta_{5 T_b}$ and $\theta_{7 T_b}$ ok. We know the phases now \sin of π by 2 is 1, but remember when we are calculating S_2 there is a minus sign before that. So, this becomes negative. So, here also the state is π by 2 $\sin \pi$ by 2 is 1, but there is a minus sign. So, negative. So, we get the polarity of S_2 . So, here it is going to be negative negative negative and positive and we choose this. So, your sine wave this here it will start like this similarly it will start like this, here also it will start like this correct and here it becomes positive.

So, I am just saying one half. So, similarly is true for the other half of this ok. And then we once we know this $S_{\phi 1}$, I know $S_{\phi 2}$ I add them together I get my S_t and notice that your S_t signal now is a continuous signal without any phase discontinuity. So, we have achieved what we desired that we want phase continuity during the inter bit switching time correct because we do not want the spread of the power spectral density ok.

Now, we will try to find out the modulator for the MSK case, look at the receiver and calculate the probability of error for the same. And finally, we will also evaluate the power spectral density of the minimum shift keying signal. And this will do in the next class.

Thank you.